

Event-Triggered Robust Stabilization of Nonlinear Input-Constrained Systems Using Single Network Adaptive Critic Designs

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Abstract—In this paper, we study the event-triggered robust stabilization problem of nonlinear systems subject to mismatched perturbations and input constraints. First, with the introduction of an infinite-horizon cost function for the auxiliary system, we transform the robust stabilization problem into a constrained optimal control problem. Then, we prove that the solution of the event-triggered Hamilton–Jacobi–Bellman (ETHJB) equation, which arises in the constrained optimal control problem, guarantees original system states to be uniformly ultimately bounded (UUB). To solve the ETHJB equation, we present a single network adaptive critic design (SN-ACD). The critic network used in the SN-ACD is tuned through the gradient descent method. By using Lyapunov method, we demonstrate that all the signals in the closed-loop auxiliary system are UUB. Finally, we provide two examples, including the pendulum system, to validate the proposed event-triggered control strategy.

Index Terms—Adaptive critic designs (ACDs), adaptive dynamic programming (ADP), event-triggered control (ETC), input constraints, neural network (NN), reinforcement learning (RL).

I. INTRODUCTION

ADAPTIVE critic designs (ACDs) have emerged as effective tools to solve optimal control problems over the past several decades [1]–[3]. The typical structure applied to implement ACDs is the actor-critic architecture, where the actor performs a control policy to environment (or controlled systems), and the critic offers an estimation of the value of that control policy and gives feedback information to the actor. In the computational intelligence community, adaptive dynamic programming (ADP) [4] and reinforcement learning (RL) [5] are nearly in the same spirits as ACDs (e.g., all of them have similar implementation architectures). Thus, they are often regarded as synonyms for ACDs. In this paper, we view ADP

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and RL as a kind of ACDs. The early contributors to ADP and RL included Werbos [6] and Sutton and Barto [7]. After that, many scholars showed their interest in ADP and RL. Thus, all kinds of ADP and RL methods were developed, such as local value/policy iterative ADP [8], [9], goal representation ADP [10], robust ADP [11], [12], single network ACDs (SN-ACDs) [13], online RL [14], [15], off-policy RL [16], and manifold RL [17].

In recent years, applications of ACDs to study robust stabilization problems have been extensively reported [18]–[22]. In the existing literature, the robust controllers for nonlinear systems are generally obtained by solving H_∞ optimal control problems or nonlinear zero-sum games under the framework of ACDs. Nevertheless, a limitation of solving H_∞ optimal control problems or zero-sum games is that one needs to make sure the existence of saddle points. Unfortunately, it is challenging to judge whether the saddle point of nonlinear systems exists or not. To avoid this difficulty, Lin and Brandt [23] introduced an indirect method, which aimed at converting the robust control problem into an H_2 optimal control problem. Then, one was able to derive the robust controller for nonlinear systems by solving the H_2 optimal control problem. Recently, the indirect method together with ACDs was proposed by Adhyaru *et al.* [24] to design the robust controller for uncertain nonlinear input-constrained systems. After that, Mu *et al.* [25] used the indirect method and ACDs together to derive a robust tracking control strategy for nonlinear systems subject to matched uncertainties. By using a similar method as [25], Qu *et al.* [26] obtained a decentralized tracking control of large-scale nonlinear systems with matched interconnections. An important difference between [25] and [26] was that [26] did not require the initial admissible control while implementing the proposed robust control scheme. Later, Zhang *et al.* [27] extended the work of [26] to design an optimal guaranteed cost sliding mode controller for constrained nonlinear systems with matched/mismatched disturbances. In all the above mentioned literature, the robust control strategies were implemented in the time-triggering mechanism. In other words, the robust control schemes were implemented periodically. According to [28], the time-triggered control schemes often had difficulties in handling the control problems with the conditions that there were only finite computation bandwidths as well as the limited communication resources.

To overcome these difficulties, many event-triggered control (ETC) approaches have been introduced [29]–[32]. Unlike

79 the time-triggered controller, the event-triggered controller is
 80 updated aperiodically. Specifically, the event-triggered controller is updated only when the deviation between the system
 81 state and the desired value crosses a prescribed threshold.
 82 Due to this characteristic, the ETC strategies can overcome
 83 the shortcomings of the time-triggered control schemes [28].
 84 Thus, many studies on robust ETC methods were reported.
 85 Wang *et al.* [33] presented an event-based robust controller
 86 for uncertain nonlinear systems via the SN-ACD. After that,
 87 with the combination of the SN-ACD and the concurrent
 88 learning technique, Zhang *et al.* [34] developed a robust
 89 ETC of nonlinear systems with mismatched disturbances.
 90 Compared with [33], Zhang *et al.* [34] relaxed the persistence
 91 of excitation (PE) condition. Recently, by applying ACDs
 92 to solve the event-triggered H_∞ optimal control problems,
 93 Mu *et al.* [35] and Zhang *et al.* [36] obtained robust ETC
 94 strategies for nonlinear systems, respectively. Owing to the
 95 existence of the aforementioned limitation in solving H_∞
 96 optimal control problems, these robust ETC schemes usually
 97 encountered difficulties in engineering applications. On the
 98 other hand, due to physical characteristics of actuators in engi-
 99 neering industries, it is necessary to take actuator saturations
 100 (i.e., input constraints) into account. To address this problem,
 101 Wang *et al.* [37] studied the constrained robust ETC problem
 102 of nonlinear input-affine systems with *matched* uncertainties
 103 using ACDs. However, to the best of our knowledge, there are
 104 few studies developing the robust ETC scheme for nonlinear
 105 input-constrained systems subject to *mismatched* perturba-
 106 tions, especially without using the H_∞ control theory [38].
 107 This motivates this paper.

108 In this paper, a robust ETC strategy is developed for nonlin-
 109 ear input-constrained systems with mismatched perturbations.
 110 First, by constructing an infinite-horizon cost function for the
 111 auxiliary system, the robust stabilization problem is converted
 112 into a constrained optimal control problem. Then, it is proved
 113 that the solution of the event-triggered Hamilton–Jacobi–
 114 Bellman (ETHJB) equation, arising in the constrained optimal
 115 control problem, keeps original system states uniformly ulti-
 116 mately bounded (UUB). To solve the ETHJB equation, the
 117 SN-ACD is proposed. The critic network used in the SN-ACD
 118 is updated by using the gradient descent method. Finally, uni-
 119 form ultimate boundedness of all the signals in the closed-loop
 120 auxiliary system is demonstrated via Lyapunov method.

121 The novelties of this paper include three aspects.

- 122 1) Different from [34] updating the augmented control in
 123 an mechanism regarded as the combination of time-
 124 triggering and event-triggering mechanisms (ETMs), this
 125 paper tunes the augmented control only in the ETM.
 126 Hence, the developed control scheme has an advantage
 127 in decreasing the computational burden.
- 128 2) Unlike [35] and [36] solving the event-triggered H_∞
 129 optimal control problems, this paper obtains the robust
 130 ETC via an indirect method. Thus, the present method
 131 relaxes the requirement of judging the existence of the
 132 saddle point, which is an indispensable procedure in
 133 solving H_∞ optimal control problems.
- 134 3) This paper extends the work of [37] to develop a robust
 135 ETC strategy for nonlinear input-constrained systems

136 with *mismatched* perturbations. Generally, robust control
 137 methods for nonlinear systems with *matched* distur-
 138 bances are not applicable to those systems with *mis-
 139 match* disturbances (*note*: the definitions of systems
 140 with *mismatch* disturbances and systems with *match*
 141 disturbances can refer to [23]). Furthermore, when con-
 142 sidering input constraints, it increases the difficulty in
 143 making such an extension.

144 It is worth emphasizing here that the knowledge of system
 145 dynamics [i.e., $f(x)$ and $g(x)$ in system (1) (*note*: see
 146 Section II-A)] is required to be known. Actually, by using
 147 a similar fuzzy technique proposed in [39], this condition can
 148 be removed. For simplicity, in this paper we assume that the
 149 information of system dynamics is available.

150 The rest of this paper is structured as follows. After
 151 briefly presenting problem descriptions and preliminaries in
 152 Section II, we propose the robust ETC scheme in Section III.
 153 Then, after discussing the stability analysis in Section IV, we
 154 provide two examples to validate the established theoretical
 155 results in Section V. Finally, several concluding remarks and
 156 future works are given in Section VI.

157 *Notation*: \mathbb{R} , \mathbb{N} , and \mathbb{N}^+ denote the sets of real numbers,
 158 non-negative integers, and positive integers, respectively. \mathbb{R}^m
 159 and $\mathbb{R}^{n \times m}$ denote the spaces of real m -vectors and $n \times m$ real
 160 matrices, respectively. I_n is the identity matrix of dimension
 161 $n \times n$. \top is the transposition symbol. $\|\alpha\| = \sqrt{\sum_{i=1}^n |\alpha_i|^2}$ is
 162 the Euclidean norm of the vector $\alpha = (\alpha_1, \alpha_2, \dots, \alpha_n)^\top \in \mathbb{R}^n$.
 163 Ω is a subset of \mathbb{R}^n , i.e., $\Omega \subset \mathbb{R}^n$. $\|A\|$ denotes the Frobenius-
 164 norm of the matrix $A \in \mathbb{R}^{n \times m}$. $V_x^* = \partial V^*(x)/\partial x$ is the partial
 165 derivative of $V^*(x)$ with respect to $x \in \mathbb{R}^n$.

II. PROBLEM DESCRIPTION AND PRELIMINARIES

A. Problem Description

166 Consider the continuous-time nonlinear system with a mis-
 167 matched perturbation given in the form

$$\dot{x}(t) = f(x(t)) + g(x(t))u(t) + k(x(t))d(x(t)) \quad (1)$$

168 where $x(t) \in \mathbb{R}^n$ is the state, $u(t) \in \mathfrak{U}$ is the control input, $\mathfrak{U} =$
 169 $\{(u_1, u_2, \dots, u_m) \in \mathbb{R}^m : |u_i| \leq \beta, i = 1, 2, \dots, m\}$, $\beta > 0$ is
 170 the upper bound, $f(x) \in \mathbb{R}^n$, $g(x) \in \mathbb{R}^{n \times m}$, and $k(x) \in \mathbb{R}^{n \times p}$
 171 (*note*: $k(x) \neq g(x)$ when $p = m$) are known smooth functions,
 172 and $d(x) \in \mathbb{R}^p$ is an *uncertain* perturbation. Here, $x_0 = x(0)$
 173 is the initial state.

174 *Assumption 1*: System (1) is controllable. Meanwhile,
 175 $x = 0$ is the equilibrium point of system (1) when letting
 176 $u(t) = 0$ and $d(x(t)) = 0$ for all $t \geq 0$.

177 *Assumption 2*: The control matrix $g(x)$ is bounded as $0 <$
 178 $g(x) \leq g_M$ ($\forall x \in \mathbb{R}^n$) with $g_M \in \mathbb{R}$ the positive constant.
 179 Meanwhile, there exist non-negative functions $\zeta_M(x) \in \mathbb{R}$ and
 180 $d_M(x) \in \mathbb{R}$ such that, for all $x \in \mathbb{R}^n$

$$\|g^+(x)k(x)d(x)\| \leq \ell_M(x) \quad \text{and} \quad \|d(x)\| \leq d_M(x) \quad (185)$$

186 with $g^+(x)$ the Moore–Penrose pseudo-inverse of $g(x)$. In
 187 addition, $\ell_M(0) = 0$, $d(0) = 0$, and $d_M(0) = 0$.

188 This paper aims at finding an appropriate state feedback
 189 controller to stabilize system (1). Owing to the existence of

system uncertainties, it is challengeable to design the stabilizing controller directly. To address this issue, we will convert the robust control problem of system (1) into a constrained optimal control problem of the auxiliary system.

194 B. Hamilton–Jacobi–Bellman Equation Related to Auxiliary 195 Systems

196 Divide the term $k(x)d(x)$ into the following two parts:

$$197 \quad k(x)d(x) = g(x)g^+(x)k(x)d(x) + h(x)d(x) \quad (2)$$

198 with

$$199 \quad h(x) = (I_n - g(x)g^+(x))k(x). \quad (3)$$

200 According to [40], the auxiliary system corresponding to (1)
201 can be described as

$$202 \quad \dot{x} = f(x) + g(x)u + h(x)v \quad (4)$$

203 where $u \in \mathcal{U}$ and $v \in \mathbb{R}^p$ is the auxiliary control.

204 The infinite-horizon cost function for system (4) is given by

$$205 \quad V^{u,v}(x(t)) = \int_t^\infty (\Psi(x(s)) + r(x(s), u(s), v(s)))ds \quad (5)$$

206 where $\Psi(x) = 2\ell_M^2(x) + \rho d_M^2(x)$, $\rho \in \mathbb{R}$ is a positive constant,
207 and

$$208 \quad r(x, u, v) = x^T Qx + \mathcal{W}(u) + \rho v^T v$$

209 with $Q \in \mathbb{R}^{n \times n}$ the positive definite matrix and $\mathcal{W}(u) \in \mathbb{R}$ the
210 semipositive definite function.

211 To overcome the bounded control, we define $\mathcal{W}(u)$ as [41]

$$212 \quad \mathcal{W}(u) = 2\beta \sum_{i=1}^m \int_0^{u_i} \psi^{-1}(\zeta/\beta) d\zeta \quad (6)$$

213 where $\psi(\cdot)$ is a bounded monotonic function with $\psi(0) = 0$.
214 Meanwhile, $\psi(\cdot)$ is an odd function with its derivative
215 bounded. Since $\psi^{-1}(\cdot)$ is a monotonic odd function, $\mathcal{W}(u)$
216 given in (6) is semipositive definite. In this paper, we let $\psi(\cdot)$
217 be the hyperbolic tangent function, i.e., $\psi(\cdot) = \tanh(\cdot)$.

218 Let $\mathcal{A}(\Omega)$ be the set of admissible control [5] defined on
219 Ω . Then, the optimal value of (5) is formulated as

$$220 \quad V^*(x) = \min_{u, v \in \mathcal{A}(\Omega)} V^{u,v}(x). \quad (7)$$

221 If $V^*(x)$ is continuously differentiable, then its derivative
222 satisfies

$$223 \quad (V_x^*)^T (f(x) + g(x)u + h(x)v) \\ 224 \quad + \Psi(x) + x^T Qx + \mathcal{W}(u) + \rho v^T v = 0.$$

225 According to [4], the Hamiltonian for V_x^* , u , and v can be
226 defined as

$$227 \quad H(x, V_x^*, u, v) = (V_x^*)^T (f(x) + g(x)u + h(x)v) + \Psi(x) \\ 228 \quad + x^T Qx + \mathcal{W}(u) + \rho v^T v. \quad (8)$$

229 Then, $V^*(x)$ can be obtained by solving the Hamilton–Jacobi–
230 Bellman (HJB) equation

$$231 \quad \min_{u, v \in \mathcal{A}(\Omega)} H(x, V_x^*, u, v) = 0 \quad (9)$$

with $V^*(0) = 0$. Based on the stationary condition [42, Th. 5.8], we can therefore derive the closed-form expressions of optimal control and optimal auxiliary control as follows [4]:

$$232 \quad u^*(x) = -\beta \tanh\left(\frac{1}{2\beta} g^T(x) V_x^*\right) \quad (10) \quad 236$$

$$233 \quad v^*(x) = -\frac{1}{2\rho} h^T(x) V_x^*. \quad (11) \quad 237$$

238 From (8)–(11), we can rewrite the HJB equation as

$$239 \quad (V_x^*)^T f(x) + \Psi(x) + x^T Qx + \mathcal{W}\left(-\beta \tanh\left(\frac{1}{2\beta} g^T(x) V_x^*\right)\right) \\ 240 \quad - \beta (V_x^*)^T g(x) \tanh\left(\frac{1}{2\beta} g^T(x) V_x^*\right) - \left\| \frac{1}{2\sqrt{\rho}} h^T(x) V_x^* \right\|^2 = 0 \quad (12) \quad 241$$

242 with $V^*(0) = 0$. According to [30], (12) is the time-triggered
243 HJB equation.

244 Similar to [43], it can be proved that the robust controller for
245 system (1) is able to be obtained by solving (12). However,
246 owing to the use of time-triggered formulations, the robust
247 control strategy is developed in the time-triggering mechanism.
248 As mentioned in [44], the time-triggered control algorithms
249 generally have low efficiency of using the limited communica-
250 tion resources between actuators and systems. In addition,
251 they often involve high-computational burdens. To overcome
252 the two deficiencies, we will develop a robust ETC scheme
253 for system (1).

III. ROBUST ETC STRATEGY

255 In this section, we first describe the robust stabilization of
256 system (1) in the ETM. Specifically, we prove that the robust
257 ETC of (1) can be obtained by solving an ETHJB equation.
258 Then, we use the SN-ACD to solve the ETHJB equation.

A. Robust Stabilization in the ETM

259 Let $\{t_j\}_{j=0}^\infty$ (note: $t_j < t_{j+1}, j \in \mathbb{N}$) be the sequence of trig-
260 gering instants, where t_j denotes the j th triggering instant. The
261 system state is sampled at the triggering instant t_j , and the
262 sampled state is written as
263

$$264 \quad \bar{x}_j = x(t_j) \quad j \in \mathbb{N}.$$

265 Since there generally exists an error between the sampled state
266 \bar{x}_j and the current state $x(t)$, we define the error as follows:
267

$$268 \quad e_j(t) = \bar{x}_j - x(t) \quad \forall t \in [t_j, t_{j+1}). \quad (13) \quad 269$$

270 From the expression $e_j(t)$ given in (13), we can judge whether
271 an event is triggered or not. Specifically, if the event is trig-
272 gered at instant $t = t_j$, then $e_j(t_j) = 0$. Based on the sampled
273 state, we can obtain the ETC law $u(\bar{x}_j)$, which is executed at
274 the triggering instant t_j . By using the zero-order hold tech-
275 nique [28], the control sequence $\{u(\bar{x}_j)\}_{j=0}^\infty$ can generate a
276 continuous-time input signal $\mu(\bar{x}_j, t)$, i.e.,

$$277 \quad \mu(\bar{x}_j, t) = u(\bar{x}_j) = u(x(t_j)) \quad \forall t \in [t_j, t_{j+1}). \quad 278$$

Let the above mentioned ETM be applied to $u^*(x)$ given in (10). Then, the optimal ETC law for system (4) with the cost function (5) can be obtained as [30] (for all $t \in [t_j, t_{j+1}]$)

$$\mu^*(\bar{x}_j, t) = u^*(\bar{x}_j) = -\beta \tanh\left(\frac{1}{2\beta} g^\top(\bar{x}_j) V_{\bar{x}_j}^*\right) \quad (14)$$

with $V_{\bar{x}_j}^* = (\partial V^*(x)/\partial x)|_{x=\bar{x}_j}$.

Similarly, applying the aforementioned ETM to $v^*(x)$ given in (11), we can derive the optimal auxiliary ETC law as

$$\vartheta^*(\bar{x}_j, t) = v^*(\bar{x}_j) = -\frac{1}{2\rho} h^\top(\bar{x}_j) V_{\bar{x}_j}^* \quad \forall t \in [t_j, t_{j+1}]. \quad (15)$$

Remark 1: For brevity, in subsequent discussion we write $\mu^*(\bar{x}_j, t)$ and $\vartheta^*(\bar{x}_j, t)$ as $\mu^*(\bar{x}_j)$ and $\vartheta^*(\bar{x}_j)$, respectively.

Before continuing the discussion, we give the following assumption used in [30] and [45].

Assumption 3: $u^*(x)$ has the Lipschitz property on Ω . That is, there exists a Lipschitz constant $K_{u^*} > 0$ such that, for all $x, \bar{x}_j \in \Omega$

$$\|u^*(x) - u^*(\bar{x}_j)\| \leq K_{u^*} \|x - \bar{x}_j\| = K_{u^*} \|e_j\|.$$

Remark 2: By using Remark 1 and (14), we can write $\mu^*(\bar{x}_j) = u^*(\bar{x}_j)$. Thus, Assumption 3 implies

$$\|u^*(x) - \mu^*(\bar{x}_j)\| \leq K_{u^*} \|e_j\| \quad (16)$$

for all $x, \bar{x}_j \in \Omega$.

Theorem 1: Let Assumptions 1–3 be valid and let $V^*(x)$ be a solution of the HJB equation (12). Then, the optimal ETC law $\mu^*(\bar{x}_j)$ given in (14) can ensure the closed-loop system (1) to be stable in the sense of uniform ultimate boundedness only if $v^*(x)$ given in (11) satisfies

$$\|v^*(x(t))\|^2 \leq \lambda_{\min}(Q) \|x(t)\|^2 \quad \forall t \geq t_s \quad (17)$$

where $t_s \geq 0$ is a threshold, and provided that the triggering condition is given by

$$\|e_j\|^2 \leq \frac{(1-2\rho)\lambda_{\min}(Q)}{4K_{u^*}^2} \|x\|^2 \triangleq \|e_T\|^2 \quad (18)$$

with $0 < \rho < 1/2$ the design parameter and e_T the triggering threshold.

Proof: We take $V^*(x)$ as the Lyapunov function candidate. From the expression $V^*(x)$ given as in (7), we can deduce that $V^*(x) > 0$ for $x \neq 0$ and $V^*(x) = 0 \Leftrightarrow x = 0$, i.e., $V^*(x)$ is positive definite.

By differentiating $V^*(x)$ along the solution of $\dot{x} = f(x) + g(x)\mu^*(\bar{x}_j) + k(x)d(x)$ and using (2), we have [note: $\dot{V}^*(x)$ denotes $dV^*(x(t))/dt$]

$$\begin{aligned} \dot{V}^*(x) &= (V_x^*)^\top (f(x) + g(x)\mu^*(\bar{x}_j) + k(x)d(x)) \\ &= (V_x^*)^\top (f(x) + g(x)u^*(x) + h(x)v^*(x)) \\ &\quad + (V_x^*)^\top g(x)(\mu^*(\bar{x}_j) - u^*(x)) \\ &\quad + (V_x^*)^\top g(x)g^\top(x)k(x)d(x) \\ &\quad + (V_x^*)^\top h(x)(d(x) - v^*(x)) \end{aligned} \quad (19)$$

with $h(x)$ defined as in (3).

On the other hand, from (8) and (9), we obtain

$$\begin{aligned} (V_x^*)^\top (f(x) + g(x)u^*(x) + h(x)v^*(x)) \\ = -\Psi(x) - x^\top Qx - \mathcal{W}(u^*(x)) - \rho \|v^*(x)\|^2. \end{aligned} \quad (20)$$

Meanwhile, from (10) and (11), we find

$$\begin{cases} (V_x^*)^\top g(x) = -2\beta (\tanh^{-1}(u^*(x)/\beta))^\top \\ (V_x^*)^\top h(x) = -2\rho (v^*(x))^\top. \end{cases} \quad (21)$$

Substituting (20) and (21) into (19), it follows:

$$\begin{aligned} \dot{V}^*(x) &= -\Psi(x) - x^\top Qx - \mathcal{W}(u^*(x)) + \rho \|v^*(x)\|^2 \\ &\quad + 2\beta \underbrace{\left(\tanh^{-1}(u^*(x)/\beta) \right)^\top (u^*(x) - \mu^*(\bar{x}_j))}_{\pi_1} \\ &\quad - 2\beta \underbrace{\left(\tanh^{-1}(u^*(x)/\beta) \right)^\top g^+(x)k(x)d(x)}_{\pi_2} \\ &\quad - 2\rho \underbrace{(v^*(x))^\top d(x)}_{\pi_3}. \end{aligned} \quad (22)$$

By using Young's inequality $2y^\top z \leq \varrho \|y\|^2 + \|z\|^2/\varrho$ ($\varrho > 0$) and (16), we can see that π_1 in (22) implies (note: $\varrho = 1/2$)

$$\begin{aligned} \pi_1 &\leq \frac{\beta^2}{2} \left\| \tanh^{-1}(u^*(x)/\beta) \right\|^2 + 2 \|u^*(x) - \mu^*(\bar{x}_j)\|^2 \\ &\leq \frac{\beta^2}{2} \sum_{i=1}^m \left(\tanh^{-1}(u_i^*(x)/\beta) \right)^2 + 2K_{u^*}^2 \|e_j\|^2. \end{aligned} \quad (23)$$

Similarly, by using the above mentioned Young's inequality and Assumption 2, we can find that π_2 and π_3 in (22) yield (note: $\varrho = 1/2$ and $\varrho = 1$, respectively)

$$\begin{aligned} \pi_2 &\leq \frac{\beta^2}{2} \left\| \tanh^{-1}(u^*(x)/\beta) \right\|^2 + 2 \|g^+(x)k(x)d(x)\|^2 \\ &\leq \frac{\beta^2}{2} \sum_{i=1}^m \left(\tanh^{-1}(u_i^*(x)/\beta) \right)^2 + 2\ell_M^2(x) \end{aligned} \quad (24)$$

$$\pi_3 \leq \rho \|v^*(x)\|^2 + \rho \|d(x)\|^2 \leq \rho \|v^*(x)\|^2 + \rho d_M^2(x). \quad (25)$$

From [46] (note: see the proof of [46, Th. 1]), we know

$$\begin{aligned} \mathcal{W}(u^*) &= 2\beta \sum_{i=1}^m \int_0^{u_i^*(x)} \tanh^{-1}(\zeta/\beta) d\zeta \\ &= \beta^2 \sum_{i=1}^m \left(\tanh^{-1}(u_i^*(x)/\beta) \right)^2 \\ &\quad - 2\beta^2 \sum_{i=1}^m \int_0^{\tanh^{-1}(u_i^*(x)/\beta)} \tau_i \tanh^2(\tau_i) d\tau_i. \end{aligned} \quad (26)$$

Observing that $\Psi(x) = 2\ell_M^2(x) + \rho d_M^2(x)$ and using (23)–(26), we can conclude that (22) yields

$$\begin{aligned} \dot{V}^*(x) &\leq -x^\top Qx + 2\rho \|v^*(x)\|^2 + 2K_{u^*}^2 \|e_j\|^2 \\ &\quad + 2\beta^2 \underbrace{\sum_{i=1}^m \int_0^{\tanh^{-1}(u_i^*(x)/\beta)} \tau_i \tanh^2(\tau_i) d\tau_i}_{\mathcal{L}(x)}. \end{aligned} \quad (27)$$

According to the proof of [46, Th. 1], we know that $\mathcal{L}(x)$ given in (27) is a bounded function. To facilitate subsequent discussion, we denote that $\|\mathcal{L}(x)\| \leq \epsilon_M$,

where $\epsilon_M > 0$ is a constant. Then, (27) can be further rewritten as

$$\begin{aligned} \dot{V}^*(x) &\leq -\lambda_{\min}(Q)\|x\|^2 + 2\rho\|v^*(x)\|^2 + 2K_{u^*}^2\|e_j\|^2 + \epsilon_M \\ &= -2\rho\left(\lambda_{\min}(Q)\|x\|^2 - \|v^*(x)\|^2\right) + 2K_{u^*}^2\|e_j\|^2 \\ &\quad - (1-2\rho)\lambda_{\min}(Q)\|x\|^2 + \epsilon_M \end{aligned} \quad (28)$$

with $\lambda_{\min}(Q)$ the minimum eigenvalue of Q .

Thus, if (17) and (18) hold, then (28) yields

$$\dot{V}^*(x) \leq -\frac{(1-2\rho)\lambda_{\min}(Q)}{2}\|x\|^2 + \epsilon_M. \quad (29)$$

Therefore, from (29), we can find that $\dot{V}^*(x) < 0$ only when $x(t)$ is out of the following set:

$$\Omega_x = \left\{ x: \|x\| \leq \sqrt{\frac{2\epsilon_M}{(1-2\rho)\lambda_{\min}(Q)}} \right\}.$$

Then, uniform ultimate boundedness of the states of system (1) is guaranteed by using Lyapunov extension theorem [47]. Specifically, this indicates that $\mu^*(\bar{x}_j)$ keeps the closed-loop system (1) stable in the sense of uniform ultimate boundedness. Meanwhile, the ultimate bound is $\sqrt{2\epsilon_M}/((1-2\rho)\lambda_{\min}(Q))$. ■

Remark 3: According to (18), the triggering instant t_j can be calculated. Then, it is possible to obtain the minimal intersample time $(\Delta t_j)_{\min}$, where $\Delta t_j = t_{j+1} - t_j, j \in \mathbb{N}$. However, if there exists $(\Delta t_j)_{\min} = 0$, then the Zeno behavior occurs [48]. In this circumstance, $\mu^*(\bar{x}_j)$ has to be redesigned. Fortunately, $(\Delta t_j)_{\min} > 0, j \in \mathbb{N}$, under Assumption 1 (note: since similar proofs have been provided in [36] and [49], we omit the proof here). In this paper, simulation results provided in Section V also show that $(\Delta t_j)_{\min} > 0, j \in \mathbb{N}$.

To obtain the optimal ETC law $\mu^*(\bar{x}_j)$, we need to solve the ETHJB equation, which is derived by substituting (14) and (15) into (9). That is,

$$\begin{aligned} &(V_x^*)^T f(x) - \beta(V_x^*)^T g(x) \tanh\left(\frac{1}{2\beta}g^T(\bar{x}_j)V_{\bar{x}_j}^*\right) \\ &\quad - \frac{1}{2\rho}(V_x^*)^T h(x)h^T(\bar{x}_j)V_{\bar{x}_j}^* + \Psi(x) + x^T Qx \\ &\quad + \mathcal{W}\left(-\beta \tanh\left(\frac{1}{2\beta}g^T(\bar{x}_j)V_{\bar{x}_j}^*\right)\right) + \left\| \frac{1}{2\sqrt{\rho}}h^T(\bar{x}_j)V_{\bar{x}_j}^* \right\|^2 = 0. \end{aligned} \quad (30)$$

Generally, it is rather hard to solve the ETHJB equation (30) analytically [50]. To conquer the difficulty, we present the SN-ACD to approximately solve (30).

B. SN-ACD for Solving the ETHJB Equation

The approximation theorem [51] guarantees that $V^*(x)$ given in (7) can be represented via a critic network over Ω as

$$V^*(x) = \omega_c^T \sigma_c(x) + \varepsilon_c(x)$$

where $\omega_c \in \mathbb{R}^{\tilde{n}_c}$ is the ideal weight vector, $\sigma_c(x) = [\sigma_{c1}(x), \sigma_{c2}(x), \dots, \sigma_{c\tilde{n}_c}(x)]^T \in \mathbb{R}^{\tilde{n}_c}$ is the basis function vector, $\sigma_{cl}(x), l = 1, 2, \dots, \tilde{n}_c$, are continuously differentiable

functions with $\sigma_{cl}(0) = 0$, $\tilde{n}_c \in \mathbb{N}^+$ is the number of basis functions, and $\varepsilon_c(x) \in \mathbb{R}$ is the approximation error.

Differentiating $V^*(x)$ at the sampled state \bar{x}_j , we have

$$V_{\bar{x}_j}^* = \nabla \sigma_c^T(\bar{x}_j) \omega_c + \nabla \varepsilon_c(\bar{x}_j) \quad \forall t \in [t_j, t_{j+1}) \quad (31)$$

where $\nabla \sigma_c(\bar{x}_j) = (\partial \sigma_c(x)/\partial x)|_{x=\bar{x}_j}$ and $\nabla \varepsilon_c(\bar{x}_j) = (\partial \varepsilon_c(x)/\partial x)|_{x=\bar{x}_j}$.

Substituting (31) into (14), we can rewrite $\mu^*(\bar{x}_j)$ as

$$\mu^*(\bar{x}_j) = -\beta \tanh(\mathcal{A}_1(\bar{x}_j)) + \varepsilon_{\mu^*}(\bar{x}_j) \quad \forall t \in [t_j, t_{j+1}) \quad (32)$$

where

$$\mathcal{A}_1(\bar{x}_j) = \frac{1}{2\beta}g^T(\bar{x}_j)\nabla \sigma_c^T(\bar{x}_j)\omega_c \quad (403)$$

and $\varepsilon_{\mu^*}(\bar{x}_j) = -(1/2)(\mathbf{1} - \tanh^2(\xi))g^T(\bar{x}_j)\nabla \varepsilon_c(\bar{x}_j)$ with $\mathbf{1} = [1, \dots, 1]^T \in \mathbb{R}^m$ and ξ chosen between $\mathcal{A}_0(\bar{x}_j)$ (note: $\mathcal{A}_0(\bar{x}_j) = (1/(2\beta))g^T(\bar{x}_j)V_x^*$) and $\mathcal{A}_1(\bar{x}_j)$.

Similarly, by using (31), $\vartheta^*(\bar{x}_j)$ given in (15) can be represented as (for all $t \in [t_j, t_{j+1})$)

$$\vartheta^*(\bar{x}_j) = -\frac{1}{2\rho}h^T(\bar{x}_j)\nabla \sigma_c^T(\bar{x}_j)\omega_c + \varepsilon_{\vartheta^*}(\bar{x}_j) \quad (33)$$

with $\varepsilon_{\vartheta^*}(\bar{x}_j) = -(1/2\rho)h^T(\bar{x}_j)\nabla \varepsilon_c(\bar{x}_j)$.

Remark 4: The difference between $\varepsilon_{\mu^*}(\bar{x}_j)$ given in (32) and $\varepsilon_{\vartheta^*}(\bar{x}_j)$ given in (33) is caused by control constraints (note: u is constrained while v is unconstrained). To make (32) be better for understanding, we provide the detailed process of deriving $\varepsilon_{\mu^*}(\bar{x}_j)$ as follows. Let

$$\mathcal{T}(\mathcal{A}_{\varpi}(x)) = -\beta \tanh(\mathcal{A}_{\varpi}(x)), \quad \varpi = 0, 1. \quad (416)$$

Then, applying the mean value theorem [42, Th. 5.10] to $\mathcal{T}(\mathcal{A}_{\varpi}(x))$, we obtain (note: $\mathcal{A}_0(x) = (1/(2\beta))g^T(x)V_x^*$)

$$\begin{aligned} \mathcal{T}(\mathcal{A}_0(x)) - \mathcal{T}(\mathcal{A}_1(x)) &= -\beta(\tanh(\mathcal{A}_0(x)) - \tanh(\mathcal{A}_1(x))) \\ &= -\frac{1}{2}(\mathbf{1} - \tanh^2(\xi))g^T(x)\nabla \varepsilon_c(x) \end{aligned} \quad (419)$$

with ξ chosen between $\mathcal{A}_0(x)$ and $\mathcal{A}_1(x)$. By using (31), we find that (14) yields

$$\begin{aligned} \mu^*(\bar{x}_j) &= -\beta \tanh(\mathcal{A}_0(\bar{x}_j)) \\ &= \mathcal{T}(\mathcal{A}_1(\bar{x}_j)) + (\mathcal{T}(\mathcal{A}_0(\bar{x}_j)) - \mathcal{T}(\mathcal{A}_1(\bar{x}_j))) \\ &= -\beta \tanh(\mathcal{A}_1(\bar{x}_j)) - \frac{1}{2}(\mathbf{1} - \tanh^2(\xi))g^T(\bar{x}_j)\nabla \varepsilon_c(\bar{x}_j). \end{aligned} \quad (423)$$

Hence, we can obtain the expression $\varepsilon_{\mu^*}(\bar{x}_j)$ given as in (32).

In general, the ideal weight vector ω_c is unavailable. Thus, we cannot implement $\mu^*(\bar{x}_j)$ given in (32). To handle this issue, we replace ω_c with the current estimated weight vector $\hat{\omega}_c$ in the critic network. Then, the approximation value function can be formulated as

$$\hat{V}(x) = \hat{\omega}_c^T \sigma_c(x). \quad (432)$$

The derivative of $\hat{V}(x)$ at the sampled state \bar{x}_j is

$$\hat{V}_{\bar{x}_j} = \nabla \sigma_c^T(\bar{x}_j) \hat{\omega}_c. \quad (434)$$

Replacing $V_{\bar{x}_j}^*$ in (14) with $\hat{V}_{\bar{x}_j}$, we derive the estimated value of $\mu^*(\bar{x}_j)$ as

$$\hat{\mu}(\bar{x}_j) = -\beta \tanh(\mathcal{A}_2(\bar{x}_j)) \quad \forall t \in [t_j, t_{j+1}) \quad (35)$$

438 where

$$439 \quad \mathcal{A}_2(\bar{x}_j) = \frac{1}{2\beta} g^\top(\bar{x}_j) \nabla \sigma_c^\top(\bar{x}_j) \hat{\omega}_c.$$

440 By the same token, the estimated value of $\vartheta^*(\bar{x}_j)$ given in (33)
441 can be obtained as

$$442 \quad \hat{\vartheta}(\bar{x}_j) = -\frac{1}{2\rho} h^\top(\bar{x}_j) \nabla \sigma_c^\top(\bar{x}_j) \hat{\omega}_c \quad \forall t \in [t_j, t_{j+1}). \quad (36)$$

443 Substituting $\hat{V}(x)$, $\hat{\mu}(\bar{x}_j)$, and $\hat{\vartheta}(\bar{x}_j)$ into (8), we can see that
444 the approximation Hamiltonian is

$$445 \quad \hat{H}(x, \hat{V}_x, \hat{\mu}(\bar{x}_j), \hat{\vartheta}(\bar{x}_j)) \\ 446 \quad = \hat{\omega}_c^\top \nabla \sigma_c(x) (f(x) + g(x) \hat{\mu}(\bar{x}_j) + h(x) \hat{\vartheta}(\bar{x}_j)) \\ 447 \quad + \Psi(x) + x^\top Q x + \mathcal{W}(\hat{\mu}^\top(\bar{x}_j)) + \rho \|\hat{\vartheta}(\bar{x}_j)\|^2.$$

448 Observe that (9) implies

$$449 \quad H(x, V_x^*, \mu^*(\bar{x}_j), \vartheta^*(\bar{x}_j)) = 0.$$

450 Thus, the error of Hamiltonian can be formulated as

$$451 \quad e_c = \hat{H}(x, \hat{V}_x, \hat{\mu}(\bar{x}_j), \hat{\vartheta}(\bar{x}_j)) - H(x, V_x^*, \mu^*(\bar{x}_j), \vartheta^*(\bar{x}_j)) \\ 452 \quad = \hat{\omega}_c^\top \phi + \Psi(x) + x^\top Q x + \mathcal{W}(\hat{\mu}^\top(\bar{x}_j)) + \rho \|\hat{\vartheta}(\bar{x}_j)\|^2 \quad (37)$$

453 where $\phi = \nabla \sigma_c(x) (f(x) + g(x) \hat{\mu}(\bar{x}_j) + h(x) \hat{\vartheta}(\bar{x}_j))$.

454 To ensure e_c given in (37) to be sufficiently small, we use
455 the gradient descent method to minimize the target function
456 $E = (1/2)e_c^\top e_c$. Then, the weight update rule for the critic
457 network is obtained as

$$458 \quad \dot{\hat{\omega}}_c = -\frac{l_c}{(1 + \phi^\top \phi)^2} \frac{\partial E}{\partial \hat{\omega}_c} \\ 459 \quad = -\frac{l_c \phi}{(1 + \phi^\top \phi)^2} e_c \quad \forall t \in [t_j, t_{j+1}) \quad (38)$$

460 with e_c defined as in (37), $l_c \in \mathbb{R}^n$ the positive parameter, and
461 $(1 + \phi^\top \phi)^{-2}$ the normalization term.

462 Let the weight estimation error of the critic network be
463 $\tilde{\omega}_c = \omega_c - \hat{\omega}_c$. Then, from (38), we can see that the weight
464 estimation error dynamics of the critic network satisfies [30]

$$465 \quad \dot{\tilde{\omega}}_c = -l_c \varphi \varphi^\top \tilde{\omega}_c + \frac{l_c \varphi}{1 + \phi^\top \phi} \varepsilon_H \quad \forall t \in [t_j, t_{j+1}) \quad (39)$$

466 where $\varphi = \phi / (1 + \phi^\top \phi)^2$ and $\varepsilon_H = -\nabla \varepsilon_c^\top(x) (f(x) +$
467 $g(x) \hat{\mu}(\bar{x}_j) + h(x) \hat{\vartheta}(\bar{x}_j))$ is the residual error.

468 From the ETM introduced in Section III-A, we can find
469 that the closed-loop system (4) is a hybrid system. Let the
470 augmented state be $\mathcal{X} = [x^\top, \bar{x}_j^\top, \tilde{\omega}_c^\top]^\top$. Then, we can describe
471 the hybrid dynamical system as follows.

472 1) *Continuous Dynamics:*

$$473 \quad \dot{\mathcal{X}}(t) = \begin{bmatrix} f(x) + \mathcal{F}(x, \bar{x}_j) \\ 0 \\ -l_c \varphi \varphi^\top \tilde{\omega}_c + \frac{l_c \varphi \varepsilon_H}{(1 + \phi^\top \phi)^2} \end{bmatrix} \quad \forall t \in [t_j, t_{j+1}) \quad (40)$$

474 where

$$475 \quad \mathcal{F}(x, \bar{x}_j) = -\beta g(x) \tanh\left(\frac{1}{2\beta} g^\top(\bar{x}_j) \nabla \sigma_c^\top(\bar{x}_j) \hat{\omega}_c\right) \\ 476 \quad - \frac{1}{2\rho} h(x) h^\top(\bar{x}_j) \nabla \sigma_c^\top(\bar{x}_j) \hat{\omega}_c.$$

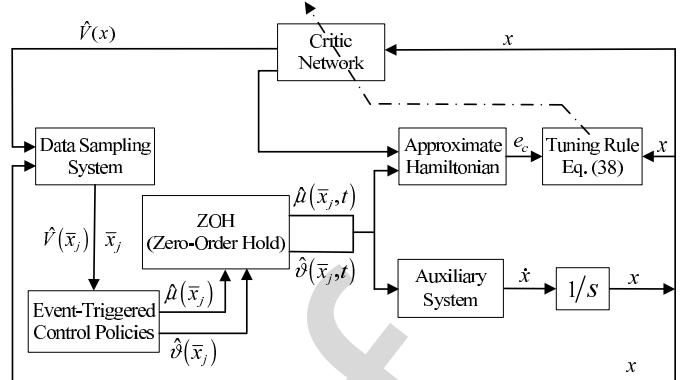


Fig. 1. Block diagram of the proposed ETC strategy.

2) *Discrete Dynamics:*

$$477 \quad \mathcal{X}(t^+) = \mathcal{X}(t) + \begin{bmatrix} 0 \\ \bar{x}_j - x(t) \\ 0 \end{bmatrix} \quad t = t_{j+1} \quad (41) \quad 478$$

479 where $\mathcal{X}(t^+) = \lim_{\eta \rightarrow 0^+} \mathcal{X}(t + \eta)$ with $\eta \in (0, t_{j+1} - t_j)$. 479

480 Based on the above mentioned analyses, we present the 480
481 block diagram of the proposed ETC strategy in Fig. 1. 481

IV. STABILITY ANALYSIS

482 Before proving stabilities of systems (40) and (41), we pro- 483
484 vide two assumptions introduced in [4] and [52], respectively. 484

485 *Assumption 4:* The derivative of the basic function vector is 485
486 bounded as $\|\nabla \sigma_c(x)\| \leq b_{\sigma_c}$ ($\forall x \in \Omega$), where b_{σ_c} is a positive 486
487 constant. In addition, there exist positive constants $b_{\varepsilon_{\mu^*}}$, $b_{\varepsilon_{\vartheta^*}}$, 487
488 and b_{ε_H} such the approximation errors $\varepsilon_{\mu^*}(\bar{x}_j)$, $\varepsilon_{\vartheta^*}(\bar{x}_j)$, and the 488
489 residual error ε_H bounded as $\|\varepsilon_{\mu^*}(\bar{x}_j)\| \leq b_{\varepsilon_{\mu^*}}$, $\|\varepsilon_{\vartheta^*}(\bar{x}_j)\| \leq 489$
490 $b_{\varepsilon_{\vartheta^*}}$, and $\|\varepsilon_H\| \leq b_{\varepsilon_H}$ ($\forall \bar{x}_j, x \in \Omega$), respectively. 490

491 Similar to (16) imposed on $u^*(x)$, we present the following 491
492 assumption for $v^*(x)$. 492

493 *Assumption 5:* $v^*(x)$ satisfies the Lipschitz condition on Ω . 493
494 That is, for all $x, \bar{x}_j \in \Omega$, there exists a Lipschitz constant 494
495 $K_{v^*} > 0$ such that 495

$$496 \quad \|v^*(x) - v^*(\bar{x}_j)\| \leq K_{v^*} \|x - \bar{x}_j\| = K_{v^*} \|e_j\|. \quad 496$$

497 Let

$$498 \quad G(\mathcal{A}_\kappa(x)) = \beta \tanh(\mathcal{A}_\kappa(x)), \quad \kappa = 1, 2 \quad (42) \quad 498$$

499 where $\mathcal{A}_1(x) = (1/(2\beta)) g^\top(x) \nabla \sigma_c^\top(x) \omega_c$ and $\mathcal{A}_2(x) = 499$
500 $(1/(2\beta)) g^\top(x) \nabla \sigma_c^\top(x) \hat{\omega}_c$. Then, using Taylor's theorem [42],
501 it follows: 501

$$502 \quad G(\mathcal{A}_1(x)) = G(\mathcal{A}_2(x)) + \frac{\partial G(\mathcal{A}_2)}{\partial \mathcal{A}_2}(\mathcal{A}_1(x) - \mathcal{A}_2(x)) \quad 502$$

$$503 \quad + O((\mathcal{A}_1(x) - \mathcal{A}_2(x))^2) \quad 503$$

$$504 \quad = G(\mathcal{A}_2(x)) + \frac{1}{2} (I_m - \mathcal{B}(\mathcal{A}_2(x))) g^\top(x) \quad 504$$

$$505 \quad \times \nabla \sigma_c^\top(x) \tilde{\omega}_c + O((\mathcal{A}_1(x) - \mathcal{A}_2(x))^2) \quad 505$$

506 with $\mathcal{B}(\mathcal{A}_2(x)) = \text{diag}\{\tanh^2(\mathcal{A}_{2i}(x))\}$, $i = 1, 2, \dots, m$, and
507 the high-order term $O((\mathcal{A}_1(x) - \mathcal{A}_2(x))^2)$. 507

508 *Lemma 1:* The high-order term given in (43) is bounded as
 509 $\|O((\mathcal{A}_1(x) - \mathcal{A}_2(x))^2)\| \leq 2\sqrt{m} + g_M b_{\sigma_c} \|\tilde{\omega}_c\|. \quad (44)$

510 *Proof:* From (43), we can find

511
$$\|O((\mathcal{A}_1(x) - \mathcal{A}_2(x))^2)\| \leq \|G(\mathcal{A}_1(x))\| + \|G(\mathcal{A}_2(x))\|$$

 512
$$+ \frac{1}{2} \|I_m - \mathcal{B}(\mathcal{A}_2(x))\| \|g(x)\|$$

 513
$$\times \|\nabla \sigma(x)\| \|\tilde{\omega}_c\|. \quad (45)$$

514 Since $\|\tanh(x)\| \leq 1$ for all $x \in \mathbb{R}^n$, we can conclude that
 515 $\|G(\mathcal{A}_\kappa(x))\| = (\sum_{i=1}^m \tanh^2(\mathcal{A}_{\kappa i}(x)))^{1/2} \leq \sqrt{m}$, $\kappa = 1, 2$, and
 516 $\|I_m - \mathcal{B}(\mathcal{A}_2(x))\| \leq 2$. Then, using Assumptions 2 and 4, we
 517 can see that (45) yields (44). ■

518 *Theorem 2:* Consider auxiliary system (4) associated with
 519 the ETHJB equation (30). Let Assumptions 1–5 be valid
 520 and take the control policies proposed as in (35) and (36).
 521 Suppose that the initial control for system (4) is admissible
 522 and the weight tuning rule for the critic network is described
 523 as (38). Then, the closed-loop system (4) and the weight esti-
 524 mation error $\tilde{\omega}_c$ are UUB only if the following event-triggering
 525 condition holds:

526
$$\|e_j\|^2 \leq \frac{(1-2\gamma)\lambda_{\min}(Q)}{4K_{\max}^2} \|x\|^2 \triangleq \|\bar{e}_T\|^2 \quad (46)$$

527 where $K_{\max} = \max\{K_{u^*}, K_{v^*}\}$, $0 < \gamma < 1/2$ is a design
 528 parameter, and \bar{e}_T is the triggering threshold, and provided
 529 that the following inequality holds:

530
$$\frac{l_c}{2} \lambda_{\min}(\varphi \varphi^\top) - (16g_M^2 + h_M^2/\rho^2) b_{\sigma_c}^2 > 0 \quad (47)$$

531 where $\lambda_{\min}(\varphi \varphi^\top)$ denotes the minimum eigenvalue of $\varphi \varphi^\top$, φ
 532 satisfies the PE condition, and h_M is the bound of $h(x)$.

533 *Proof:* We take the Lyapunov function candidate as

534
$$L(t) = \underbrace{V^*(\bar{x}_j)}_{L_1(t)} + \underbrace{V^*(x(t))}_{L_2(t)} + \underbrace{(1/2)\tilde{\omega}_c^\top \tilde{\omega}_c}_{L_3(t)}.$$

535 Since the closed-loop system (4) is a hybrid system, we present
 536 the stability analysis from following two circumstances.

537 *Situation I:* Events are not triggered, i.e., $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$.
 538 Then, we have $\dot{L}_1(t) = \dot{V}^*(\bar{x}_j) = 0$.

539 Taking the derivative of $L_2(t)$ and using the trajectory
 540 generated from $\dot{x} = f(x) + g(x)\hat{\mu}(\bar{x}_j) + h(x)\hat{\vartheta}(\bar{x}_j)$, we have

541
$$\begin{aligned} \dot{L}_2(t) &= (V_x^*)^\top (f(x) + g(x)\hat{\mu}(\bar{x}_j) + h(x)\hat{\vartheta}(\bar{x}_j)) \\ &= (V_x^*)^\top (f(x) + g(x)u^*(x) + h(x)v^*(x)) \\ &\quad + (V_x^*)^\top g(x)(\hat{\mu}(\bar{x}_j) - u^*(x)) \\ &\quad + (V_x^*)^\top h(x)(\hat{\vartheta}(\bar{x}_j) - v^*(x)). \end{aligned} \quad (48)$$

545 Substituting (20) and (21) into (48), it follows:

546
$$\begin{aligned} \dot{L}_2(t) &= -\Psi(x) - x^\top Qx - \rho \|u^*(x)\|^2 - \mathcal{W}(u^*(x)) \\ &\quad + \underbrace{2\beta \left(\tanh^{-1}(u^*(x)/\beta) \right)^\top (u^*(x) - \hat{\mu}(\bar{x}_j))}_{\Xi} \\ &\quad + 2\rho (v^*(x))^\top (v^*(x) - \hat{\vartheta}(\bar{x}_j)). \end{aligned} \quad (49)$$

549 Applying Young's inequality $2y^\top z \leq \|y\|^2 + \|z\|^2$ to Ξ in (49),
 550 we obtain

551
$$\begin{aligned} \Xi &\leq \beta^2 \left\| \tanh^{-1}(u^*(x)/\beta) \right\|^2 + \|u^*(x) - \hat{\mu}(\bar{x}_j)\|^2 \\ &= \beta^2 \sum_{i=1}^m \left(\tanh^{-1}(u_i^*(x)/\beta) \right)^2 + \|u^*(x) - \hat{\mu}(\bar{x}_j)\|^2. \end{aligned} \quad (52)$$

553 Then, by using (26), we can see that

554
$$-\mathcal{W}(u^*(x)) + \Xi \leq \mathfrak{f}(x) + \|u^*(x) - \hat{\mu}(\bar{x}_j)\|^2 \quad (50)$$

555 with $\mathfrak{f}(x)$ defined as in (27). As indicated in the proof
 556 of Theorem 1, $\mathfrak{f}(x)$ is bounded as $\|\mathfrak{f}(x)\| \leq \epsilon_M$. Thus,
 557 combining (49) and (50), we have

558
$$\begin{aligned} \dot{L}_2(t) &\leq -\Psi(x) - x^\top Qx - \rho \left\| \hat{\vartheta}(\bar{x}_j) \right\|^2 + \epsilon_M \\ &\quad + \underbrace{\|u^*(x) - \hat{\mu}(\bar{x}_j)\|^2}_{\Lambda_1} + \underbrace{\rho \|v^*(x) - \hat{\vartheta}(\bar{x}_j)\|^2}_{\Lambda_2}. \end{aligned} \quad (51)$$

560 Applying the inequality $\|y+z\|^2 \leq 2\|y\|^2 + 2\|z\|^2$ to Λ_1 in (51)
 561 and using Assumption 3 as well as (32) and (35), it follows:

562
$$\begin{aligned} \Lambda_1 &= \|(u^*(x) - \mu^*(\bar{x}_j)) + (\mu^*(\bar{x}_j) - \hat{\mu}(\bar{x}_j))\|^2 \\ &\leq 2\|\mu^*(\bar{x}_j) - \hat{\mu}(\bar{x}_j)\|^2 + 2\|u^*(x) - \mu^*(\bar{x}_j)\|^2 \\ &\leq 2\|G(\mathcal{A}_2(\bar{x}_j)) - G(\mathcal{A}_1(\bar{x}_j)) + \varepsilon_{\mu^*}(\bar{x}_j)\|^2 + 2K_{u^*}^2 \|e_j\|^2 \end{aligned} \quad (52)$$

566 where $G(\mathcal{A}_\kappa(\bar{x}_j)) = G(\mathcal{A}_\kappa(x))|_{x=\bar{x}_j}$ with $G(\mathcal{A}_\kappa(x))$ defined
 567 as in (42). By using (43) and Lemma 1 as well as Young's
 568 inequality, we derive

569
$$\begin{aligned} &2\|G(\mathcal{A}_2(\bar{x}_j)) - G(\mathcal{A}_1(\bar{x}_j)) + \varepsilon_{\mu^*}(\bar{x}_j)\|^2 \\ &\leq 2\left(2g_M b_{\sigma_c} \|\tilde{\omega}_c\| + 2\sqrt{m} + b_{\varepsilon_{\mu^*}}\right)^2 \\ &\leq 16g_M^2 b_{\sigma_c}^2 \|\tilde{\omega}_c\|^2 + 4a_0^2 \end{aligned} \quad (53)$$

572 with $a_0 = 2\sqrt{m} + b_{\varepsilon_{\mu^*}}$.

573 Thus, combining (52) and (53), it follows:

574
$$\Lambda_1 \leq 2K_{u^*}^2 \|e_j\|^2 + 16g_M^2 b_{\sigma_c}^2 \|\tilde{\omega}_c\|^2 + 4a_0^2. \quad (54)$$

575 Similar to the process of calculating Λ_1 , we obtain

576
$$\begin{aligned} \Lambda_2 &= \|(v^*(x) - \vartheta^*(\bar{x}_j)) + (\vartheta^*(\bar{x}_j) - \hat{\vartheta}(\bar{x}_j))\|^2 \\ &\leq 2\|\vartheta^*(\bar{x}_j) - \hat{\vartheta}(\bar{x}_j)\|^2 + 2\|v^*(x) - \vartheta^*(\bar{x}_j)\|^2 \\ &\leq 2\left\| -\frac{h^\top(\bar{x}_j)}{2\rho} \nabla \sigma_c^\top(\bar{x}_j) \tilde{\omega}_c + \varepsilon_{\vartheta^*}(\bar{x}_j) \right\|^2 + 2K_{v^*}^2 \|e_j\|^2 \\ &\leq 2K_{v^*}^2 \|e_j\|^2 + \left(h_M^2 b_{\sigma_c}^2 / \rho^2 \right) \|\tilde{\omega}_c\|^2 + 4b_{\varepsilon_{\vartheta^*}}^2. \end{aligned} \quad (55)$$

580 Note that $\Psi(x)$ given in (5) and $\rho \|\hat{\vartheta}(\bar{x}_j)\|^2$ are non-negative
 581 functions. Then, from (51), (54), and (55), we get

582
$$\begin{aligned} \dot{L}_2(t) &\leq -\lambda_{\min}(Q) \|x\|^2 + 4K_{\max}^2 \|e_j\|^2 \\ &\quad + \left(16g_M^2 + h_M^2 / \rho^2 \right) b_{\sigma_c}^2 \|\tilde{\omega}_c\|^2 + 4a_0^2 + 4b_{\varepsilon_{\vartheta^*}}^2 + \epsilon_M \end{aligned} \quad (56)$$

584 with $K_{\max} = \max\{K_{u^*}, K_{v^*}\}$.

586 Taking the time derivative of $L_3(t)$ and using the weight
 587 estimation error dynamics (39), it follows:

$$588 \quad \dot{L}_3(t) = -l_c \tilde{\omega}_c^\top \varphi \varphi^\top \tilde{\omega}_c + l_c \frac{\tilde{\omega}_c^\top \varphi}{1 + \phi^\top \phi} \varepsilon_H. \quad (57)$$

589 Noticing that $1 + \phi^\top \phi \geq 1$ and using the above men-
 590 tioned Young's inequality, we develop the last term in
 591 (57) as

$$592 \quad \frac{l_c}{1 + \phi^\top \phi} \tilde{\omega}_c^\top \varphi \varepsilon_H \leq \frac{l_c}{2(1 + \phi^\top \phi)} \left(\tilde{\omega}_c^\top \varphi \varphi^\top \tilde{\omega}_c + \varepsilon_H^\top \varepsilon_H \right)$$

$$593 \quad \leq \frac{l_c}{2} \tilde{\omega}_c^\top \varphi \varphi^\top \tilde{\omega}_c + \frac{l_c}{2} \varepsilon_H^\top \varepsilon_H.$$

594 Then, (57) yields

$$595 \quad \dot{L}_3(t) \leq -\frac{l_c}{2} \tilde{\omega}_c^\top \varphi \varphi^\top \tilde{\omega}_c + \frac{l_c}{2} \varepsilon_H^\top \varepsilon_H$$

$$596 \quad \leq -\frac{l_c}{2} \lambda_{\min}(\varphi \varphi^\top) \|\tilde{\omega}_c\|^2 + \frac{l_c}{2} b_{\varepsilon_H}^2. \quad (58)$$

597 By using (56) and (58), we can see that

$$598 \quad \dot{L}(t) \leq -2\gamma \lambda_{\min}(Q) \|x\|^2 - (1 - 2\gamma) \lambda_{\min}(Q) \|x\|^2$$

$$599 \quad - \left(\frac{l_c}{2} \lambda_{\min}(\varphi \varphi^\top) - (16g_M^2 + h_M^2/\rho^2) b_{\sigma_c}^2 \right) \|\tilde{\omega}_c\|^2$$

$$600 \quad + 4K_{\max}^2 \|e_j\|^2 + 4a_0^2 + 4b_{\varepsilon_{\theta^*}}^2 + \epsilon_M. \quad (59)$$

601 If the condition (46) holds, then (59) yields

$$602 \quad \dot{L}(t) \leq -2\gamma \lambda_{\min}(Q) \|x\|^2 + 4a_0^2 + 4b_{\varepsilon_{\theta^*}}^2 + \epsilon_M$$

$$603 \quad - \left(\frac{l_c}{2} \lambda_{\min}(\varphi \varphi^\top) - (16g_M^2 + h_M^2/\rho^2) b_{\sigma_c}^2 \right) \|\tilde{\omega}_c\|^2.$$

604 Under the condition (47), we can find that $\dot{L}(t) < 0$
 605 only if we can ensure one of the following inequalities
 606 holds:

$$607 \quad \|x\| > \sqrt{\frac{4a_0^2 + 4b_{\varepsilon_{\theta^*}}^2 + \epsilon_M}{2\gamma \lambda_{\min}(Q)}} \triangleq c_1 \quad (60)$$

608 or

$$609 \quad \|\tilde{\omega}_c\| > \sqrt{\frac{8a_0^2 + 8b_{\varepsilon_{\theta^*}}^2 + 2\epsilon_M}{l_c \lambda_{\min}(\varphi \varphi^\top) - (32g_M^2 + 2h_M^2/\rho^2) b_{\sigma_c}^2}} \triangleq c_2. \quad (61)$$

610 Then, uniform ultimate boundedness of both $x(t)$ and $\tilde{\omega}_c$
 611 is obtained by using Lyapunov extension theorem [47].
 612 Meanwhile, the ultimate bounds of $x(t)$ and $\tilde{\omega}_c$ are c_1 given
 613 in (60) and c_2 given in (61), respectively.

614 *Situation II:* Events are triggered, i.e., $t = t_j, j \in \mathbb{N}$. Then,
 615 we take the difference of Lyapunov function candidate $L(t_j)$
 616 into account, that is

$$617 \quad \Delta L(t_j) = V^*(\bar{x}_{j+1}) - V^*(\bar{x}_j) + \Pi(x(t_j^+), \bar{x}_j)$$

618 where $x(t_j^+) = \lim_{\eta \rightarrow 0^+} x(t_j + \eta)$ with $\eta \in (0, t_{j+1} - t_j)$, and

$$619 \quad \Pi(x(t_j^+), \bar{x}_j) = V^*(x(t_j^+)) - V^*(x(t_j))$$

$$620 \quad + \frac{1}{2} \tilde{\omega}_c^\top(t_j^+) \tilde{\omega}_c(t_j^+) - \frac{1}{2} \tilde{\omega}_c^\top(t_j) \tilde{\omega}_c(t_j).$$

Suppose that $x(t) \notin E_{c_1} = \{x(t) \in \mathbb{R}^n \mid \|x(t)\| \leq c_1\}$ or $\tilde{\omega}_c \notin E_{c_2} = \{\tilde{\omega}_c \in \mathbb{R}^n \mid \|\tilde{\omega}_c\| \leq c_2\}$. Then, from
 621 Situation I, we have $dL(t)/dt < 0 \forall t \in [t_j, t_{j+1})$. That is,
 622 $L(t)$ is strictly monotonically decreasing on $[t_j, t_{j+1})$. Thus, it
 623 implies

$$624 \quad L(t_j) > L(t_j + \eta) \quad \forall \eta \in (0, t_{j+1} - t_j). \quad (62)$$

625 Taking $\eta \rightarrow 0^+$ over both sides of (62), we can conclude

$$626 \quad L(t_j) \geq \lim_{\eta \rightarrow 0^+} L(t_j + \eta) = L(t_j^+). \quad (628)$$

627 Thus, we have

$$628 \quad V^*(x(t_j)) + \frac{1}{2} \tilde{\omega}_c^\top(t_j) \tilde{\omega}_c(t_j) \geq V^*(x(t_j^+)) + \frac{1}{2} \tilde{\omega}_c^\top(t_j^+) \tilde{\omega}_c(t_j^+). \quad (630)$$

631 From (63), it follows:

$$632 \quad \Pi(x(t_j^+), \bar{x}_j) \leq 0. \quad (64)$$

633 On the other hand, the uniform ultimate boundedness of the
 634 state $x(t)$ in Situation I implies

$$635 \quad V^*(\bar{x}_{j+1}) \leq V^*(\bar{x}_j). \quad (65)$$

636 Therefore, under the condition that $x(t) \notin E_{c_1} = \{x(t) \in \mathbb{R}^n \mid \|x(t)\| \leq c_1\}$ (or $\tilde{\omega}_c \notin E_{c_2} = \{\tilde{\omega}_c \in \mathbb{R}^n \mid \|\tilde{\omega}_c\| \leq c_2\}$),
 637 we can conclude that $\Delta L(t_j) < 0$ based on (64) and (65).
 638 According to [53], uniform ultimate boundedness of $x(t)$
 639 and $\tilde{\omega}_c$ is guaranteed. Meanwhile, the ultimate bounds of
 640 $x(t)$ and $\tilde{\omega}_c$ are c_1 given in (60) and c_2 given in (61),
 641 respectively. \blacksquare

V. SIMULATION STUDY

644 This section presents two examples to show the effec-
 645 tiveness and applicabilities of the established theoretical
 646 results.

A. Example 1: Nonlinear Plants

647 We study the continuous-time nonlinear system with a
 648 mismatched perturbation given by

$$649 \quad \dot{x}_1 = -x_1 + x_2 + \delta_1 x_1 \cos\left(\frac{1}{x_2 + \delta_2}\right) + \delta_3 x_2 \sin(x_1 x_2) \quad (651)$$

$$650 \quad \dot{x}_2 = -0.5(x_1 + x_2) + 0.5x_2 \sin^2(x_1) + \sin(x_1)u \quad (66)$$

651 where $x = [x_1, x_2]^\top \in \mathbb{R}^2$ is the state, $u \in \{u \in \mathbb{R} : |u| \leq \beta\}$ is
 652 the control input, and $\delta_\zeta, \zeta = 1, 2, 3$, are unknown parameters.
 653 In this example, we set $\beta = 2$ and randomly choose $\delta_1 \in$
 654 $[-\sqrt{2}/2, \sqrt{2}/2]$, $\delta_2 \in [-100, 100]$, and $\delta_3 \in [-\sqrt{2}/2, \sqrt{2}/2]$.
 655 The initial state is $x_0 = [0.5, -0.5]^\top$.

656 The mismatched perturbation in system (66) is

$$657 \quad d(x) = \delta_1 x_1 \cos\left(\frac{1}{x_2 + \delta_2}\right) + \delta_3 x_2 \sin(x_1 x_2). \quad (659)$$

658 After making some computations, we obtain $\|d(x)\| \leq \|x\|$.
 659 Hence, we can let $d_M(x) = \|x\|$. Since $g(x) = [0, \sin(x_1)]^\top$
 660 and $k(x) = [1, 0]^\top$, we have $g(x)g^+(x)k(x) = 0$. Then,
 661

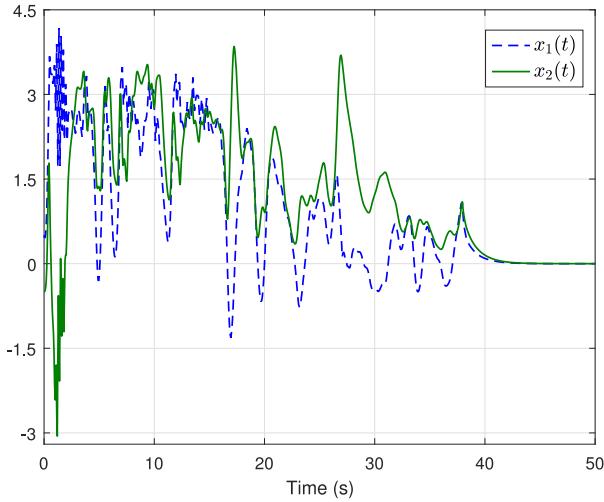


Fig. 2. Evolution of auxiliary system states $x_1(t)$ and $x_2(t)$ in Example 1.

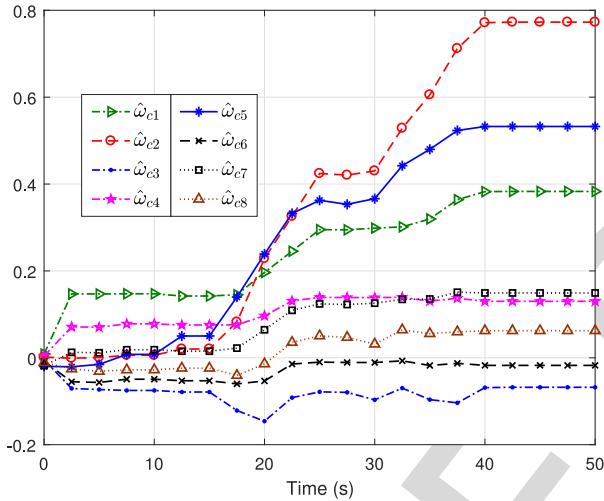


Fig. 3. Convergence of critic network weight vector $\hat{\omega}_c$ in Example 1.

663 based on (4), the auxiliary system related to (66) can be
 664 proposed as

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -x_1 + x_2 \\ -0.5(x_1 + x_2 - x_2 \sin^2(x_1)) \end{bmatrix} + g(x)u + k(x)v \quad (67)$$

667 with $g(x) = [0, \sin(x_1)]^T$, $k(x) = [1, 0]^T$, and $v \in \mathbb{R}$.

668 Since $\|g^+(x)k(x)d(x)\| = 0$, we choose $\ell_M(x) = 0$. Let
 669 $\rho = 0.25$ and $Q = I_2$. Then, based on (5), the cost function
 670 for system (67) can be given as

$$V^{u,v}(x_0) = \int_0^\infty \left(1.25 \|x\|^2 + \mathcal{W}(u) + 0.25 v^\top v \right) dt$$

672 where

$$\begin{aligned} 673 \quad \mathcal{W}(u) &= 2 \int_0^u \beta \tanh^{-1}(\xi/\beta) d\xi \\ 674 \quad &= 2\beta u \tanh^{-1}(u/\beta) + \beta^2 \ln\left(1 - u^2/\beta^2\right). \end{aligned} \quad (68)$$

675 To obtain the optimal ETC, we employ the critic
 676 network (34). The parameters used in the critic network
 677 and the event-triggering condition (46) are given as follows:

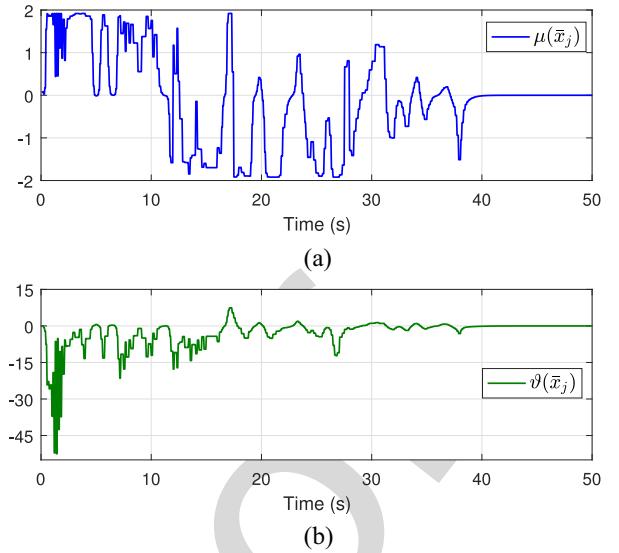


Fig. 4. (a) ETC $\mu(\bar{x}_j)$. (b) Auxiliary ETC $\vartheta(\bar{x}_j)$.

$l_c = 0.5$, $\tilde{n}_c = 8$, $\gamma = 0.3$, and $K_{\max} = 3.5$. The basic function vector used in the critic network is

$$\sigma_c(x) = \left[x_1^2, x_2^2, x_1 x_2, x_1^4, x_2^4, x_1^3 x_2, x_1^2 x_2^2, x_1 x_2^3 \right]^\top. \quad 680$$

The weight vector in the critic network is written as $\hat{w}_c = [\hat{w}_{c1}, \hat{w}_{c2}, \dots, \hat{w}_{c8}]^\top$. Furthermore, to ensure φ to be persistently exciting, we add the exploration noise $n_e(t) = 12e^{-0.05t}[\sin^2(t)\cos(t) + \sin^2(2t)\cos(0.1t) + \sin^2(-1.2t)\cos(0.5t) + \sin^5(t) + \sin^2(1.12t) + \cos(2.4t)\sin^3(2.4t)]$ into the control at the first 38 s.

Remark 5: Choosing appropriate basic function vectors for critic networks is a challenging issue. Because the number of the elements in the basis function vector is closely associated with the number of neurons in the critic network. In this example, the basic function vector is determined via computer simulations. We find that selecting the basic function vector as the above mentioned $\sigma_c(x)$ can lead to desirable results. In Example 2, the basic function vector is also determined by using computer simulations.

The evolution of auxiliary system states is illustrated in Fig. 2. As displayed in Fig. 2, the auxiliary system states $x_1(t)$ and $x_2(t)$ turn out to be asymptotically stable. The convergence of the critic network weight vector $\hat{\omega}_c$ is shown in Fig. 3. It can be seen that, after the first 40 s, the critic network weight vector converges to $\hat{\omega}_c^{\text{final}} = [0.383, 0.773, -0.068, 0.130, 0.533, -0.018, 0.149, 0.062]^T$. Fig. 4(a) and (b) describes the ETC $\mu(\bar{x}_j)$ and the auxiliary ETC $\vartheta(\bar{x}_j)$. The validity of condition (17) is verified by presenting Fig. 5. As indicated in Fig. 5, (17) holds only if $t \geq 40$ s (i.e., $t_s = 40$ s). Fig. 6(a) and (b) shows the two norms (i.e., the norm of the event-triggering condition $\|e_j\|$ and the norm of the event-triggering threshold $\|\bar{e}_T\|$) and the sampling period T_s , respectively. Observing Fig. 6(b), we can find that $\min T_s = 0.1$ s. Actually, there are 289 state samples in Fig. 6(b), which indicates that only 289 state samples are necessary to implement the ETC algorithm. Nevertheless, under the same

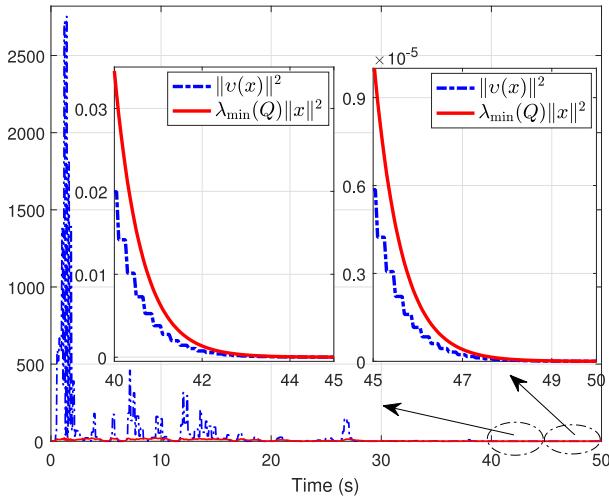
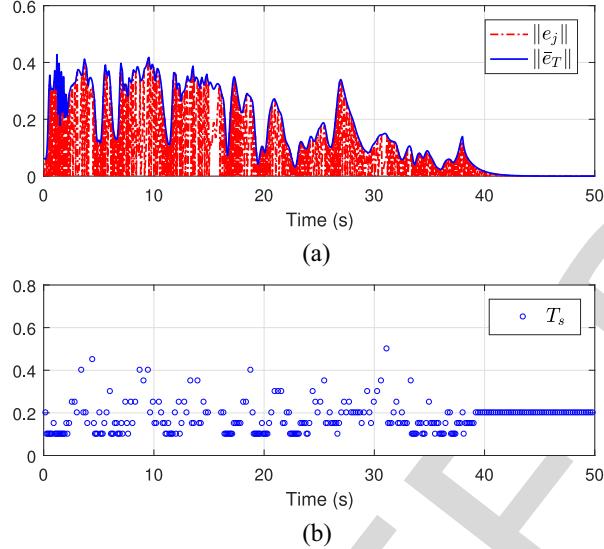


Fig. 5. Verification of condition (17) in Example 1.

Fig. 6. (a) Norm of the event-triggering condition $\|e_j\|$ and norm of the event-triggering threshold $\|\bar{e}_T\|$. (b) Sampling period T_s in Example 1.

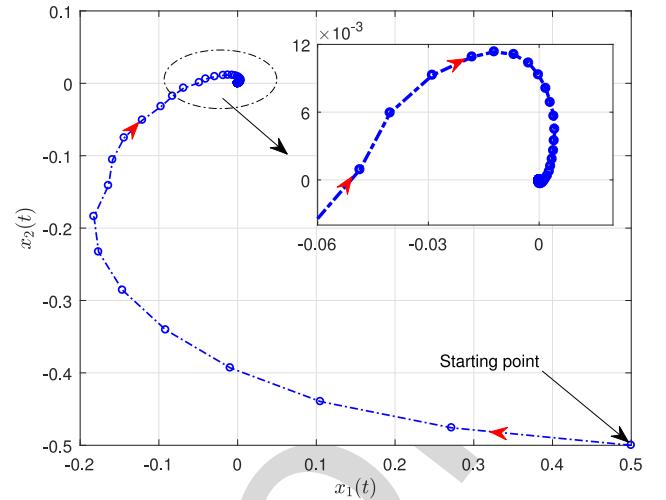
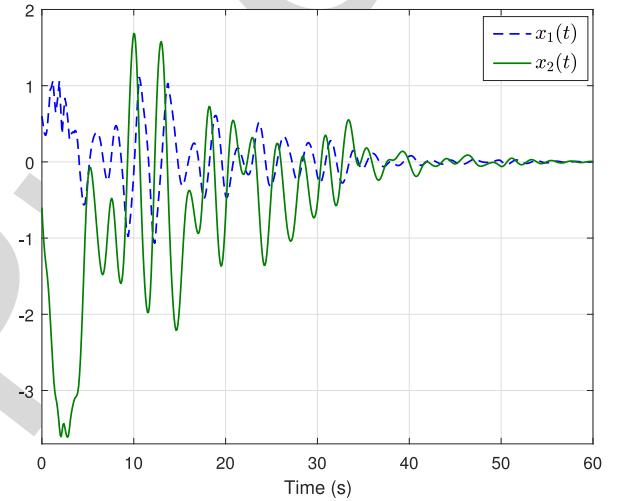
condition, there are 1000 state samples utilized to implement the time-triggering control algorithm. Consequently, the present ETC strategy reduces the controller updates up to 71.1%. In this sense, the computational burden is remarkably decreased. Substituting the above obtained weight vector $\hat{\omega}_c^{\text{final}}$ to (35), we derive the approximate optimal ETC. Fig. 7 displays the states of closed-loop system (66). From Fig. 7, we can see that the states of system (66) are stable under the approximate ETC.

723 B. Example 2: Application to the Pendulum System

724 We consider the pendulum system given in [19] as

$$725 \quad \begin{cases} \frac{d\theta}{dt} = v + d \\ J \frac{dv}{dt} = u - Mgl \sin(\theta) - f_d \frac{d\theta}{dt} \end{cases} \quad (69)$$

726 with the current angle position of the pendulum $\theta \in \mathbb{R}$, the 727 angular velocity $v \in \mathbb{R}$, the perturbation $d \in \mathbb{R}$, the control 728 $u \in \{u \in \mathbb{R} : |u| \leq 1\}$ (note: $\beta = 1$), the mass of the pendulum

Fig. 7. States $x_1(t)$ and $x_2(t)$ of closed-loop system (66).Fig. 8. Evolution of auxiliary system states $x_1(t)$ and $x_2(t)$ in Example 2.

$M = 4/3$ kg, the acceleration of gravity $g = 9.8$ m/s², the length of the pendulum $l = 3/2$ m, the rotary inertia $J = 4/3Ml^2$ kg · m², and the frictional factor $f_d = 0.2$. We assume that $d = \delta_1 \theta \sin(\delta_2 v)$ with δ_1 and δ_2 randomly chosen within the intervals $[-\sqrt{2}/2, \sqrt{2}/2]$ and $[-2, 2]$, respectively.

Let $x_1 = \theta$ and $x_2 = v$. Observing that $g(x) = [0, 0.25]^\top$ and $k(x) = [1, -0.2]^\top$, we obtain $h(x) = (I_2 - g(x)g^+(x))k(x) = [1, 0]^\top$. Then, according to (4), the auxiliary system related to (69) can be proposed as

$$738 \quad \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ -4.9 \sin(x_1) - 0.2x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 0.25 \end{bmatrix}u + \begin{bmatrix} 1 \\ 0 \end{bmatrix}v. \quad (70)$$

The initial state is $x_0 = [0.6, -0.6]^\top$. Since $d(x) = \delta_1 x_1 \sin(\delta_2 x_2)$, we have that $\|d(x)\| \leq \sqrt{2}/2\|x\|$ and $\|g^+(x)k(x)d(x)\| \leq 0.4\sqrt{2}\|x\|$. Therefore, we choose $d_M(x) = (\sqrt{2}/2)\|x\|$ and $\ell_M(x) = 0.4\sqrt{2}\|x\|$ to satisfy Assumption 2. Selecting $\rho = 0.4$, $Q = I_2$, and using (5), we can present the cost function for system (70) as follows:

$$745 \quad V^{u,v}(x_0) = \int_0^\infty (1.84\|x\|^2 + \mathcal{W}(u) + 0.4v^\top v) dt$$

with $\mathcal{W}(u)$ defined as in (68).

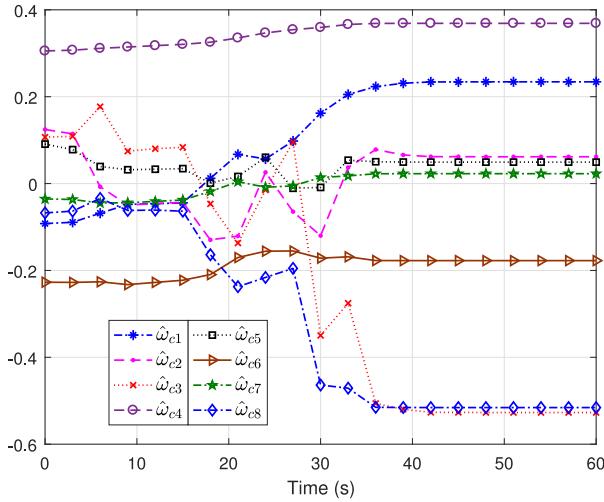
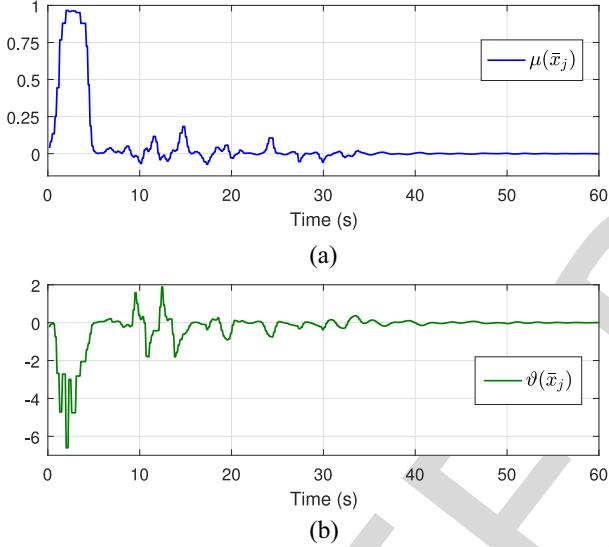
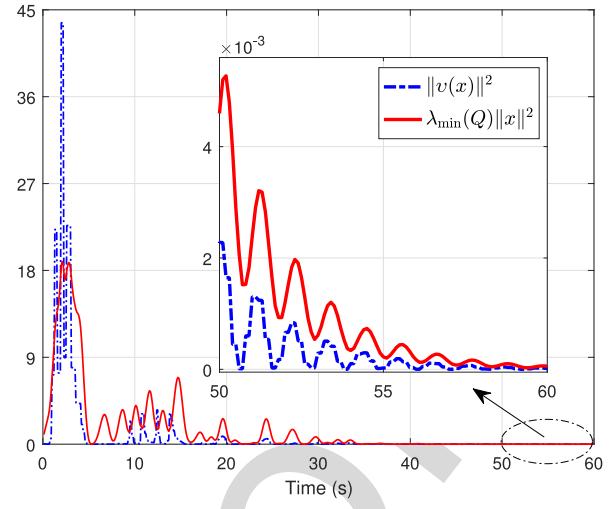
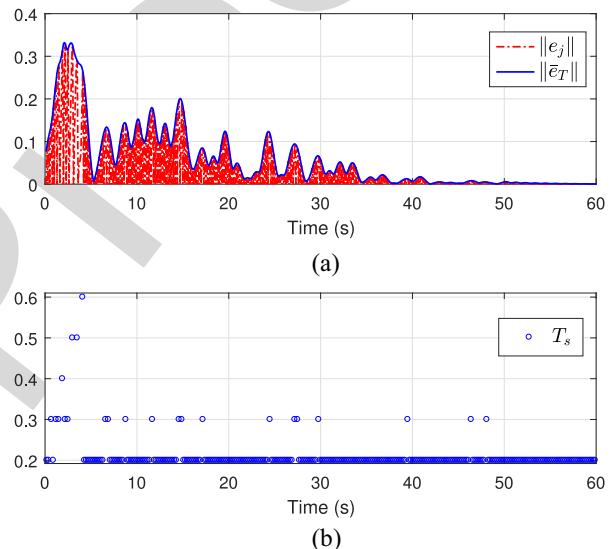
Fig. 9. Convergence of critic network weight vector $\hat{\omega}_c$ in Example 2.Fig. 10. (a) ETC $\mu(\bar{x}_j)$. (b) Auxiliary ETC $\vartheta(\bar{x}_j)$.

Fig. 11. Verification of condition (17) in Example 2.

Fig. 12. (a) Norm of the event-triggering condition $\|\bar{e}_j\|$ and norm of the event-triggering threshold $\|\bar{e}_T\|$. (b) Sampling period T_s in Example 2.

The critic network given as in (34) is applied to derive the optimal ETC law for system (70). The parameters employed in the critic network and the event-triggering condition (46) are the same as in Example 1. The basic function vector used in the critic network is

$$\sigma_c(x) = [x_1^2, x_2^2, x_1 x_2, x_1^4, x_2^4, x_1^3 x_2, x_1^2 x_2^2, x_1 x_2^3]^T$$

and the weight vector in the critic network is denoted as $\hat{\omega}_c = [\hat{\omega}_{c1}, \hat{\omega}_{c2}, \dots, \hat{\omega}_{c8}]^T$. To ensure φ to be persistently exciting, we add the following exploration noise $n_e(t) = 3e^{-0.15t}[\sin^2(t)\cos(t) + \sin^2(2t)\cos(0.1t) + \dots + \sin^2(-1.2t)\cos(0.5t) + \sin^5(t) + \sin^2(1.12t) + \cos(2.4t)\sin^3(2.4t)]$ into the control at the first 50 s.

The evolution of auxiliary system states is displayed in Fig. 8. As illustrated in Fig. 8, the auxiliary system states $x_1(t)$ and $x_2(t)$ are UUB. The convergence of the weight vector $\hat{\omega}_c$ is depicted in Fig. 9. It can be observed that, after the first 50 s, the critic network weight vector converges to $\hat{\omega}_c^{\text{final}} = [0.234, 0.062, -0.527, 0.369, 0.049, -0.177, 0.023, -0.516]^T$. Fig. 10(a) and (b) indicates the ETC $\mu(\bar{x}_j)$ and the auxiliary ETC $\vartheta(\bar{x}_j)$. The validity of condition (17) is verified through Fig. 11. As shown in Fig. 11, (17) holds only if $t \geq 50$ s (i.e., $t_s = 50$ s). Fig. 12(a) and (b) describes the two norms (i.e., the norm of the event-triggering condition $\|\bar{e}_j\|$ and the norm of the event-triggering threshold $\|\bar{e}_T\|$) and the sampling period T_s , respectively. From Fig. 12(b), we can see that $\min T_s = 0.2$ s. In fact, there are 284 state samples in Fig. 12(b), which shows that only 284 state samples are necessary to implement the ETC algorithm. Nonetheless, under the same condition, there are 600 state samples required to implement the time-triggering control algorithm. Accordingly, the present ETC strategy reduces the controller updates up to 52.67%. In this sense, the computational burden is remarkably decreased. Substituting the above derived weight vector $\hat{\omega}_c^{\text{final}}$ to (35), we obtain the approximate optimal ETC. Fig. 13 displays the states of closed-loop system (69). Observing Fig. 13,

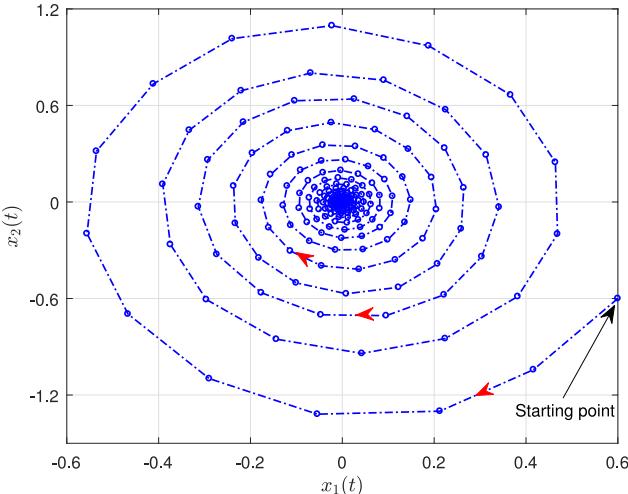


Fig. 13. States $x_1(t)$ and $x_2(t)$ of closed-loop system (69).

782 we can find that the approximate optimal ETC keeps the states
783 of system (69) stable.

VI. CONCLUSION

784 This paper has presented a robust ETC method for nonlinear
785 ear input-constrained systems with mismatched perturbations.
786 By using an SN-ACD, the robust ETC law was obtained via
787 solving the H_2 optimal control problem. Thus, the proposed
788 control method can avoid the difficulty in judging the exist-
789 ence of saddle points, which arises in H_∞ optimal control
790 problems. However, the developed robust ETC approach has
791 to calculate the Moore–Penrose pseudo-inverse of the control
792 matrix function and offer its upper bounded function. Indeed,
793 this is a limitation when applying the present control strategy
794 to nonlinear systems with complicated structures. Hence, how
795 to relax this condition is one subject of our future studies.

796 On the other hand, this robust ETC approach is devel-
797 oped for a unique agent system. In recent years, ACDs
798 have been utilized to study optimal regulations of multiagent
799 systems [54], [55]. It is well-known that multiagent plants
800 exist widely in engineering applications, such as the coordi-
801 nation of unmanned aerial vehicles. In addition, packet loss
802 is an important issue arising in multiagent systems, espe-
803 cially networked multiagent systems. Recently, Lu *et al.* [56]
804 proposed an effective model predictive tracking control strat-
805 egy for networked systems subject to random packet loss
806 and uncertainties. Accordingly, how to extend the SN-ACD
807 to develop robust ETC schemes for multiagent systems with
808 random packet loss and uncertainties is also one direction in
809 our future research.

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