

# Field-Plated Ga<sub>2</sub>O<sub>3</sub> Trench Schottky Barrier Diodes with a $BV^2/R_{on,sp}$ of up to 0.95 GW/cm<sup>2</sup>

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**Abstract**—We report the realization of field-plated vertical Ga<sub>2</sub>O<sub>3</sub> trench Schottky barrier diodes (SBDs). The trench SBDs show significantly lower leakage current than regular SBDs. With employment of field plate, a breakdown voltage (BV) of 2.89 kV is achieved, which is ~500 V higher than those without field plate. Trench sidewall depletion is observed, and the average depletion width is extracted using an analytical model. The trench SBDs have a differential specific on-resistance ( $R_{on,sp}$ ) of 10.5 (8.8) mΩ·cm<sup>2</sup> from DC (pulsed) measurements, which leads to a Baliga's figure-of-merit ( $BV^2/R_{on,sp}$ ) of 0.80 (0.95) GW/cm<sup>2</sup> — the highest among Ga<sub>2</sub>O<sub>3</sub> power devices to date.

**Index Terms**—Ga<sub>2</sub>O<sub>3</sub>, Schottky diodes, Power semiconductor devices, HVPE, Wide band gap semiconductors, MIS devices

## I. Introduction

BETA-Ga<sub>2</sub>O<sub>3</sub> is seen as a promising ultra-wide-bandgap material for power semiconductor devices. On top of the ultra-high critical electric-field ( $E_c$ ) of ~8 MV/cm [1][2], which leads to a projected Baliga's figure-of-merit (FOM) higher than those of SiC and GaN [1], β-Ga<sub>2</sub>O<sub>3</sub> enjoys readily available melt-grown substrates [3], which enable high-quality homoepitaxy as well as a potentially low-cost device platform [4]. The device performance is rapidly progressing in recent years in both lateral and vertical topologies. Over 2 kV breakdown voltage (BV) has been achieved in both lateral Schottky barrier diodes (SBDs) [5] and lateral transistors [6][7]. For high voltage and high current applications, vertical power devices are typically preferred over lateral devices. To date, normally-off vertical fin transistors have been demonstrated with a BV up to 1.6 kV [8][9], alongside with a number of

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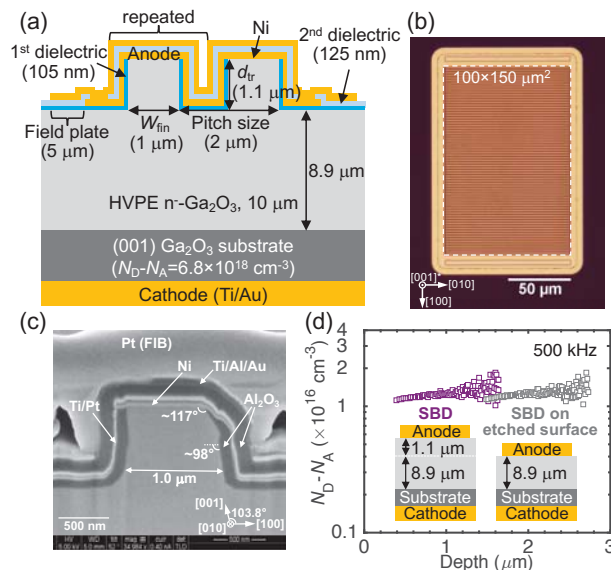


Fig. 1. (a) Schematic cross-section of the field-plated Ga<sub>2</sub>O<sub>3</sub> trench SBDs. (b) Top-view optical image of a fabricated device, where the active device region is outlined by a white dashed rectangle with an area of 100×150 μm<sup>2</sup>. (c) Cross-sectional SEM image of a fin channel. (d) Extracted  $N_D-N_A$  profile from  $C-V$  measurements on regular diodes with a diameter of 200 μm. Inset: schematic cross-section of the two types regular diodes.

reports on vertical SBDs with a BV exceeding 1 kV [10-18].

Despite the encouraging demonstrations of high BV, the Baliga's FOM ( $BV^2/R_{on,sp}$ ,  $R_{on,sp}$  being the differential specific on-resistance of the device) of the present Ga<sub>2</sub>O<sub>3</sub> power devices is still much lower than the material limit, with a highest value of 0.6 GW/cm<sup>2</sup> reached in vertical Schottky barrier diodes [16][17]. To reach the projected high FOM in Ga<sub>2</sub>O<sub>3</sub>, a high parallel-plane electric-field close to  $E_c$  has to be sustained. In the case of vertical SBDs with a typical barrier height of <1.5 eV, however, the surface electric field has to be suppressed to a value much lower than  $E_c$ ; otherwise the leakage current will be excessive [19]. These seemingly conflicting requirements can be fulfilled by the trench MOS barrier Schottky (TMBS) structure (here referred to as trench SBD) [20], which possesses the much-needed charge-coupling effect for surface field reduction without requiring p-type doping. Effective leakage reduction in Ga<sub>2</sub>O<sub>3</sub> trench SBDs has been demonstrated [21, 13-16], as well as an increase of the BV with respect to the regular SBDs, with a highest BV of 2.44 kV [15].

Through the analysis of our previous trench SBDs, we found that the BV is limited by field crowding at either the trench

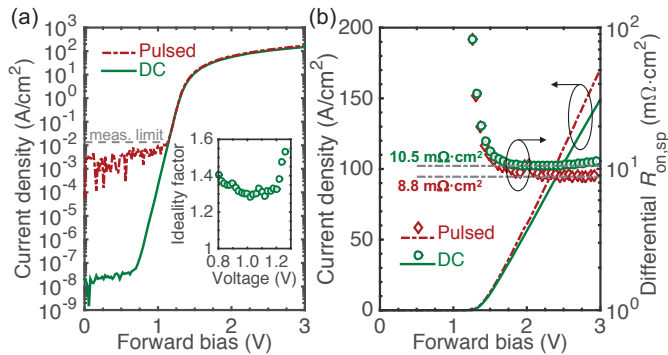


Fig. 2. Forward  $I$ - $V$  characteristics in (a) log scale and (b) linear scale of the trench SBDs with an active area of  $100 \times 150 \mu\text{m}^2$  (also see Fig. 1b for the definition of the device active area). Both DC and pulsed measurements are taken. Inset in (a): extracted ideality factor from DC measurements. The differential  $R_{\text{on,sp}}$  is shown in (b). A base voltage of 0 V, a pulse width of  $1 \mu\text{s}$  and a duty cycle of 0.1% is used in the pulsed measurements.

corners [15] or the device edge [22], and the  $R_{\text{on,sp}}$  is influenced by the fin area ratio ( $A.R.$ ) [15]. To mitigate the field crowding at trench corners, a smaller fin width ( $W_{\text{fin}}$ ) can be designed, since the peak electric field is found to decrease with reducing fin width [15]. The field crowding at trench corners is caused by the positive space charge inside the fin channels. By reducing the fin channel width, the total space charge inside the fin channels is decreased, leading to a reduction of the field crowding at trench corners. To address the field crowding at the device edge, adding a field plate (FP) at the device periphery is a viable technique.

In this work, we improved upon the previous devices by adopting both of these aforementioned approaches: a  $1\text{-}\mu\text{m}$  fin width is designed with a high area ratio of 50%, together with the implementation of a field plate. With such improvements, a BV of 2.89 kV and a record-high Baliga's FOM of 0.8 (0.95)  $\text{GW}/\text{cm}^2$  are achieved under DC (pulsed) condition.

## II. DEVICE FABRICATION

The schematic device cross-section is shown in Fig. 1(a). The  $10\text{-}\mu\text{m}$  epitaxial layer is grown by halide vapor phase epitaxy (HVPE) on a (001) n-type  $\text{Ga}_2\text{O}_3$  substrate. The default fin width and pitch sizes are  $1 \mu\text{m}$  and  $2 \mu\text{m}$ , respectively, leading to a fin area ratio of 50%. The trench depth ( $d_{\text{tr}}$ ) is reduced from  $1.55 \mu\text{m}$  in our previous design [15] to  $1.1 \mu\text{m}$ . The fin channels are designed along [010] direction, which is found to result in the highest forward current due to the lowest interface charge density on the (100)-like fin sidewalls [22]. The length of the field plate is designed to be  $5 \mu\text{m}$ .

The fabrication process flow prior to the addition of the field plate is largely the same as reported previously [15]. After the formation of the fin channel by dry etching, the wafer was soaked in HF for 35 min under ultrasonication to remove the dry etch damage, during which the Ti/Pt (40/60 nm) etching mask was simultaneously removed. Then, Ti/Au (75/150 nm) cathode ohmic contact was formed, followed by the deposition of the 1<sup>st</sup>  $\text{Al}_2\text{O}_3$  dielectric of 105 nm by atomic layer deposition (ALD). After opening of the  $\text{Al}_2\text{O}_3$  on top of the fin channel by dry etching, Ni (30 nm) Schottky contact and then Ti/Pt (40/40 nm) sidewall coverage were deposited by electron-beam

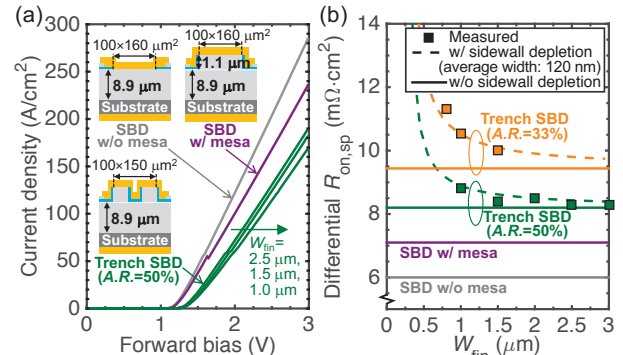


Fig. 3. (a) Comparison of the forward  $I$ - $V$  characteristics under the pulsed condition between trench SBDs and regular SBDs without and with mesa. Insets show the schematic cross-section of the devices. The same pulsed condition as in Fig. 2 is used. (b) Extracted differential  $R_{\text{on,sp}}$  of the trench SBDs under the pulsed condition vs. the fin width ( $W_{\text{fin}}$ ). Dashed and solid line correspond to the calculated  $R_{\text{on,sp}}$  with and without sidewall depletion, respectively.

evaporation and sputtering, respectively. To form the field plate, a 2<sup>nd</sup> ALD  $\text{Al}_2\text{O}_3$  dielectric layer of 125 nm was deposited. Then, contact holes were formed by dry etching to expose the anode metal. Lastly, the Ti/Al/Cu/Au (10/70/10/60 nm) field-plate metal stack was deposited. Figure 1(b) shows the top-view optical image of a fabricated device with an active region area of  $100 \times 150 \mu\text{m}^2$ , which includes both the fins and the regions between fins and is used to calculate the current density and  $R_{\text{on,sp}}$ . Figure 1(c) shows the cross-section of a fin channel imaged by scanning electron microscopy (SEM). A slanted facet is observed at the upper half of the right sidewall. This is likely a result of metal-assisted chemical etching by HF before the complete removal of the Ti/Pt dry-etch mask [23].

## III. RESULTS AND DISCUSSION

Capacitance-voltage measurements were performed on the co-fabricated regular SBDs. As shown in Fig. 1(d), the net doping concentration  $N_{\text{D}}-N_{\text{A}}$  is extracted to be  $1.25 \times 10^{16} \text{cm}^{-3}$ .

Figure 2 shows the forward  $I$ - $V$  characteristics from both DC and pulsed measurements of the trench SBDs with an active area of  $100 \times 150 \mu\text{m}^2$ . An apparent Schottky barrier height of 1.38 eV and an ideality factor of  $\sim 1.3$  is extracted from the thermionic emission model. The pulsed measurements show a higher current density than DC measurements, especially at higher forward bias. This is likely due to the reduction of self-heating and sidewall trapping effect under pulsed conditions [14][24]. The differential  $R_{\text{on,sp}}$  is extracted to be  $10.5 \text{m}\Omega\cdot\text{cm}^2$  from DC and  $8.8 \text{m}\Omega\cdot\text{cm}^2$  from pulsed measurements.

Figure 3(a) shows the comparison of the forward  $I$ - $V$  characteristics under pulsed condition among three types of devices: regular SBDs with and without mesa, and trench SBDs. Similar with our previous observations [15], the current density is highest in SBDs without mesa and lowest in trench SBDs as a result of the extra conduction path in the  $1.1\text{-}\mu\text{m}$  upper drift region as well as the restriction of the current path due to the fin channels. In addition, we observed that the current density is slightly lower in trench SBDs with smaller  $W_{\text{fin}}$ , even with the same fin area ratio. We ascribe it to sidewall depletion in the fin channels due to negative interface charges [22]. Considering an average sidewall depletion width of  $W_{\text{d}}$ , the  $R_{\text{on,sp}}$  of trench SBDs ( $R_{\text{on,sp,trench}}$ ) can be generically modeled by

$$R_{on,sp,trench} = R_1 + \frac{R_2}{\left(1 - \frac{2W_d}{W_{fin}}\right) \cdot A.R.},$$

where  $R_1$  is the  $R_{on,sp}$  of all the components *below* the fin channels, including the drift layer (8.9- $\mu\text{m}$  thick, also see Fig. 1a), the substrate and the back contact, while  $R_2$  is the  $R_{on,sp}$  of the fin channel (1.1- $\mu\text{m}$  tall in the trench SBDs). Since the device active area is nearly the same in the three types of devices shown in Fig. 3(a), the current spreading effect below the 1.1- $\mu\text{m}$  upper drift region should be similar. Thus,  $R_1$  can be approximated with  $R_{on,sp}$  of the SBD without mesa. If  $A.R.$  is taken to be 100%, the trench SBD will become the SBD with mesa, thus  $R_2$  can be obtained by the difference in  $R_{on,sp}$  between the SBDs with and without mesa.

Figure 3(b) shows the measured  $R_{on,sp}$  of the trench SBDs with 2 different  $A.R.$  and a few  $W_{fin}$  values, all fabricated on the same wafer. If there is no depletion or accumulation in the fin channel,  $W_d=0$  thus  $R_{on,sp,trench}$  should be independent of the fin width  $W_{fin}$ . The increase of  $R_{on,sp}$  with decreasing  $W_{fin}$  suggests that sidewall depletion does exist and impact the  $R_{on,sp}$  negatively. With the analytical model above, an average depletion width  $W_d$  of  $120\pm 30$  nm is extracted. Detailed study of the interface charge responsible for the sidewall depletion is beyond the scope of this letter and will be presented in our future papers. In these devices, the interface-state density ( $D_{it}$ ) on the fin sidewall is estimated to be  $\sim 8\text{-}9 \times 10^{11} \text{ cm}^{-2} \text{ eV}^{-1}$ , and the total interface-charge density as a function of the stress voltage behaves similarly with our previous devices [14]. Without the sidewall depletion, the  $R_{on,sp,trench}$  is expected to be  $8.2 \text{ m}\Omega\cdot\text{cm}^2$  for an  $A.R.$  of 50% based on the model. If accumulation condition at the MIS-interface can be reached, the  $R_{on,sp,trench}$  can be even lower, comparable to that of regular SBDs.

Figure 4(a) shows the reverse  $I$ - $V$  characteristics of the trench SBDs. In comparison with the regular SBDs with mesa and FP, the field-plated trench SBDs have much lower leakage current as well as a much higher BV of 2.89 kV. The breakdown of the trench SBDs is destructive, with visible breakdown craters near the device edge. To investigate the breakdown mechanism of

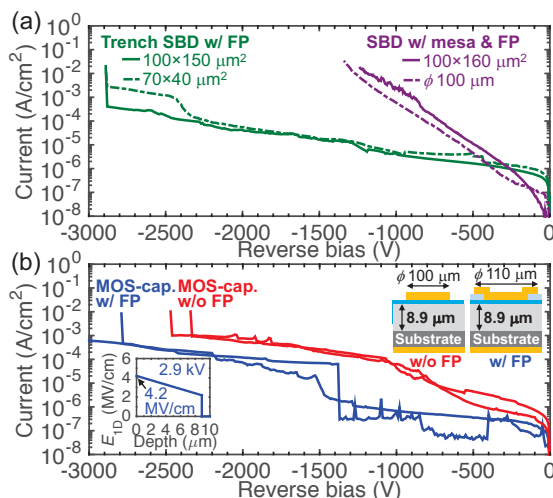


Fig. 4. Comparison of the reverse  $I$ - $V$  characteristics between (a) field-plated trench SBDs and SBDs with mesa, (b) MOS-capacitors with and without field plate. Insets in (b) shows the schematic structures of the MOS-capacitors and the calculated one-dimensional electric field profile along the vertical direction at 2.9 kV in the MOS-capacitors.

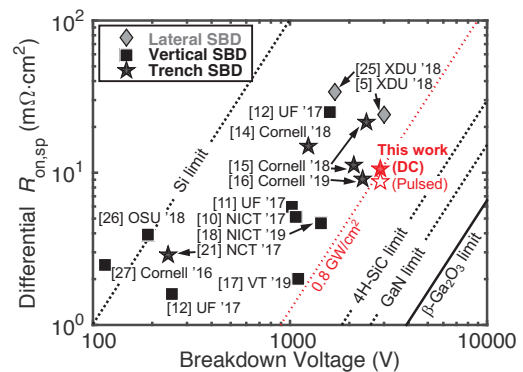


Fig. 5. Benchmark plot of state-of-the-art  $\beta$ - $\text{Ga}_2\text{O}_3$  SBDs, including lateral [5][25], vertical [10-12,17,18,26,27] and trench SBDs [21,14-16].

the trench SBDs, breakdown tests were also performed on MOS-capacitors without and with field plate, which are co-fabricated on the same wafer and have the same dielectric thickness and field-plate structure as the trench SBDs. As shown in Fig. 4(b), with addition of the field plate, the destructive BV of MOS-capacitors increases from  $\sim 2.4$  kV to  $\sim 2.9$  kV. The observed trend in BV also suggests that the breakdown is at the device edge, evidenced by the breakdown craters observed near the edge of the MOS-capacitors. The increase in BV with the addition of the field plate confirms its effectiveness in alleviating the edge field crowding, as also shown in our previously reported  $\text{Ga}_2\text{O}_3$  transistors [9] and SBDs [16]. Since the BV of trench-SBDs is very close to that of field-plated MOS-capacitors, the breakdown of the trench SBDs is most-likely limited by the edge field crowding, which is found to be quite prominent despite the presence of the field plate [16]. In field-plated MOS-capacitors, the parallel-plane one-dimensional electric field ( $E_{1D}$ ) at the  $\text{Ga}_2\text{O}_3$  surface is calculated to be 4.2 MV/cm at 2.9 kV (see inset in Fig. 4(b)), which is higher than the  $E_c$  of SiC and GaN, but still much lower than the  $E_c$  of  $\text{Ga}_2\text{O}_3$ . With a more ideal edge termination structure, a higher  $E_{1D}$  can be reached.

Figure 5 shows the benchmark plot of reported  $\beta$ - $\text{Ga}_2\text{O}_3$  lateral SBDs [5][25], vertical SBDs [10-12,17,18,26,27] and trench SBDs [21,14-16]. The field-plated trench SBD in this work achieves a FOM ( $\text{BV}^2/R_{on,sp}$ ) of  $0.80 \text{ GW}/\text{cm}^2$  from DC measurements, which is the highest among all  $\text{Ga}_2\text{O}_3$  power devices to date. From pulsed measurements, the FOM is even higher at  $0.95 \text{ GW}/\text{cm}^2$  as a result of the slightly lower  $R_{on,sp}$  of  $8.8 \text{ m}\Omega\cdot\text{cm}^2$ . As shown in Ref. 17, at a FOM higher than  $0.6 \text{ GW}/\text{cm}^2$ , the performance of  $\text{Ga}_2\text{O}_3$  SBDs is already comparable with the state-of-the-art GaN SBDs. The gap between the present device performance and the material limit can be bridged with better electric field managements.

#### IV. CONCLUSION

We demonstrated field-plated  $\text{Ga}_2\text{O}_3$  vertical trench Schottky barrier diodes. The addition of field plate is found to increase the breakdown voltage by  $\sim 500$  V. We observed the existence of sidewall depletion with an extracted average width of  $120\pm 30$  nm. The field-plated trench SBDs have a  $R_{on,sp}$  of  $10.5$  ( $8.8$ )  $\text{m}\Omega\cdot\text{cm}^2$  from DC (pulsed) measurements and a breakdown voltage of 2.89 kV, corresponding to a record-high Baliga's figure-of-merit of  $0.80$  ( $0.95$ )  $\text{GW}/\text{cm}^2$ .

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