

Low-Latency Multiuser Two-Way Wireless Relaying for Spectral and Energy Efficiencies

Zhichao Sheng , Hoang Duong Tuan , Trung Q. Duong , *Senior Member, IEEE*,
H. Vincent Poor , *Fellow, IEEE*, and Yong Fang , *Senior Member, IEEE*

Abstract—This paper considers two possible approaches, which enable multiple pairs of users to exchange information via multiple multi-antenna relays within one time-slot to save communication bandwidth in low-latency communications. The first approach is to deploy full-duplexes for both users and relays to make their simultaneous signal transmission and reception possible. In the second approach, the users use a fraction of a time slot to send their information to the relays and the relays use the remaining complementary fraction of the time slot to send the beamformed signals to the users. The inherent loop self-interference in the duplexes and inter-full-duplexing-user interference in the first approach are absent in the second approach. Under both these approaches, the joint design of the users' power allocation and relays' beamformers to either optimize the users' exchange of information or maximize the energy-efficiency subject to user quality-of-service (QoS) constraints in terms of minimal rate thresholds leads to complex nonconvex optimization problems. Path-following algorithms are developed for their computational solutions. Numerical examples show the advantages of the second approach over the first approach.

Index Terms—Full-duplex, time-fraction allocation, relay beamforming, power allocation, spectral efficiency, energy efficiency, multi-user communication, path-following methods.

I. INTRODUCTION

FULL-DUPLEXING (FD) [1]–[5] is a technique for simultaneous transmission and reception in the same time slot and over the same frequency band while two-way relaying

(TWR) [6]–[9] allows pairs of users to exchange their information in one step. FD deployed at both users and relays thus enables the users to exchange information via relays within a single time-slot [10]. This is in contrast to the conventional one-way relaying which needs four time slots, and the half-duplexing (HD) TWR [8], [11]–[13], which needs two time slots for the same task. Thus, FD TWR promises to be a very attractive tool for device-to-device (D2D) and machine-to-machine (M2M) communications [14], [15] and low latency communication [16]–[18] for Internet of Things (IoT) applications.

The major issue in FD is the loop self-interference (SI) due to the co-location of transmit antennas and receive antennas. Despite considerable progress [3]–[5], it is still challenging to attenuate the FD SI to a level such that FD can use techniques of signal processing to outperform the conventional half-duplexing in terms of spectral and energy efficiencies [19], [20]. Similarly, it is not easy to manage TWR multi-channel interference, which becomes double as compared to one-way relaying [21], [22]. The FD-based TWR suffers even more severe interference than the FD one-way relaying, which may reduce any throughput gain achieved by using fewer time slots [10].

There is another approach to implement HD TWR within a single time slot, which avoids FD at both users and relays. In a fraction of a time slot, the HD users send the information intended for their partners to the relays and then the relays send the beamformed signals to the users within the remaining fraction of the time slot. In contrast to FD relays, which use half of their available antennas for simultaneous transmission and reception, the HD relays now can use all their antennas for separate transmissions and receptions. Thus, compared with FD users, which need two antennas for simultaneous transmission and reception, the HD users now need only one antenna for separate transmission and reception.

In this paper, we consider the problem of joint design of users' power allocation and relays' beamformers to either maximize the user information exchange throughput or the network energy efficiency [23] subject to user quality-of-service (QoS) constraints in terms of minimal rate thresholds. As they constitute optimization of nonconvex objective functions subject to nonconvex constraints under both these approaches, finding a feasible point is already challenging computationally. Nevertheless, like [13] we develop efficient path-following algorithms for their computation, which not only converge rapidly but also invoke a low-complexity convex quadratic optimization prob-

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Z. Sheng and H. D. Tuan are with the School of Electrical and Data Engineering, University of Technology Sydney, Ultimo, NSW 2007, Australia (e-mail: zhichaosheng@163.com; Tuan.Hoang@uts.edu.au).

T. Q. Duong is with the Queen's University Belfast, Belfast BT7 1NN, U.K. (e-mail: trung.q.duong@qub.ac.uk).

H. V. Poor is with the Department of Electrical Engineering, Princeton University, Princeton, NJ 08544 USA (e-mail: poor@princeton.edu).

Y. Fang is with the Key Laboratory of Specialty Fiber Optics and Optical Access Networks, Shanghai Institute for Advanced Communication and Data Science, Shanghai University, Shanghai 200444, China (e-mail: yfang@staff.shu.edu.cn).

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lem at each iteration for generating a new and better feasible point. The numerical examples demonstrate the full advantage of the second approach over the first approach. Some transformations proposed in [13] to transform nonconvex constraints to convex constraints for computational tractability are also used in this paper. However, compared to [13] this paper offers the following further developments:

- To address the optimization problems in an FD-based TWR setting, we propose a new and tighter bound for the non-concave objective functions, which is based on one step of approximation instead of multiple steps of approximation as in [13]. This helps to expand the search area for locating an optimal solution to accelerate the computational convergence.
- The presence of time fractions as an additional optimization variable in the optimization problems in time-fraction (TF)-wise HD TWR setting makes the transformations proposed in [13] no longer sufficient for transforming all nonconvex constraints to convex constraints. We develop new complementary transformations for transforming the nonconvex-still constraints to convex constraints, preserving the convexity of the existing convex constraints and making the objective functions more computationally tractable. Novel lower bounding approximations for the new objective functions, which are based on newly obtained inequalities, are then derived for developing the corresponding efficient path-following algorithms.

The rest of this paper is organized as follows. Section II considers the two aforementioned nonconvex problems under an FD-based TWR setting. Section III considers them under the TF-wise HD TWR setting. Section IV verifies the full advantage of the TF-wise HD TWR over FD-based TWR via numerical examples. Section V concludes the paper. The appendix provides some fundamental inequalities, which play a crucial role in the development of the path-following algorithms in the previous sections.

Notation: Bold-faced characters denote matrices and column vectors, with upper case used for the former and lower case for the latter. $\mathbf{X}(n, \cdot)$ represents the n th row of the matrix $\mathbf{X}$ while $\mathbf{X}(n, m)$ is its (n, m) th entry. $\langle \mathbf{X} \rangle$ is the trace of the matrix $\mathbf{X}$. $(\cdot)^T$ and $(\cdot)^H$ respectively are the transpose and complex transpose operators. The inner product between vectors $\mathbf{x}$ and $\mathbf{y}$ is defined as $\langle \mathbf{x}, \mathbf{y} \rangle = \mathbf{x}^H \mathbf{y}$. $\|\cdot\|$ denotes either the Euclidean vector squared norm or the Frobenius matrix squared norm. Accordingly, $\|\mathbf{X}\|^2 = \langle \mathbf{X}^H \mathbf{X} \rangle$ for any complex $\mathbf{X}$. Lastly, $\mathbf{x} \sim \mathcal{CN}(\bar{\mathbf{x}}, \mathbf{R}_x)$ means that $\mathbf{x}$ is a vector of Gaussian random variables with mean $\bar{\mathbf{x}}$ and covariance $\mathbf{R}_x$.

II. FULL-DUPLEXING BASED TWO-WAY RELAYING

Fig. 1 illustrates an FD TWR network consisting of K pairs of FD users (UEs) and M FD relays indexed by $m \in \mathcal{M} \triangleq \{1, \dots, M\}$. Each FD user uses one transmit antenna and one receive antenna, while each FD relay uses N_R receive antennas and N_R transmit antennas. Without loss of generality, the k th UE (UE k) and $(k + K)$ th UE (UE $k + K$) are assumed to exchange information with each other via the relays. The pairing operator

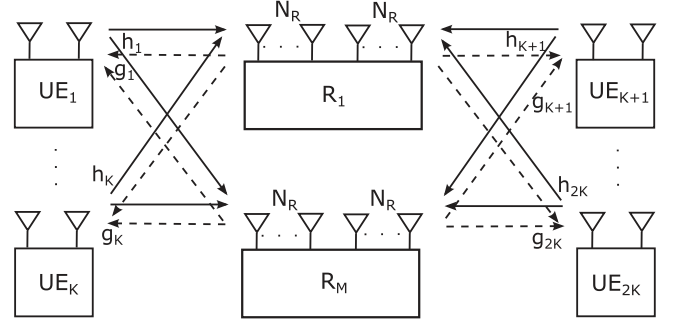


Fig. 1. Two-way relay networks with multiple two-antenna users and multiple multi-antenna relays.

is thus defined as $a(k) = K + k$ for $k \leq K$ and $a(k) = k - K$ if $k > K$. For each $k \in \mathcal{K} \triangleq \{1, \dots, 2K\}$, define the set of UEs that are on the same side of the relays as k th UE as

$$\mathcal{U}(k) = \begin{cases} 1, 2, \dots, K & \text{for } 1 \leq k \leq K \\ K + 1, \dots, 2K & \text{for } k \geq K + 1. \end{cases}$$

Under simultaneous transmission and reception, FD UEs in $\mathcal{U}(k)$ interfere each other. This kind of interference is called inter-FD-user interference.

Let $\mathbf{s} = [s_1, \dots, s_{2K}] \in \mathcal{C}^{2K}$ be the vectors of information symbols s_k transmitted from UEs, which are independent and have unit energy, i.e. $\mathbb{E}[\mathbf{s}\mathbf{s}^H] = \mathbf{I}_{2K}$. For $\mathbf{h}_{\ell,m} \in \mathbb{C}^{N_R}$ as the vector of channels from UE ℓ to relay m , the received signal at relay m is

$$\mathbf{r}_m = \sum_{\ell \in \mathcal{K}} \sqrt{p_\ell} \mathbf{h}_{\ell,m} s_\ell + e_{LI,m} + \mathbf{n}_{R,m}, \quad (1)$$

where $\mathbf{n}_{R,m} \sim \mathcal{CN}(0, \sigma_R^2 \mathbf{I}_{N_R})$ is the background noise, and $\mathbf{p} = (p_1, \dots, p_{2K})$ is a UE power allocation vector, while $e_{LI,m} \in \mathbb{C}^{N_R}$ models the effect of analog circuit non-ideality and the limited dynamic range of the analog-to-digital converter (ADC) at FD relay m .

The transmit power of UEs is physically limited by $P^{U,\max}$ as

$$p_k \leq P^{U,\max}, \quad k \in \mathcal{K}. \quad (2)$$

The total transmit power of UEs is bounded by $P_{\text{sum}}^{U,\max}$ to prevent their excessive interference to other networks as

$$P_{\text{sum}}^U(\mathbf{p}) = \sum_{k \in \mathcal{K}} p_k \leq P_{\text{sum}}^{U,\max}. \quad (3)$$

Relay m processes the received signal by applying the beam-forming matrix $\mathbf{W}_m \in \mathbb{C}^{N_R \times N_R}$ for transmission:

$$\begin{aligned} \mathbf{r}_{m,b} &= \mathbf{W}_m \mathbf{r}_m \\ &= \sum_{\ell \in \mathcal{K}} \sqrt{p_\ell} \mathbf{W}_m \mathbf{h}_{\ell,m} s_\ell + \mathbf{W}_m (e_{LI,m} + \mathbf{n}_{R,m}). \end{aligned} \quad (4)$$

For simplicity it is assumed that $\mathbf{W}_m e_{LI,m} \sim \mathcal{CN}(0, \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m) \mathbf{I}_{N_R})$ with the relay channel's instantaneous residual

SI attenuation level σ_{SI} .¹ This gives

$$\mathbb{E}[\|\mathbf{W}_m e_{LI,m}\|^2] = \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m),$$

in calculating the transmit power at relay m in a closed-form as

$$\begin{aligned} P_m^A(\mathbf{p}, \mathbf{W}_m) &= \mathbb{E}[\|\mathbf{r}_{m,b}\|^2] \\ &= \sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell,m}\|^2 + \sigma_R^2 \|\mathbf{W}_m\|^2 \\ &\quad + \mathbb{E}[\|\mathbf{W}_m e_{LI,m}\|^2] \\ &= \left[\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell,m}\|^2 + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] / (1 - \sigma_{SI}^2). \end{aligned} \quad (5)$$

This transmit power at relay m must be physically limited by a physical parameter $P^{A,\max}$ as

$$P_m^A(\mathbf{p}, \mathbf{W}_m) \leq P^{A,\max}, m \in \mathcal{M}, \quad (6)$$

and their sum is also bounded by $P_{\text{sum}}^{R,\max}$ to control the network emission to other networks:

$$\begin{aligned} P_{\text{sum}}^R(\mathbf{p}, \mathbf{W}) &= \sum_{m \in \mathcal{M}} P_m^A(\mathbf{p}, \mathbf{W}_m) \\ &= \sum_{m \in \mathcal{M}} \left[\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell,m}\|^2 + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] / (1 - \sigma_{SI}^2) \\ &\leq P_{\text{sum}}^{R,\max}. \end{aligned} \quad (7)$$

The relays transmit the processed signals to all UEs. For the vector channel $\mathbf{g}_{m,k} \in \mathcal{C}^{N_R}$ from relay m to UE k and channel $\chi_{\eta,k}$ from UE $\eta \in \mathcal{U}(k)$ to UE k , the received signal at UE k is given by

$$\begin{aligned} y_k &= \sum_{m \in \mathcal{M}} \mathbf{g}_{m,k}^T \mathbf{r}_{m,b} + \sum_{\eta \in \mathcal{U}(k)} \chi_{\eta,k} \sqrt{p_\eta} \tilde{s}_\eta + n_k \\ &= \sum_{m \in \mathcal{M}} \mathbf{g}_{m,k}^T \left[\sum_{\ell \in \mathcal{K}} \sqrt{p_\ell} \mathbf{W}_m \mathbf{h}_{\ell,m} s_\ell + \mathbf{W}_m (e_{LI,m} + \mathbf{n}_{R,m}) \right] \\ &\quad + \sum_{\eta \in \mathcal{U}(k)} \chi_{\eta,k} \sqrt{p_\eta} \tilde{s}_\eta + n_k, \end{aligned} \quad (8)$$

where $n_k \sim \mathcal{CN}(0, \sigma_k^2)$ is the background noise, and $|\chi_{k,k}|^2 = \sigma_{SI}^2$ as $\chi_{k,k} \tilde{s}_k$ represents the loop interference at UE k . We can rewrite (8) as

$$\begin{aligned} y_k &= \sqrt{p_{a(k)}} \sum_{m \in \mathcal{M}} \mathbf{g}_{m,k}^T \mathbf{W}_m \mathbf{h}_{a(k),m} s_{a(k)} \\ &\quad + \sqrt{p_k} \sum_{m \in \mathcal{M}} \mathbf{g}_{m,k}^T \mathbf{W}_m \mathbf{h}_{k,m} s_k \\ &\quad + \sum_{m \in \mathcal{M}} \mathbf{g}_{m,k}^T \left[\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \sqrt{p_\ell} \mathbf{W}_m \mathbf{h}_{\ell,m} s_\ell \right. \end{aligned}$$

¹It is more practical to assume $e_{LI,m} \sim \mathcal{CN}(0, \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m) I_{N_R})$ so $\mathbf{W}_m e_{LI,m} \sim \mathcal{CN}(0, \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m) \mathbf{W}_m \mathbf{W}_m^H)$ resulting in $\mathbb{E}[\|\mathbf{W}_m e_{LI,m}\|^2] = \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m) \|\mathbf{W}_m\|^2$. Usually $\|\mathbf{W}_m\|^2 \leq \nu$ can be assumed so $\mathbb{E}[\|\mathbf{W}_m e_{LI,m}\|^2] = \sigma_{SI}^2 P_m^A(\mathbf{p}, \mathbf{W}_m)$ for $\sigma_{SI}^2 = \nu \sigma_{SI}^2$.

$$\begin{aligned} &+ \mathbf{W}_m (e_{LI,m} + \mathbf{n}_{R,m}) \Big] \\ &+ \sum_{\eta \in \mathcal{U}(k)} \chi_{\eta,k} \sqrt{p_\eta} \tilde{s}_\eta + n_k. \end{aligned} \quad (9)$$

Note that the first term in (9) is the desired signal component, the third term is the inter-pair interference and the last two terms are noise. UE k can cancel the self-interference represented by the second term using the channel state information of the forward channels $\mathbf{h}_{k,m}$ from itself to the relays and backward channels $\mathbf{g}_{m,k}$ from the relays to itself as well as the beamforming matrix $\mathbf{W}_m$. The challenge here is that the loop SI term $\sum_{\eta \in \mathcal{U}(k)} \chi_{\eta,k} \sqrt{p_\eta} \tilde{s}_\eta$, which may be strong due to the proximity of UEs in $\mathcal{U}(k)$, cannot be nulled out. This means more power should be given to the relays, but that leads to more FD SI at the relays.

Furthermore, for $\mathbf{f}_{m,k}^H \triangleq \mathbf{g}_{m,k}^T$, the signal-to-interference-plus-noise ratio (SINR) at UE k 's receiver can be calculated as

$$\begin{aligned} \gamma_k(\mathbf{p}, \mathbf{W}) &= p_{a(k)} \left| \sum_{m \in \mathcal{M}} \mathbf{f}_{m,k}^H \mathbf{W}_m \mathbf{h}_{a(k),m} \right|^2 \\ &\quad / \left[\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} p_\ell \left| \sum_{m \in \mathcal{M}} \mathbf{f}_{m,k}^H \mathbf{W}_m \mathbf{h}_{\ell,m} \right|^2 \right. \\ &\quad \left. + \sigma_R^2 \sum_{m \in \mathcal{M}} \|\mathbf{f}_{m,k}^H \mathbf{W}_m\|^2 \right. \\ &\quad \left. + \frac{\sigma_{SI}^2}{1 - \sigma_{SI}^2} \sum_{m \in \mathcal{M}} \|\mathbf{g}_{m,k}\|^2 \left(\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell,m}\|^2 \right. \right. \\ &\quad \left. \left. + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) + \sum_{\eta \in \mathcal{U}(k)} |\chi_{\eta,k}|^2 p_\eta + \sigma_k^2 \right]. \end{aligned} \quad (10)$$

Under the definitions

$$\mathcal{L}_{k,\ell}(\mathbf{W}) \triangleq \sum_{m \in \mathcal{M}} \mathbf{f}_{m,k}^H \mathbf{W}_m \mathbf{h}_{\ell,m},$$

$$\mathcal{L}_k(\mathbf{W}) \triangleq [\mathbf{f}_{1,k}^H \mathbf{W}_1 \mathbf{f}_{2,k}^H \mathbf{W}_2 \dots \mathbf{f}_{M,k}^H \mathbf{W}_M], \quad (11)$$

it follows that

$$\begin{aligned} \gamma_k(\mathbf{p}, \mathbf{W}) &= p_{a(k)} |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 \\ &\quad / \left[\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} p_\ell |\mathcal{L}_{k,\ell}(\mathbf{W})|^2 + \sigma_R^2 \|\mathcal{L}_k(\mathbf{W})\|^2 \right. \\ &\quad \left. + \frac{\sigma_{SI}^2}{1 - \sigma_{SI}^2} \sum_{m \in \mathcal{M}} \|\mathbf{g}_{m,k}\|^2 \left(\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell,m}\|^2 \right. \right. \\ &\quad \left. \left. + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) + \sum_{\eta \in \mathcal{U}(k)} |\chi_{\eta,k}|^2 p_\eta + \sigma_k^2 \right]. \end{aligned} \quad (12)$$

In FD TWR, the performance of interest is the information exchange throughput of UE pairs:

$$R_k(\mathbf{p}, \mathbf{W}) = \ln(1 + \gamma_k(\mathbf{p}, \mathbf{W})) + \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W})),$$

$$k = 1, \dots, K. \quad (13)$$

The problem of maximin information exchange throughput optimization subject to transmit power constraints is then formulated as

$$\max_{\mathbf{W}, \mathbf{p}} \min_{k=1, \dots, K} [\ln(1 + \gamma_k(\mathbf{p}, \mathbf{W})) + \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W}))] \quad (14a)$$

$$\text{s.t.} \quad (2), (3), (6), (7). \quad (14b)$$

Another problem, which has attracted recent attention in 5G [23], [24] is the following problem of maximizing the network energy-efficiency (EE) subject to UE QoS in terms of the information exchange throughput thresholds:

$$\max_{\mathbf{W}, \mathbf{p}} \sum_{k=1}^K [\ln(1 + \gamma_k(\mathbf{p}, \mathbf{W})) + \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W}))]$$

$$/ [\zeta(P_{\text{sum}}^U(\mathbf{p}) + P_{\text{sum}}^R(\mathbf{p}, \mathbf{W})) + MP^R + 2KP^U] \quad (15a)$$

$$\text{s.t.} \quad (2), (3), (6), (7), \quad (15b)$$

$$R_k(\mathbf{p}, \mathbf{W}) \geq r_k, k = 1, \dots, K, \quad (15c)$$

where ζ , P^R and P^U are the reciprocal of the drain efficiency of power amplifier, and the circuit powers of the relay and UE, respectively, and r_k sets the exchange throughput threshold for UE pairs.

The next two subsections are devoted to computational solutions for problems (14) and (15), respectively.

A. FD TWR Maximin Information Exchange Throughput Optimization

By introducing new nonnegative variables

$$\beta_k = 1/p_k^2 > 0, k \in \mathcal{K}, \quad (16)$$

and functions

$$\Psi_{k,\ell}(\mathbf{W}, \alpha, \beta) \triangleq |\mathcal{L}_{k,\ell}(\mathbf{W})|^2 / \sqrt{\alpha\beta}, (k, \ell) \in \mathcal{K} \times \mathcal{K},$$

$$\Upsilon_k(\mathbf{W}, \alpha) \triangleq \|\mathcal{L}_k(\mathbf{W})\|^2 / \sqrt{\alpha}, k \in \mathcal{K},$$

$$\Phi_{\ell,m}(\mathbf{W}_m, \alpha, \beta) \triangleq \|\mathbf{h}_{\ell,m}^H \mathbf{W}_m\|^2 / \sqrt{\alpha\beta}, (\ell, m) \in \mathcal{K} \times \mathcal{M}, \quad (17)$$

which are convex [25], (12) can be re-expressed as

$$\gamma_k(\mathbf{p}, \mathbf{W}) = |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\beta_{a(k)}} \times \left[\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \Psi_{k,\ell}(\mathbf{W}, 1, \beta_\ell) + \sigma_R^2 \Upsilon_k(\mathbf{W}, 1) + \frac{\sigma_{SI}^2}{1 - \sigma_{SI}^2} \sum_{m \in \mathcal{M}} \|\mathbf{g}_{m,k}\|^2 \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \langle \mathbf{W}_m^H \mathbf{W}_m \rangle \right) + \sum_{\eta \in \mathcal{U}(k)} |\chi_{\eta,k}|^2 / \sqrt{\beta_\eta} + \sigma_k^2 \right]. \quad (18)$$

Similarly to [26] and [13, Th. 1] we can prove the following result.

Theorem 1: The optimization problem (14), which is maximization of nonconcave objective function over a nonconvex set, can be equivalently rewritten as the following problem of maximizing a nonconcave objective function over a set of convex constraints:

$$\max_{\mathbf{W}, \alpha, \beta} f(\mathbf{W}, \alpha, \beta) \triangleq \min_{k=1, \dots, K} \left[\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}}) + \ln(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k}) \right] \quad (19a)$$

$$\text{s.t.} \quad \sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \Psi_{k,\ell}(\mathbf{W}, \alpha_k, \beta_\ell) + \sigma_R^2 \Upsilon_k(\mathbf{W}, \alpha_k) + \sum_{\eta \in \mathcal{U}(k)} |\chi_{\eta,k}|^2 / \sqrt{\alpha_k \beta_\eta} + \frac{\sigma_{SI}^2}{1 - \sigma_{SI}^2} \sum_{m \in \mathcal{M}} \|\mathbf{g}_{m,k}\|^2 \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, \alpha_k, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 / \sqrt{\alpha_k} \right) + \sigma_k^2 / \sqrt{\alpha_k} \leq 1, \quad (19b)$$

$$\beta_k \geq 1/(P^{U,\max})^2, k \in \mathcal{K}, \quad (19c)$$

$$P_{\text{sum}}^U(\beta) := \sum_{k \in \mathcal{K}} 1/\sqrt{\beta_k} \leq P_{\text{sum}}^{U,\max}, \quad (19d)$$

$$\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \leq (1 - \sigma_{SI}^2) P_m^{A,\max}, m \in \mathcal{M}, \quad (19e)$$

$$\sum_{m \in \mathcal{M}} \left[\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] \leq (1 - \sigma_{SI}^2) P_{\text{sum}}^{R,\max}. \quad (19f)$$

As in [13] the main issue is how to handle the nonconcave objective function in (19a). Indeed, one can use [13, Th. 2] for lower bounding the objective function in (19a) by a concave function, which is a reciprocal of a positive linear function over a complex trust region involving all concerned variables. By the

following theorem we provide a new and better lower bound under a simpler trust region involving only the beamforming matrix $\mathbf{W}$, which results in expanded local search areas, accelerating convergence of the designed algorithm. This is a one-step approximation that is in contrast to the multi-step approximation in [13].

Theorem 2: At any $(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)})$ feasible for the convex constraints (19b)–(19f), it is true that

$$\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}}) \geq f_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}) \quad (20)$$

over the trust region

$$2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 > 0, \quad (21)$$

for

$$f_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}) = \ln(1 + x_{k,a(k)}^{(\kappa)}) + a_{k,a(k)}^{(\kappa)} \left[2 - \frac{|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2}{2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2} - \sqrt{\alpha_k \beta_{a(k)}} / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}} \right] \quad (22)$$

with $x_{k,a(k)}^{(\kappa)} \triangleq |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}$ and $a_{k,a(k)}^{(\kappa)} \triangleq x_{k,a(k)}^{(\kappa)} / (x_{k,a(k)}^{(\kappa)} + 1) > 0$.

Proof: (22) follows by applying inequality (59) in the appendix for

$$x = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2, y = \sqrt{\alpha_k \beta_{a(k)}}$$

and

$$\bar{x} = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2, \bar{y} = \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}$$

and then the inequality

$$1/|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 \leq 1 / \left(2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 \right) \quad (23)$$

over the trust region (21). ■

By Algorithm 1 we propose a path-following procedure for computing (19), which solves the following convex optimization problem of inner approximation at the κ th iteration to generate the next feasible point $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$:

$$\begin{aligned} \max_{\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}} \quad & \min_{k=1, \dots, K} [f_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}) \\ & + f_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \beta_k)] \\ \text{s.t.} \quad & (19b) - (19f), (21). \end{aligned} \quad (24)$$

Algorithm 1: Path-Following Algorithm for FD TWR Exchange Throughput Optimization.

initialization: Set $\kappa = 0$. Initialize a feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)})$ for the convex constraints (19b)–(19f) and $R_1 = f(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)})$.

repeat

- $R_0 = R_1$.
- Solve the convex optimization problem (24) to obtain the solution $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$.
- Update $R_1 = f(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$.
- Reset $\kappa \rightarrow \kappa + 1$.

until $\frac{R_1 - R_0}{R_0} \leq \epsilon$ for given tolerance $\epsilon > 0$.

Similarly to [13, Alg. 1], it can be shown that the sequence $\{(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)})\}$ generated by Algorithm 1 converges at least to a locally optimal solution of (19).²

B. FD TWR Energy-Efficiency Maximization

We return to consider the optimization problem (15), which can be shown similarly to Theorem 1 to be equivalent to the following optimization problem under the variable change (16):

$$\max_{\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}} F(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \quad \text{s.t.} \quad (19b) - (19f), \quad (25a)$$

$$\tilde{R}_k(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \geq r_k, k = 1, \dots, K, \quad (25b)$$

for

$$F(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \triangleq \left[\sum_{k=1}^K \tilde{R}_k(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \right] / \pi(\boldsymbol{\beta}, \mathbf{W}),$$

$$\tilde{R}_k(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \triangleq \ln \left(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}} \right) + \ln \left(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k} \right),$$

and

$$\begin{aligned} \pi(\boldsymbol{\beta}, \mathbf{W}) \triangleq & \sum_{k \in \mathcal{K}} \zeta / \sqrt{\beta_k} + (\zeta / (1 - \sigma_{SI}^2)) \\ & \times \sum_{m \in \mathcal{M}} \left[\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) \right. \\ & \left. + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] + MP^R + 2KP^U. \end{aligned} \quad (26)$$

The objective function in (25a) is nonconcave and the constraint (25b) is nonconvex.

Suppose that $(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)})$ is a feasible point for (25) found from the $(\kappa - 1)$ th iteration. Applying inequality (58) in the appendix for

$$x = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2, y = \sqrt{\alpha_k \beta_{a(k)}}, t = \pi(\boldsymbol{\beta}, \mathbf{W})$$

and

$$\bar{x} = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2, \bar{y} = \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}, \bar{t} = \pi(\boldsymbol{\beta}^{(\kappa)}, \mathbf{W}^{(\kappa)})$$

²As mentioned in [27, Remark] this desired property of a limit point indeed does not require the differentiability of the objective function.

and using inequality (23) yield the following new and tighter bound compared to [13, (36)] for the terms of the objective function in (25a), which involves only one approximation step:

$$\left[\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}}) \right] / \pi(\boldsymbol{\beta}, \mathbf{W}) \geq F_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}) \quad (27)$$

over the trust region (21), where

$$F_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}) \triangleq \frac{p_{k,a(k)}^{(\kappa)} + q_{k,a(k)}^{(\kappa)} \left[2 - \frac{|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2}{2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2} - \sqrt{\alpha_k \beta_{a(k)}} / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}} \right] - r_{k,a(k)}^{(\kappa)} \pi(\boldsymbol{\beta}, \mathbf{W}),$$

and

$$\begin{aligned} x_{k,a(k)}^{(\kappa)} &= |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}, \\ t^{(\kappa)} &= \pi(\boldsymbol{\beta}^{(\kappa)}, \mathbf{W}^{(\kappa)}), \\ p_{k,a(k)}^{(\kappa)} &= 2 \left[\ln(1 + x_{k,a(k)}^{(\kappa)}) \right] / t^{(\kappa)} > 0, \\ q_{k,a(k)}^{(\kappa)} &= x_{k,a(k)}^{(\kappa)} / ((x_{k,a(k)}^{(\kappa)} + 1)t^{(\kappa)}) > 0, \\ r_{k,a(k)}^{(\kappa)} &= \left[\ln(1 + x_{k,a(k)}^{(\kappa)}) \right] / (t^{(\kappa)})^2 > 0. \end{aligned} \quad (28)$$

Furthermore, we use $f_{k,a(k)}^{(\kappa)}$ defined from (20) to provide the following inner convex approximation for the nonconvex constraint (25b):

$$f_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}) + f_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \beta_k) \geq r_k. \quad (29)$$

By Algorithm 2 we propose a path-following procedure for computing (25), which solves the following convex optimization problem at the κ th iteration to generate the next feasible point $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$:

$$\begin{aligned} \max_{\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}} \quad & F^{(\kappa)}(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}) \triangleq \sum_{k=1}^K [F_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}) \\ & + F_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \boldsymbol{\beta})] \end{aligned} \quad (30a)$$

$$\text{s.t.} \quad (19b) - (19f), (21), (29). \quad (30b)$$

Analogously to Algorithm 1, the sequence $\{(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)})\}$ generated by Algorithm 2 converges at least to a locally optimal solution of (25).

An initial feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)})$ for initializing Algorithm 2 can be found by using Algorithm 1 for computing (14), which terminates upon

$$\min_{k=1, \dots, K} R_k(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)}) / r_k \geq 1 \quad (31)$$

to satisfy (25b).

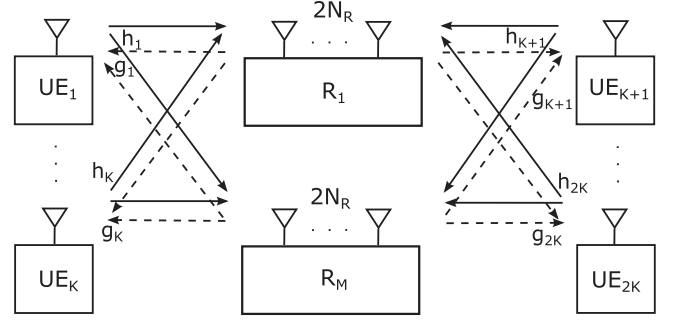


Fig. 2. Two-way relay networks with multiple single-antenna users and multiple multi-antenna relays.

Algorithm 2: Path-Following Algorithm for FD TWR Energy-Efficiency.

initialization: Set $\kappa = 0$. Initialize a feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)})$ for (25) and $e_1 = F(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)})$.

repeat

- $e_0 = e_1$.
- Solve the convex optimization problem (30) to obtain the solution $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$.
- Update $e_1 = F(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)})$.
- Reset $\kappa \rightarrow \kappa + 1$.

until $\frac{e_1 - e_0}{e_0} \leq \epsilon$ for given tolerance $\epsilon > 0$.

III. TIME-FRACTION-WISE HD TWO-WAY RELAYING

Through the FD-based TWR detailed in the previous section one can see the following obvious issues for its practical implementation:

- It is difficult to attenuate FD SI at the UEs and relays to a level that realizes the benefits of FD. The FD SI is even more severe at the relays, which are equipped with multiple antennas;
- Inter-FD-user interference cannot be controlled;
- It is technically difficult to implement FD at UEs, which particularly requires two antennas per UE.

We now propose a new technique for UE information exchange via HD TWR within the time slot as illustrated by Fig. 2, where during time-fraction $0 < \tau < 1$ all UEs send information to the relays and during the remaining time fraction $(1 - \tau)$ the relays send the beamformed signals to the UEs. This alternative has the following advantages:

- Each relay uses all available $2N_R$ antennas for separated receiving and transmitting signals;
- Each UE needs only a single antenna to implement the conventional HD, which transmits signal and receive signals in separated time fractions.

Suppose that UE k uses the power τp_k to send information to the relay. The following physical limitation is imposed:

$$p_k \leq \bar{P}_{\text{UE}}, k \in \mathcal{K}, \quad (32)$$

where $\bar{P}_{\text{UE}}$ is a physical parameter to signify the hardware limit in transmission during time-fractions. Typically, $\bar{P}_{\text{UE}} = 3P^{U, \max}$ for $P^{U, \max}$ defined from (2).

As in (3), the power budget of all UEs is $P_{\text{sum}}^{U, \max}$:

$$P_{\text{sum}}^U(\mathbf{p}) = \tau \sum_{k \in \mathcal{K}} p_k \leq P_{\text{sum}}^{U, \max}. \quad (33)$$

The received signal at relay m can be simply written as

$$\mathbf{r}_m = \sum_{\ell \in \mathcal{K}} \sqrt{\tau p_\ell} \mathbf{h}_{\ell, m} s_\ell + \mathbf{n}_{R, m}^{(\tau)}, \quad (34)$$

where $\mathbf{n}_{R, m}^{(\tau)} \in \mathcal{CN}(0, \tau \sigma_R^2 \mathbf{I}_{2N_R})$ and $\mathbf{h}_{\ell, m} \in \mathbb{C}^{2N_R}$ is the vector of channels from UE ℓ to relay m .

Relay m processes the received signal by applying the beamforming matrix $\mathbf{W}_m \in \mathbb{C}^{2N_R \times 2N_R}$ for transmission:

$$\mathbf{r}_{m, b} = \mathbf{W}_m \mathbf{r}_m = \sum_{\ell \in \mathcal{K}} \sqrt{\tau p_\ell} \mathbf{W}_m \mathbf{h}_{\ell, m} s_\ell + \mathbf{W}_m \mathbf{n}_{R, m}^{(\tau)}. \quad (35)$$

Given the physical parameter $P^{A, \max}$ as in (6) and then $\bar{P}_R = 3P^{A, \max}$, the transmit power at relay m is physically limited as

$$P_m^A(\mathbf{p}, \mathbf{W}_m, \tau) = \tau \left[\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell, m}\|^2 + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] \leq \bar{P}_R, \quad m \in \mathcal{M}. \quad (36)$$

Given a budget $P_{\text{sum}}^{R, \max}$ as in (7), the sum power transmitted by the relays is also constrained as

$$\begin{aligned} P_{\text{sum}}^R(\tau, \mathbf{p}, \mathbf{W}) &= (1 - \tau) \sum_{m \in \mathcal{M}} P_m^A(\mathbf{p}, \mathbf{W}_m, \tau) \\ &= (1 - \tau) \tau \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} p_\ell \|\mathbf{W}_m \mathbf{h}_{\ell, m}\|^2 + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) \\ &\leq P_{\text{sum}}^{R, \max}. \end{aligned} \quad (37)$$

The received signal at UE k can be expressed as

$$\begin{aligned} y_k &= \sqrt{\tau p_{a(k)}} \sum_{m \in \mathcal{M}} \mathbf{g}_{m, k}^T \mathbf{W}_m \mathbf{h}_{a(k), m} s_{a(k)} \\ &+ \sqrt{\tau p_k} \sum_{m \in \mathcal{M}} \mathbf{g}_{m, k}^T \mathbf{W}_m \mathbf{h}_{k, m} s_k \\ &+ \sum_{m \in \mathcal{M}} \mathbf{g}_{m, k}^T \left(\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \sqrt{\tau p_\ell} \mathbf{W}_m \mathbf{h}_{\ell, m} s_\ell + \mathbf{W}_m \mathbf{n}_{R, m}^{(\tau)} \right) + n_k. \end{aligned} \quad (38)$$

Under the definitions

$$\begin{aligned} \mathcal{L}_{k, \ell}(\mathbf{W}) &= \sum_{m \in \mathcal{M}} \mathbf{f}_{m, k}^H \mathbf{W}_m \mathbf{h}_{\ell, m}, \\ \mathcal{L}_k(\mathbf{W}) &= [\mathbf{f}_{1, k}^H \mathbf{W}_1 \mathbf{f}_{2, k}^H \mathbf{W}_2 \dots \mathbf{f}_{M, k}^H \mathbf{W}_M], \end{aligned} \quad (39)$$

the SINR at UE k can be calculated as

$$\begin{aligned} \gamma_k(\mathbf{p}, \mathbf{W}, \tau) &= \frac{p_{a(k)} |\mathcal{L}_{k, a(k)}(\mathbf{W})|^2}{\left[\sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} p_\ell |\mathcal{L}_{k, \ell}(\mathbf{W})|^2 + \sigma_R^2 \|\mathcal{L}_k(\mathbf{W})\|^2 + \sigma_k^2 / \tau \right]}. \end{aligned} \quad (40)$$

Thus, the throughput at the k th UE pair is defined by the following function of the beamforming matrix $\mathbf{W} = \{\mathbf{W}_m\}_{m \in \mathcal{M}}$, power allocation vector $\mathbf{p}$ and time-fraction τ :

$$\begin{aligned} R_k(\tau, \mathbf{p}, \mathbf{W}) &= (1 - \tau) \ln(1 + \gamma_k(\mathbf{p}, \mathbf{W}, \tau)) \\ &+ (1 - \tau) \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W}, \tau)), \\ k &= 1, \dots, K. \end{aligned} \quad (41)$$

Similarly to (14), the problem of maximin information exchange throughput optimization subject to transmit power constraints is formulated as

$$\begin{aligned} \max_{0 < \tau < 1, \mathbf{W}, \mathbf{p}} \quad & \min_{k=1, \dots, K} (1 - \tau) [\ln(1 + \gamma_k(\mathbf{p}, \mathbf{W}, \tau)) \\ &+ \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W}, \tau))] \end{aligned} \quad (42a)$$

$$\text{s.t.} \quad (32), (33), (36), (37), \quad (42b)$$

while the problem of maximizing the network EE subject to UE QoS in terms of the information exchange throughput thresholds is formulated similarly to (15) as

$$\begin{aligned} \max_{0 < \tau < 1, \mathbf{W}, \mathbf{p}} \quad & \sum_{k=1}^K (1 - \tau) [\ln(1 + \gamma_k(\mathbf{p}, \mathbf{W}, \tau)) \\ &+ \ln(1 + \gamma_{a(k)}(\mathbf{p}, \mathbf{W}, \tau))] / [\zeta(P_{\text{sum}}^U(\tau, \mathbf{p}) \\ &+ P_{\text{sum}}^R(\tau, \mathbf{p}, \mathbf{W})) + MP^R + 2KP^U] \end{aligned} \quad (43a)$$

$$\text{s.t.} \quad (32), (33), (36), (37) \quad (43b)$$

$$R_k(\tau, \mathbf{p}, \mathbf{W}) \geq r_k, \quad k = 1, \dots, K. \quad (43c)$$

The next two subsections are devoted to their computation.

A. TF-Wise HD TWR Maximin Information Exchange Throughput Optimization

Similarly to (19), problem (42) of maximin information exchange throughput optimization is equivalently expressed by the following optimization problem with using new variables

$\boldsymbol{\beta} = (\beta_1, \dots, \beta_{2K})^T$ defined from (16):

$$\max_{\substack{0 < \tau < 1, \mathbf{W}, \\ \boldsymbol{\alpha}, \boldsymbol{\beta}}} \min_{k=1, \dots, K} (1 - \tau) \times \left[\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}}) + \ln(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k}) \right] \quad (44a)$$

$$\text{s.t.} \quad \sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \Psi_{k,\ell}(\mathbf{W}, \alpha_k, \beta_\ell) + \sigma_R^2 \Upsilon_k(\mathbf{W}, \alpha_k) + \sigma_k^2 / \tau \sqrt{\alpha_k} \leq 1, k \in \mathcal{K}, \quad (44b)$$

$$\beta_k \geq 1/(\bar{P}_{\text{UE}})^2, k \in \mathcal{K}, \quad (44c)$$

$$\sum_{k \in \mathcal{K}} \tau / \sqrt{\beta_k} \leq P_{\text{sum}}^{U, \max}, \quad (44d)$$

$$\tau \left[\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \right] \leq \bar{P}_R, m \in \mathcal{M}, \quad (44e)$$

$$(1 - \tau) \tau \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) \leq P_{\text{sum}}^{R, \max}. \quad (44f)$$

In contrast to the power constraints (19e) and (19f), which are convex, the last constraints (44e) and (44f) are no longer convex due to the presence of the new time fraction variable τ , which also makes the objective function in (44a) much more complex compared to that in (19a). To address (44) properly we now provide a new variable transformation to transform (19e) and (19f) to convex constraints, preserving the convexity of constraints (44b)–(44d) and even making the objective function in (44a) more computationally tractable, for which we will provide a new bounding technique. To this end, recalling the definition (17), rewrite (44d)–(44f) by

$$\begin{aligned} \sum_{k \in \mathcal{K}} 1/\sqrt{\beta_k} &\leq P_{\text{sum}}^{U, \max} / \tau, \\ \sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 &\leq \bar{P}_R / \tau, m \in \mathcal{M}, \\ \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) &\leq P_{\text{sum}}^{R, \max} / (1 - \tau) \tau. \end{aligned}$$

Introduce the new variables $t_1 > 0$ and $t_2 > 0$ to express $1/\tau^2$ and $1/(1 - \tau)$, which satisfy the convex constraint

$$1/\sqrt{t_1} + 1/t_2 \leq 1. \quad (45)$$

Then, (44) is equivalent to

$$\begin{aligned} \max_{\substack{\mathbf{W} \in \mathbb{C}^{N \times N}, t_1, t_2 \\ \boldsymbol{\alpha} \in \mathbb{R}^{2K}, \boldsymbol{\beta} \in \mathbb{R}^{2K}}} \varphi(\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}, t_2) &\triangleq \\ \min_{k=1, \dots, K} (1/t_2) &\left[\ln \left(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}} \right) \right. \\ &\left. + \ln \left(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k} \right) \right] \quad (46a) \end{aligned}$$

$$\text{s.t.} \quad \sum_{\ell \in \mathcal{K} \setminus \{k, a(k)\}} \Psi_{k,\ell}(\mathbf{W}, \alpha_k, \beta_\ell) + \sigma_R^2 \Upsilon_k(\mathbf{W}, \alpha_k) + \sigma_k^2 / \tau \sqrt{\alpha_k} \leq 1, \quad (46b)$$

$$\beta_k \geq 1/(\bar{P}_{\text{UE}})^2, k \in \mathcal{K}, \quad (46c)$$

$$\sum_{k \in \mathcal{K}} 1/\sqrt{\beta_k} \leq P_{\text{sum}}^{U, \max} \sqrt{t_1}, \quad (46d)$$

$$\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \leq \bar{P}_R \sqrt{t_1}, \quad (46e)$$

$$m \in \mathcal{M},$$

$$\begin{aligned} \frac{1}{\sqrt{t_1}} \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) + \sigma_R^2 \|\mathbf{W}_m\|^2 \right) &\leq t_2 P_{\text{sum}}^{R, \max}, \quad (46f) \end{aligned}$$

where all constraints (46b)–(46f) are convex. Therefore, the next step is to approximate the objective function in (46a).

Suppose $(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)}, t_1^{(\kappa)}, t_2^{(\kappa)})$ is a feasible point for (46) found at the $(\kappa - 1)$ th iteration. Applying (58) in the appendix for

$$x = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2, y = \sqrt{\alpha_k \beta_{a(k)}}, t = t_2$$

and

$$\bar{x} = 1/|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2, \bar{y} = \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}, \bar{t} = t_2^{(\kappa)}$$

and using inequality (23) yields

$$(1/t_2) \ln \left(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}} \right) \geq \Gamma_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}, t_2) \quad (47)$$

over the trust region (21), for

$$\begin{aligned} x_{k,a(k)}^{(\kappa)} &= |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}} \\ c_{k,a(k)}^{(\kappa)} &= (2/t_2^{(\kappa)}) \ln \left(1 + x_{k,a(k)}^{(\kappa)} \right) > 0, \\ d_{k,a(k)}^{(\kappa)} &= x_{k,a(k)}^{(\kappa)} / (x_{k,a(k)}^{(\kappa)} + 1) t_2^{(\kappa)} > 0, \\ e_{k,a(k)}^{(\kappa)} &= (1/t_2^{(\kappa)})^2 \ln \left(1 + x_{k,a(k)}^{(\kappa)} \right) > 0, \end{aligned}$$

Algorithm 3: Path-Following Algorithm for TF-Wise HD TWR Exchange Throughput Optimization.

initialization: Set $\kappa = 0$. Initialize a feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)}, t_1^{(0)}, t_2^{(0)})$ for the convex constraints (46b)–(46f) and $R_1 = \varphi(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)}, t_2^{(0)})$.

repeat

- $R_0 = R_1$.
- Solve the convex optimization problem (49) to obtain the solution $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_1^{(\kappa+1)}, t_2^{(\kappa+1)})$.
- Update $R_1 = \varphi(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_2^{(\kappa+1)})$.
- Reset $\kappa \rightarrow \kappa + 1$.

until $\frac{R_1 - R_0}{R_0} \leq \epsilon$ for given tolerance $\epsilon > 0$.

and

$$\Gamma_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}, t_2) \triangleq \frac{c_{k,a(k)}^{(\kappa)} + d_{k,a(k)}^{(\kappa)}}{2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2} \left[2 - \frac{|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2}{\sqrt{\alpha_k \beta_{a(k)}} / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}} \right] - e_{k,a(k)}^{(\kappa)} t_2. \quad (48)$$

By Algorithm 3 we propose a path-following procedure for computing (46), which solves the following convex optimization problem at the κ th iteration to generate the next feasible point $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_1^{(\kappa+1)}, t_2^{(\kappa+1)})$:

$$\begin{aligned} \max_{\mathbf{W}, \boldsymbol{\alpha}, \boldsymbol{\beta}, t_1, t_2} \quad & \min_{k=1, \dots, K} \left[G_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}, t_1, t_2) \right. \\ & \left. + G_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \beta_k, t_1, t_2) \right] \\ \text{s.t.} \quad & (45), (46b), (46c), (46d), (46e), (46f), (21). \end{aligned} \quad (49)$$

Analogously to Algorithm 1, the sequence $\{(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)}, t_1^{(\kappa)}, t_2^{(\kappa)})\}$ generated by Algorithm 3 converges at least to a locally optimal solution of (46).

B. TF-Wise HD TWR Energy-Efficiency Maximization

Similarly to (46), problem (43) of TF-wise HD TWR energy efficiency can be equivalently expressed as

$$\max_{\mathbf{W}, t_1, t_2, \boldsymbol{\alpha}, \boldsymbol{\beta}} \quad \Theta(\mathbf{W}, \boldsymbol{\beta}, t_2) \quad (50a)$$

$$\text{s.t.} \quad (46b), (46c), (46d), (46e), (46f), \quad (50b)$$

$$\begin{aligned} & \ln \left(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}} \right) \\ & + \ln \left(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k} \right) \geq t_2 r_k, \\ & k = 1, \dots, K, \end{aligned} \quad (50c)$$

where

$$\begin{aligned} \Theta(\mathbf{W}, \boldsymbol{\beta}, t_2) \triangleq & \sum_{k=1}^K \left[\ln \left(1 + \frac{|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2}{\sqrt{\alpha_k \beta_{a(k)}}} \right) \right. \\ & \left. + \ln \left(1 + \frac{|\mathcal{L}_{a(k),k}(\mathbf{W})|^2}{\sqrt{\alpha_{a(k)} \beta_k}} \right) \right] / t_2 \pi(\boldsymbol{\beta}, \mathbf{W}, t_1) \end{aligned}$$

with the consumed power function $\pi(\boldsymbol{\beta}, \mathbf{W})$ defined by

$$\begin{aligned} \pi(\boldsymbol{\beta}, \mathbf{W}, t_1) \triangleq & \zeta \left[\sum_{k \in \mathcal{K}} \frac{1}{\sqrt{\beta_k t_1}} + \left(1 - \frac{1}{\sqrt{t_1}} \right) \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) \right. \right. \\ & \left. \left. + \sigma_R^2 \|\mathbf{W}_m\|^2 / \sqrt{t_1} \right) \right] + MP^R + 2KP^U. \end{aligned} \quad (51)$$

Using the inequalities

$$\begin{aligned} & \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) / \sqrt{t_1} \geq \\ & \Phi_{\ell,m}(\mathbf{W}_m^{(\kappa)}, 1, \beta_\ell^{(\kappa)}) / \sqrt{t_1^{(\kappa)}} \\ & + 2 \langle (\mathbf{W}_m^{(\kappa)})^H \mathbf{h}_{\ell,m} \mathbf{h}_{\ell,m}^H, \mathbf{W}_m - \mathbf{W}_m^{(\kappa)} \rangle / \sqrt{\beta_\ell^{(\kappa)} t_1^{(\kappa)}} \\ & - \Phi_{\ell,m}(\mathbf{W}_m^{(\kappa)}, 1, \beta_\ell^{(\kappa)}) (t_1 - t_1^{(\kappa)}) / 2(t_1^{(\kappa)})^{3/2} \\ & - (\beta_\ell - \beta_\ell^{(\kappa)}) \|\mathbf{h}_{\ell,m}^H \mathbf{W}_m^{(\kappa)}\|^2 / 2\sqrt{t_1^{(\kappa)}} (\beta_\ell^{(\kappa)})^{3/2} \end{aligned}$$

and

$$\begin{aligned} \frac{\|\mathbf{W}_m\|^2}{t_1} \geq & \frac{\|\mathbf{W}_m^{(\kappa)}\|^2}{t_1^{(\kappa)}} + 2 \left\langle \frac{\mathbf{W}_m^{(\kappa)}}{t_1^{(\kappa)}}, \mathbf{W}_m - \mathbf{W}_m^{(\kappa)} \right\rangle \\ & - \frac{\|\mathbf{W}_m^{(\kappa)}\|^2}{(t_1^{(\kappa)})^2} (t_1 - t_1^{(\kappa)}) \end{aligned}$$

which follow from the convexity of functions defined in (17), one can obtain

$$\pi(\boldsymbol{\beta}, \mathbf{W}, t_1) \leq \pi^{(\kappa)}(\boldsymbol{\beta}, \mathbf{W}, t_1) \quad (52)$$

where

$$\begin{aligned} \pi^{(\kappa)}(\boldsymbol{\beta}, \mathbf{W}, t_1) \triangleq & \zeta \left[\sum_{k \in \mathcal{K}} \frac{1}{\sqrt{\beta_k t_1}} + \sum_{m \in \mathcal{M}} \left(\sum_{\ell \in \mathcal{K}} \Phi_{\ell,m}(\mathbf{W}_m, 1, \beta_\ell) \right. \right. \\ & \left. \left. + \sigma_R^2 \frac{\|\mathbf{W}_m\|^2}{\sqrt{t_1}} \right) - \sum_{m \in \mathcal{M}} \sum_{\ell \in \mathcal{K}} \left(\frac{\Phi_{\ell,m}(\mathbf{W}_m^{(\kappa)}, 1, \beta_\ell^{(\kappa)})}{\sqrt{t_1^{(\kappa)}}} \right. \right. \\ & \left. \left. + 2 \left\langle \frac{(\mathbf{W}_m^{(\kappa)})^H \mathbf{h}_{\ell,m} \mathbf{h}_{\ell,m}^H}{\sqrt{\beta_\ell^{(\kappa)} t_1^{(\kappa)}}}, \mathbf{W}_m - \mathbf{W}_m^{(\kappa)} \right\rangle \right. \right. \\ & \left. \left. - \frac{\Phi_{\ell,m}(\mathbf{W}_m^{(\kappa)}, 1, \beta_\ell^{(\kappa)})}{2(t_1^{(\kappa)})^{3/2}} (t_1 - t_1^{(\kappa)}) \right) \right] \end{aligned}$$

$$\begin{aligned}
& - \frac{\|\mathbf{h}_{\ell,m}^H \mathbf{W}_m^{(\kappa)}\|^2}{2\sqrt{t_1^{(\kappa)}(\beta_\ell^{(\kappa)})^{3/2}}}(\beta_\ell - \beta_\ell^{(\kappa)}) \\
& - \sum_{m \in \mathcal{M}} \sigma_R^2 \left(\frac{\|\mathbf{W}_m^{(\kappa)}\|^2}{t_1^{(\kappa)}} + 2\langle \frac{\mathbf{W}_m^{(\kappa)}}{t_1^{(\kappa)}}, \mathbf{W}_m - \mathbf{W}_m^{(\kappa)} \rangle \right. \\
& \quad \left. - \frac{\|\mathbf{W}_m^{(\kappa)}\|^2}{(t_1^{(\kappa)})^2}(t_1 - t_1^{(\kappa)}) \right) \Big] + MP^R + 2KP^U,
\end{aligned}$$

which is a convex function.

Suppose that $(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)}, t_1^{(\kappa)}, t_2^{(\kappa)})$ is a feasible point for (50) found from the $(\kappa - 1)$ th iteration. Applying inequality (61) in the appendix for

$$\begin{aligned}
x &= 1/|\mathcal{L}_{k,a(k)}(\mathbf{W})|^2, y = \sqrt{\alpha_k \beta_{a(k)}}, \\
z &= \pi(\boldsymbol{\beta}, \mathbf{W}, t_1), t = t_2
\end{aligned}$$

and

$$\begin{aligned}
\bar{x} &= 1/|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2, \bar{y} = \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}, \\
\bar{z} &= \pi(\boldsymbol{\beta}^{(\kappa)}, \mathbf{W}^{(\kappa)}, t_1^{(\kappa)}), \bar{t} = t_2^{(\kappa)}
\end{aligned}$$

and using inequality (23) yield

$$\frac{\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}})}{t_2 \pi(\boldsymbol{\beta}, \mathbf{W}, t_1)} \geq \frac{\tilde{F}_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}, t_2)}{t_2 \pi(\boldsymbol{\beta}^{(\kappa)}, \mathbf{W}^{(\kappa)}, t_1^{(\kappa)})} \quad (53)$$

over the trust region (21) for

$$\begin{aligned}
x_{k,a(k)}^{(\kappa)} &= |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2 / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}}, \\
p_{k,a(k)}^{(\kappa)} &= 3 \left[\ln(1 + x_{k,a(k)}^{(\kappa)}) \right] / t_2^{(\kappa)} t^{(\kappa)} > 0, \\
q_{k,a(k)}^{(\kappa)} &= x_{k,a(k)}^{(\kappa)} / (x_{k,a(k)}^{(\kappa)} + 1) t_2^{(\kappa)} t^{(\kappa)} > 0, \\
r_{k,a(k)}^{(\kappa)} &= \left[\ln(1 + x_{k,a(k)}^{(\kappa)}) \right] / (t_2^{(\kappa)})^2 t^{(\kappa)} > 0, \\
s_{k,a(k)}^{(\kappa)} &= \left[\ln(1 + x_{k,a(k)}^{(\kappa)}) \right] / t_2^{(\kappa)} (t^{(\kappa)})^2 > 0, \quad (54)
\end{aligned}$$

and

$$\begin{aligned}
\tilde{F}_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}, t_2) &\triangleq p_{k,a(k)}^{(\kappa)} + q_{k,a(k)}^{(\kappa)} \left[2 - \frac{|\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2}{2\Re\{\mathcal{L}_{k,a(k)}(\mathbf{W})(\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)}))^*\} - |\mathcal{L}_{k,a(k)}(\mathbf{W}^{(\kappa)})|^2} \right. \\
&\quad \left. - \sqrt{\alpha_k \beta_{a(k)}} / \sqrt{\alpha_k^{(\kappa)} \beta_{a(k)}^{(\kappa)}} \right] - r_{k,a(k)}^{(\kappa)} t_2 \\
&\quad - s_{k,a(k)}^{(\kappa)} \pi^{(\kappa)}(\boldsymbol{\beta}, \mathbf{W}, t_1). \quad (55)
\end{aligned}$$

By Algorithm 4 we propose a path-following procedure for computing (50), which solves the following convex optimization problem at the κ th iteration to generate the next feasible point

Algorithm 4: Path-Following Algorithm for TF-Wise HD TWR Energy-Efficiency Optimization.

initialization: Set $\kappa = 0$. Initialize a feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)}, t_1^{(0)}, t_2^{(0)})$ for the convex constraints (50a)–(50c) and $e_1 = \Theta(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)}, t_2^{(0)})$.

repeat

- $e_0 = e_1$.
- Solve the convex optimization problem (56) to obtain the solution $(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_1^{(\kappa+1)}, t_2^{(\kappa+1)})$.
- Update $e_1 = \Theta(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_2^{(\kappa+1)})$.
- Reset $\kappa \rightarrow \kappa + 1$.

until $\frac{e_1 - e_0}{e_0} \leq \epsilon$ for given tolerance $\epsilon > 0$.

$$(\mathbf{W}^{(\kappa+1)}, \boldsymbol{\alpha}^{(\kappa+1)}, \boldsymbol{\beta}^{(\kappa+1)}, t_1^{(\kappa+1)}, t_2^{(\kappa+1)}):$$

$$\max_{\mathbf{W}, t_1, t_2, \boldsymbol{\alpha}, \boldsymbol{\beta}} \sum_{k=1}^K [\tilde{F}_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \boldsymbol{\beta}, t_2) + \tilde{F}_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \boldsymbol{\beta}, t_2)] \quad (56a)$$

$$\text{s.t.} \quad (45), (46b) - (46f), (21), \quad (56b)$$

$$\begin{aligned}
f_{k,a(k)}^{(\kappa)}(\mathbf{W}, \alpha_k, \beta_{a(k)}) + f_{a(k),k}^{(\kappa)}(\mathbf{W}, \alpha_{a(k)}, \beta_k) &\geq t_2 r_k, \\
k &= 1, \dots, K, \quad (56c)
\end{aligned}$$

where $f_{k,a(k)}^{(\kappa)}$ are defined from (20).

Analogously to Algorithm 1, the sequence $\{(\mathbf{W}^{(\kappa)}, \boldsymbol{\alpha}^{(\kappa)}, \boldsymbol{\beta}^{(\kappa)}, t_1^{(\kappa)}, t_2^{(\kappa)})\}$ generated by Algorithm 4 converges at least to a locally optimal solution of (50).

An initial feasible point $(\mathbf{W}^{(0)}, \boldsymbol{\alpha}^{(0)}, \boldsymbol{\beta}^{(0)}, t_1^{(0)}, t_2^{(0)})$ for initializing Algorithm 4 can be found by using Algorithm 3 for computing (46), which terminates upon

$$\begin{aligned}
\min_{k=1, \dots, K} \left[\ln(1 + |\mathcal{L}_{k,a(k)}(\mathbf{W})|^2 / \sqrt{\alpha_k \beta_{a(k)}}) \right. \\
\left. + \ln(1 + |\mathcal{L}_{a(k),k}(\mathbf{W})|^2 / \sqrt{\alpha_{a(k)} \beta_k}) \right] / t_2 r_k \geq 1 \quad (57)
\end{aligned}$$

to satisfy (50a)–(50c).

IV. NUMERICAL RESULTS

In this section, simulation results are presented to demonstrate the advantage of the TF-wise HD TWR considered in Section III over FD-based TWR considered in Section II and HD TWR considered in [13]. The channel $\mathbf{h}_{\ell,m}$ from UE ℓ to relay m and the channel $\mathbf{g}_{m,k}$ from relay m to UE k are assumed to be Rayleigh fading, which are modelled by independent circularly-symmetric complex Gaussian random variables with zero means and unit variances. The power of the background noises $\mathbf{n}_{R,m}$ at relay m and n_k at UE k are normalized to $\sigma_R^2 = \sigma_k^2 = 1$. The tolerance for Algorithms 1–4 is set as $\epsilon = 10^{-4}$. Each point of the numerical results is the average of 1,000 random channel realizations. Other settings

TABLE I
MINIMUM PAIR EXCHANGE THROUGHPUT BY TF-WISE TWR FOR $\tau = 1/2$

(K, M, N_R)	(2, 1, 8)	(2, 2, 4)	(2, 4, 2)	(3, 1, 8)	(3, 2, 4)	(3, 4, 2)
$\min_k R_k$	6.29	6.11	5.78	5.66	5.38	4.78

TABLE II
AVERAGE NUMBER OF ITERATIONS FOR COMPUTING (14) BY ALGORITHM 1 WITH $K = 2$

σ_{SI}^2 (dB)	-150	-140	-130	-120	-110
$(K, M, N_R) = (2, 1, 8)$	13.02	14.36	13.24	15.69	18.83
$(K, M, N_R) = (2, 2, 4)$	18.16	17.92	16.80	17.06	19.53
$(K, M, N_R) = (2, 4, 2)$	23.25	18.03	21.09	19.57	21.61

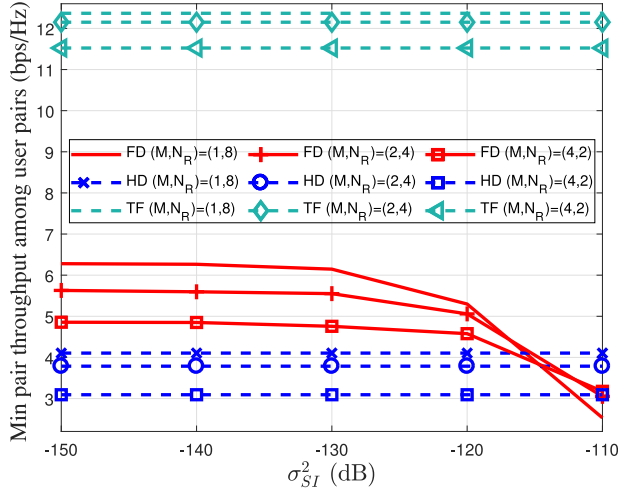


Fig. 3. Minimum pair exchange throughput versus σ_{SI}^2 with $K = 2$.

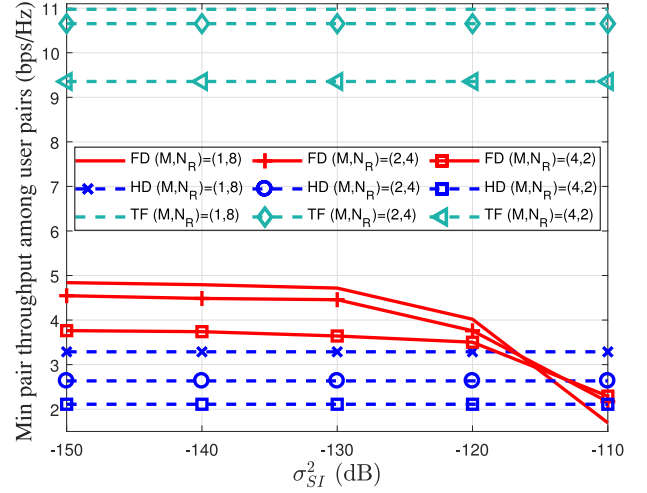


Fig. 4. Minimum pair exchange throughput versus σ_{SI}^2 with $K = 3$.

are: $P_{\text{sum}}^{U, \max} = KP_{\text{sum}}^{U, \max}$ and $P_{\text{sum}}^{R, \max} = MP_{\text{sum}}^{A, \max}/2$, where $P_{\text{sum}}^{U, \max}$ and $P_{\text{sum}}^{R, \max}$ are fixed at 10 dBW and 15 dBW; the drain efficiency of power amplifier $1/\zeta$ is 40%; the circuit powers of each antenna at the relay and UE are 0.97 dBW and -13 dBW. In the algorithms' implementation, the convex solver CVX [28] is used to solve convex optimization problems. Also, the performance graphs are plotted against the self-interference attenuation level σ_{SI}^2 as the latter is the most decisive parameter in FD technologies.

The scenarios of $K \in \{2, 3\}$ pairs and $(M, N_R) \in \{(1, 8), (2, 4), (4, 2)\}$ are simulated.

A. Maximin Information Exchange Throughput Optimization

To confirm the negative effect of the FD SI attenuation level σ_{SI} , Figs. 3 and 4 plot the achievable minimum pair exchange throughput versus SI σ_{SI}^2 with $K \in \{2, 3\}$. For small values of σ_{SI} that put FD SI at the level of the background noise, the minimum pair exchange throughput achieved by FD-based TWR still enjoys the gain offered by FD as is better than that obtained by HD TWR. However, FD cannot offset for larger values of σ_{SI} that make FD SI larger than the background noise, so the former becomes worse than the latter. In contrast, the minimum

pair exchange throughput by TF-wise HD TWR is free of FD SI and it is significantly better than that achieved by the other two. Certainly, using all antennas for separated reception and transmission in time fractions within the time unit is not only much easier to implement but is much better than FD with simultaneous reception and transmission. It has been also shown in [29] and [30] that separated information and energy transfer in time fractions within unit time is more efficient and secure than the simultaneous information and energy transfer. Table I provides the achievable minimum pair exchange throughput attained by TF-wise HD TWR at $\tau = 0.5$, where the users use half of a time slot to send their information to the relays and the relays use the remaining half of the time slot to send the beamformed signals to the users. Comparing with Figs. 3 and 4 reveals that TF-wise HD TWR under this suboptimal time-fraction allocation still outperforms FD TWR slightly and outperforms HD TWR essentially.

Tables II and III provide a computational experience in implementing Algorithm 1, which converges in less than 23 and 36 iterations in all considered FD SI scenarios for solving (14) with $K = 2$ and $K = 3$, respectively. A computational experience in implementing Algorithm 3 is provided by Table IV,

TABLE III
AVERAGE NUMBER OF ITERATIONS FOR COMPUTING (14) BY ALGORITHM 1 WITH $K = 3$

σ_{SI}^2 (dB)	-150	-140	-130	-120	-110
$(K, M, N_R) = (3, 1, 8)$	30.49	27.81	30.26	35.76	26.22
$(K, M, N_R) = (3, 2, 4)$	24.86	26.02	26.31	27.05	31.33
$(K, M, N_R) = (3, 4, 2)$	36.10	24.85	33.47	34.35	22.96

TABLE IV
AVERAGE NUMBER OF ITERATIONS FOR COMPUTING (42) BY ALGORITHM 3

Iterations	K=2	K=3
$(M, N_R) = (1, 8)$	23.55	22.42
$(M, N_R) = (2, 4)$	25.64	25.75
$(M, N_R) = (4, 2)$	25.32	21.43

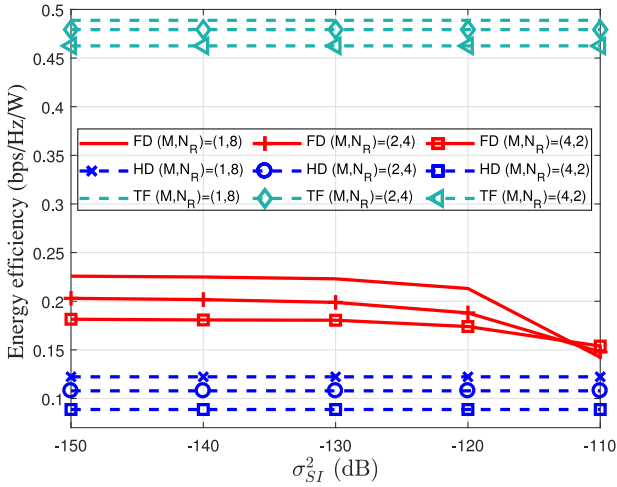


Fig. 5. Energy efficiency versus σ_{SI}^2 with $K = 2$.

which shows that Algorithm 3 converges in less than 25 iterations for solving (42) with $K = 2$ and $K = 3$.

B. EE Maximization

To include a comparison with HD TWR [13], the exchange throughput threshold r_k in (15) and (43) is set as the half of the optimal value of the maximin exchange throughput optimization problem for HD TWR that is computed by [13, Alg. 1].

Fig. 5 plots the energy efficiency attained by the three schemes for $K = 2$. As expected, the two other schemes cannot compete with TF-wise HD TWR. The corresponding sum throughput and transmit power plotted in Figs. 6 and 7 particularly explain the superior performance of TF-wise HD TWR. The sum throughput achieved by TF-wise HD TWR is more than double that achieved by FD-based TWR and HD TWR thanks to its using more power for relay beamforming. In contrast, Fig. 7 shows that the transmit power in FD-based TWR must be controlled to make sure that its transmission does not too severely interfere with its reception. Nevertheless, FD-based TWR always achieves better EE than HD TWR in the considered range of

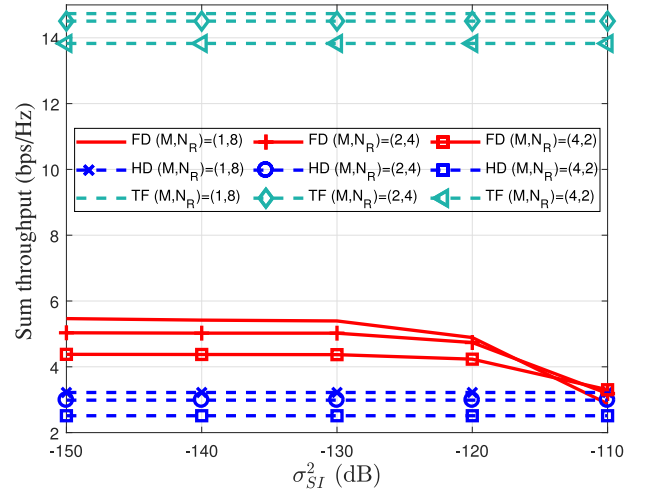


Fig. 6. Sum throughput versus σ_{SI}^2 with $K = 2$.

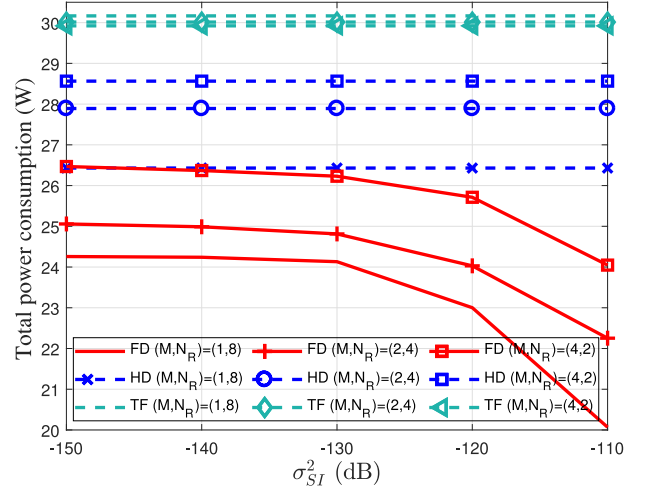


Fig. 7. Total power versus σ_{SI}^2 with $K = 2$.

σ_{SI}^2 though the gap becomes narrower as σ_{SI}^2 . For small σ_{SI}^2 , FD-based TWR achieves higher sum throughput with less transmit power as compared to HD TWR. For larger σ_{SI}^2 , the former achieves almost the same sum throughput as the latter does but with much less transmission power, keeping its EE higher than the latter. Fig. 8 for $K = 3$ follows a similar pattern.

Lastly, Tables V, VI and VII provide a computational experience in implementing Algorithm 2 for solving (15) and Algorithm 4 for solving (43). Algorithm 2 needs less than

TABLE V
AVERAGE NUMBER OF ITERATIONS FOR COMPUTING (15) BY ALGORITHM 3 WITH $K = 2$

σ_{SI}^2 (dB)	-150	-140	-130	-120	-110
$(K, M, N_R) = (2, 1, 8)$	24.85	26.18	21.02	26.63	29.43
$(K, M, N_R) = (2, 2, 4)$	26.49	27.76	26.04	24.18	27.09
$(K, M, N_R) = (2, 4, 2)$	23.87	23.24	24.31	24.65	22.83

TABLE VI
AVERAGE NUMBER OF ITERATIONS FOR SOLVING (15) BY ALGORITHM 3 WITH $K = 3$

σ_{SI}^2 (dB)	-150	-140	-130	-120	-110
$(K, M, N_R) = (3, 1, 8)$	29.40	28.59	30.42	37.31	40.46
$(K, M, N_R) = (3, 2, 4)$	27.81	28.17	30.65	32.45	31.19
$(K, M, N_R) = (3, 4, 2)$	31.75	24.44	26.13	25.37	30.38

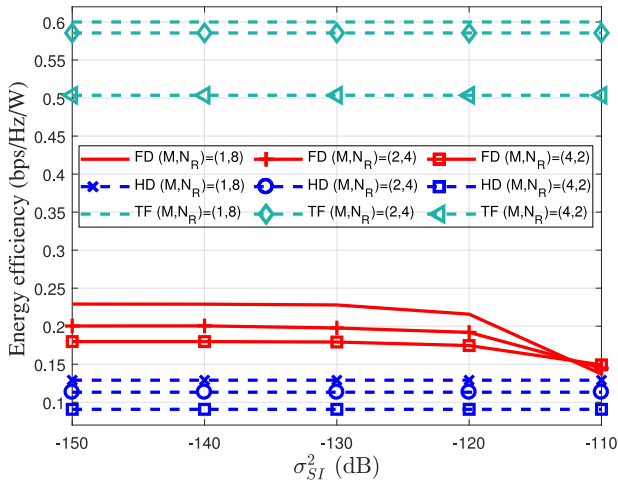


Fig. 8. Energy efficiency versus σ_{SI}^2 with $K = 3$.

TABLE VII
AVERAGE NUMBER OF ITERATIONS FOR COMPUTING (43) BY ALGORITHM 4

Iterations	K=2	K=3
$(M, N_R) = (1, 8)$	20.25	19.38
$(M, N_R) = (1, 8)$	21.51	21.19
$(M, N_R) = (1, 8)$	23.13	24.08

29 and 40 iterations on average for $K = 2$ and $K = 3$, while Algorithm 4 need less than 23 and 24 iterations.

V. CONCLUSION

This paper has considered two possible approaches for multiple pairs of users to exchange information via multiple relays within one time slot. The first approach is based on full-duplexing at the users and relays, while the second approach is based on separated time-fraction-wise half-duplexing signal transmission and reception by the users and relays. It is much easier to implement the second approach than the first approach. In order to compare their capability, we have considered two fundamental problems of joint design of UE power allocation and relay beamforming to optimize the spectral efficiency and energy efficiency. Path-following op-

timization algorithms have been devised for their computation. Simulation results have confirmed their rapid convergence. TF-wise HD TWR has been shown to easily outperform FD-based TWR and HD TWR. The throughput of a network is not only dependent on the bandwidth but is also dependent on the transmit power and interference and noise. Certainly, one does not need to double bandwidth to achieve the same throughput within a half time slot, but one needs to manage the transmit power and interference, as our proposed TF-based approach particularly shows. Additionally, the TF-based approach could exploit more the relay's diversity as it enables the use of all relay antennas for receiving and transmitting signals, which significantly helps to improve the network throughput.

APPENDIX

Let $\mathbb{R}_+^N \triangleq \{(x_1, \dots, x_N) : x_i > 0, i = 1, 2, \dots, N\}$ and $\mathbb{R}_+ \triangleq (0, +\infty)$. In [31], it was proved that the function $\psi(x, y, t) = (\ln(1 + 1/xy))/t$ is convex on $\mathbb{R}_+^3$. Therefore [32]

$$\begin{aligned}
 \frac{\ln(1 + 1/xy)}{t} &= \psi(x, y, t) \\
 &\geq \psi(\bar{x}, \bar{y}, \bar{t}) \\
 &\quad + \langle \nabla \psi(\bar{x}, \bar{y}, \bar{t}), (x, y, t) - (\bar{x}, \bar{y}, \bar{t}) \rangle \\
 &= 2 \frac{\ln(1 + 1/\bar{x}\bar{y})}{\bar{t}} + \frac{1}{(\bar{x}\bar{y} + 1)\bar{t}} \left(2 - \frac{x}{\bar{x}} - \frac{y}{\bar{y}} \right) \\
 &\quad - \frac{\ln(1 + 1/\bar{x}\bar{y})}{\bar{t}^2} t \\
 &\forall (x, y, t) \in \mathbb{R}_+^3, (\bar{x}, \bar{y}, \bar{t}) \in \mathbb{R}_+^3. \tag{58}
 \end{aligned}$$

The right-hand-side (RHS) of (58) agrees with the left-hand-side (LHS) at $(\bar{x}, \bar{y}, \bar{t})$.

Particularly,

$$\begin{aligned}
 \ln(1 + 1/xy) &\geq \ln(1 + 1/\bar{x}\bar{y}) + \frac{1}{(\bar{x}\bar{y} + 1)} \left(2 - \frac{x}{\bar{x}} - \frac{y}{\bar{y}} \right) \\
 &\forall (x, y) \in \mathbb{R}_+^2, (\bar{x}, \bar{y}) \in \mathbb{R}_+^2. \tag{59}
 \end{aligned}$$

Lemma 1: If a function $f(\mathbf{x}, t)$ is convex in $\mathbf{x}$ and $t \in \mathbb{R}_+$ and also is decreasing in t , then the function $f(\mathbf{x}, \sqrt{yz})$ is convex in $\mathbf{x}$ and $(y, z) \in \mathbb{R}_+^2$.

Proof: Since $\sqrt{yz}$ is a concave function, it is true that

$$\begin{aligned} \sqrt{(\alpha y_1 + \beta y_2)(\alpha_1 z_1 + \alpha_2 z_2)} &\geq \alpha_1 \sqrt{y_1 z_1} + \alpha_2 \sqrt{y_2 z_2} \\ \forall \alpha_i &\geq 0, \alpha_1 + \alpha_2 = 1, y_i \geq 0, z_i \geq 0, i = 1, 2. \end{aligned}$$

Therefore

$$\begin{aligned} f(\alpha_1 \mathbf{x}_1 + \alpha_2 \mathbf{x}_2, \sqrt{(\alpha_1 y_1 + \alpha_2 y_2)(\alpha_1 z_1 + \alpha_2 z_2)}) \\ \leq f(\alpha_1 \mathbf{x}_1 + \alpha_2 \mathbf{x}_2, \alpha_1 \sqrt{y_1 z_1} + \alpha_2 \sqrt{y_2 z_2}) \\ \leq \alpha_1 f(\mathbf{x}_1, \sqrt{y_1 z_1}) + \alpha_2 f(\mathbf{x}_2, \sqrt{y_2 z_2}), \end{aligned}$$

showing the convexity of $f(\mathbf{x}, \sqrt{yz})$.

Lemma 2: The Function $f(x, y, t) = (\ln(1 + 1/xy))/t^2$ is convex on $\mathbb{R}_+^3$.

Proof: One has

$$\begin{aligned} \nabla^2 f(x, y, t) &= \begin{bmatrix} \frac{2xy+1}{x^2(xy+1)^2 t^2} & \frac{1}{(xy+1)^2 t^2} & \frac{2}{t^3(xy+1)x} \\ \frac{1}{(xy+1)^2 t^2} & \frac{2xy+1}{y^2(xy+1)^2 t^2} & \frac{2}{t^3(xy+1)y} \\ \frac{2}{t^3(xy+1)x} & \frac{2}{t^3(xy+1)y} & \frac{6 \ln(1+1/xy)}{t^4} \end{bmatrix} \\ &\succeq (x^2 y^2 (xy+1)^2 t^4)^{-1} \\ &\quad \begin{bmatrix} (2xy+1)y^2 t^2 & x^2 y^2 t^2 & 2t(xy+1)xy^2 \\ x^2 y^2 t^2 & (2xy+1)x^2 t^2 & 2t(xy+1)x^2 y \\ 2t(xy+1)xy^2 & 2t(xy+1)x^2 y & 6(xy+1)x^2 y^2 \end{bmatrix}, \quad (60) \end{aligned}$$

because $\ln(1 + 1/t) \geq 1/(t+1) \quad \forall t > 0$ [31, Lemma 1]. Here $\mathbf{A} \succeq \mathbf{B}$ for real symmetric matrices $\mathbf{A}$ and $\mathbf{B}$ means that $\mathbf{A} - \mathbf{B}$ is positive definite.

Then, calculating the subdeterminants of the matrix on the RHS of (60) yields

$$\begin{aligned} (2xy+1)y^2 t^2 &> 0, \\ \begin{vmatrix} (2xy+1)y^2 t^2 & x^2 y^2 t^2 \\ x^2 y^2 t^2 & (2xy+1)x^2 t^2 \end{vmatrix} &= \\ x^2 y^2 t^4 (3x^2 y^2 + 4xy + 1) &> 0, \end{aligned}$$

and

$$\begin{aligned} \begin{vmatrix} (2xy+1)y^2 t^2 & x^2 y^2 t^2 & 2t(xy+1)xy^2 \\ x^2 y^2 t^2 & (2xy+1)x^2 t^2 & 2t(xy+1)x^2 y \\ 2t(xy+1)xy^2 & 2t(xy+1)x^2 y & 6(xy+1)x^2 y^2 \end{vmatrix} &= \\ 12(xy+1)^2 x^5 y^5 t^4 &> 0. \end{aligned}$$

Therefore the matrix on the RHS of (60) is positive definite, implying that the Hessian $\nabla^2 f(x, y, t)$ is positive definite too, which is a necessary and sufficient condition for the convexity of f [32]. ■

By applying Lemmas 1 and 2, the function $\psi(x, y, z, t) = (\ln(1 + 1/xy))/zt$ is convex on $\mathbb{R}_+^4$. Therefore, for all

$(x, y, z, t) \in \mathbb{R}_+^4$, and $(\bar{x}, \bar{y}, \bar{z}, \bar{t}) \in \mathbb{R}_+^4$, it is true that [32]

$$\begin{aligned} \frac{\ln(1 + 1/xy)}{zt} &= \\ \psi(x, y, z, t) &\geq \\ \psi(\bar{x}, \bar{y}, \bar{z}, \bar{t}) + \langle \nabla \psi(\bar{x}, \bar{y}, \bar{z}, \bar{t}), (x, y, z, t) - (\bar{x}, \bar{y}, \bar{z}, \bar{t}) \rangle &= \\ 3 \frac{\ln(1 + 1/\bar{x}\bar{y})}{\bar{z}\bar{t}} + \frac{1}{(\bar{x}\bar{y} + 1)\bar{z}\bar{t}} \left(2 - \frac{x}{\bar{x}} - \frac{y}{\bar{y}} \right) & \\ - \frac{\ln(1 + 1/\bar{x}\bar{y})}{\bar{z}^2 \bar{t}} z - \frac{\ln(1 + 1/\bar{x}\bar{y})}{\bar{z} \bar{t}^2} t. & \quad (61) \end{aligned}$$

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Zhichao Sheng was born in Yangzhou, China. He received the B.S. degree in communication engineering from Nanjing University of Information Science and Technology, Nanjing, China, in 2008, the M.S. degree in signal and information processing from Jiangsu University of Science and Technology, Zhenjiang, China, in 2012, and the Ph.D. degree in electrical engineering from the University of Technology, Sydney, NSW, Australia, in 2018. His research interests include optimization methods for wireless communication and signal processing.



Hoang Duong Tuan received the Diploma (Hons.) and Ph.D. degrees in applied mathematics from Odessa State University, Odessa, Ukraine, in 1987 and 1991, respectively. From 1994 to 2003, he spent nine academic years in Japan. From 1994 until 1999, he was an Assistant Professor with the Department of Electronic-Mechanical Engineering, Nagoya University. From 1999 to 2003, he was an Associate Professor with the Department of Electrical and Computer Engineering, Toyota Technological Institute, Nagoya, Japan. From 2003 to 2011, he was a Professor with the School of Electrical Engineering and Telecommunications, University of New South Wales. He is currently a Professor with the School of Electrical and Data Engineering, University of Technology Sydney, Ultimo, NSW, Australia. His research interests include optimization, control, signal processing, wireless communication, and biomedical engineering.

Professor with the School of Electrical Engineering and Telecommunications, University of New South Wales. He is currently a Professor with the School of Electrical and Data Engineering, University of Technology Sydney, Ultimo, NSW, Australia. His research interests include optimization, control, signal processing, wireless communication, and biomedical engineering.



lished in scientific journals (174 articles) and presented at international conferences (128 papers).

Dr. Duong currently serves as an Editor for the IEEE TRANSACTIONS ON WIRELESS COMMUNICATIONS, IEEE TRANSACTIONS ON COMMUNICATIONS, IET COMMUNICATIONS, and a Lead Senior Editor for IEEE COMMUNICATIONS LETTERS. He was the recipient of Best Paper Awards at the IEEE Vehicular Technology Conference in 2013, IEEE International Conference on Communications in 2014, IEEE Global Communications Conference in 2016, and IEEE Digital Signal Processing Conference in 2017. He is the recipient of a Royal Academy of Engineering Research Fellowship (2016–2021) and received the Newton Prize in 2017.



H. Vincent Poor (S'72–M'77–SM'82–F'87) received the Ph.D. degree in EECS from Princeton University, Princeton, NJ, USA, in 1977.

From 1977 to 1990, he was on the faculty at the University of Illinois at Urbana-Champaign. Since 1990, he has been on the faculty of Princeton, where he is the Michael Henry Strater University Professor of Electrical Engineering. From 2006 to 2016, he was the Dean of Princeton's School of Engineering and Applied Science. He has also held visiting appointments at several other institutions, including

most recently at Berkeley and Cambridge. His research interests include information theory and signal processing, and their applications in wireless networks, energy systems and related fields. Among his publications in these areas is the recent book *Information Theoretic Security and Privacy of Information Systems* (Cambridge University Press, 2017).

Dr. Poor is a member of the National Academy of Engineering and the National Academy of Sciences, and is a foreign member of the Chinese Academy of Sciences, the Royal Society and other national and international academies. He received the Technical Achievement and Society Awards of the IEEE Signal Processing Society in 2007 and 2011, respectively. Recent recognition of his work includes the 2017 IEEE Alexander Graham Bell Medal, Honorary Professorships from Peking University and Tsinghua University, both conferred in 2017, and a D.Sc. *honoris causa* from Syracuse University, awarded in 2017.



Yong Fang (SM'12) received the Ph.D. degree in electronic engineering from City University of Hong Kong, Hong Kong, in 1999. He is currently a Professor with the School of Communication and Information Engineering, Shanghai University, Shanghai, China. His research interests include communication signal processing, blind signal processing, and adaptive information system.