

Design and Fabrication of Flow-Based Edge Detection Memristor Crossbar Circuits¹

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Abstract—We design and fabricate a flow-based circuit for edge detection in images that exploits device-level parallelism in nanoscale memristor crossbars. In our approach, a corpus of human-labeled edges in BSDS500 images is used to learn an edge detection function with ternary values: true, false, and don't-care. A Boolean crossbar design implementing an approximation of this ternary function using in-memory flow-based computing is then obtained using a massively parallel simulated annealing search executed on GPUs. We demonstrate the success of our approach by fabricating the memristor circuit on a 300mm wafer platform using a custom 65nm CMOS/ReRAM process technology. We demonstrate that our flow-based computing approach is either faster, more energy-efficient or produces fewer incorrect edges than other competing approaches. We show that our design has power and area requirements that are 3.3x and 2.5x lower, respectively, than the previous state-of-the-art.

Index Terms—Computer-aided design, memristor, flow-based computing, vision, AI.

I. INTRODUCTION

MULTIPLE computer vision and AI algorithms rely on edge detection as a preliminary step [1]. Fast and efficient edge detection in images using dedicated hardware can enhance the usability of these algorithms, specially in edge computing and IoT. Flow-based in-memory computations has been shown to be both time and energy-efficient for simple arithmetic operations. These advantages of flow-based computing are obtained by bypassing the memory bottleneck of traditional von Neumann architectures and exploiting the device-level parallelism of nanoscale memristor crossbars.

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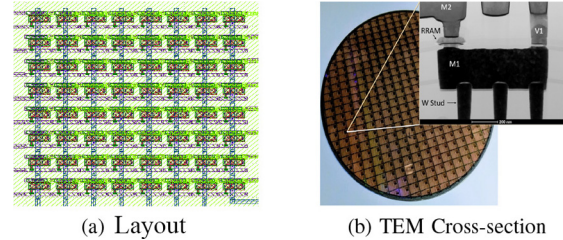


Fig. 1. A layout and TEM cross-sectional image of the HfO₂ ReRAM device implementing our edge detection crossbar design [3]. The custom ReRAM module was fabricated on a 300mm wafer using a 65nm CMOS process.

In this brief, we design and fabricate an edge detector that leverages the time and energy efficiency of flow-based computing on memristor crossbars. We create a ternary-valued function derived from manually segmented images of the BSDS500 dataset as the ground truth for edge detection [2]. The ternary function maps a pixel pair to a true, false, or don't-care value, which corresponds to an edge, not an edge, and an uncertainty about the existence of an edge between a pixel pair. A crossbar implementing this ternary function is then found using simulated annealing on more than 4000 GPU cores.

An objective function based on the human perception of image similarity is used to control the simulated annealing based search. A massively parallel simulated annealing search is employed to find crossbar designs of multiple sizes with varying accuracy and energy requirements. We design memristor crossbars of sizes 5×5 , 6×6 , 7×7 , and 8×8 , and then fabricate them on a custom ReRAM module on a 300mm wafer platform using a 65nm CMOS process technology. We experimentally demonstrate that our designs can be successfully fabricated on physical devices. We make the following new contributions in this brief:

- 1) We exploit a massively-parallel simulated annealing search for the optimal crossbar design using two Tesla V100s with more than 4000 GPU cores. We design a new method to calculate the output of crossbar circuits efficiently. This parallel approach combined with an efficient calculation of crossbar output allows us to search for smaller crossbars designs that have lower power consumption. When compared to earlier work [4], our design produces crossbars that are up to 2.5x smaller and have up to 3.3x lower power requirements.

- 2) We employ the human perception of similarity between two images as the cost metric for the search algorithm. This allows us to create crossbar designs that generate edges that are visually similar to the ground truth. We compare the edges generated by our designs to those generated by earlier work [4] and find a 2.78x improvement in the perceptual difference score.
- 3) We have fabricated and tested our designs on a physical 1T1R ReRAM module to show that our solution is both feasible and practical, shown in Figure 1(a). A single fabricated ReRAM device implementing the edge detection crossbar design is shown in Figure 1(b). Previous work to create a flow-based ReRAM device was published in ISCAS 2016 by Alamgir *et al.* [5]. But, that work only fabricated a relatively simple one-bit adder. This brief brings the advantages of in-memory flow-based computing to a fundamental computer vision problem.

II. RELATED WORK

There has been continued interest in logic design using memristor crossbars [6]–[9]. Various techniques that leverage the unique properties of memristor crossbars to perform neuromorphic computing have been proposed in literature [10]–[13]. These machine learning applications of memristor crossbars are impressive, but rely on integration with CMOS devices to perform computations. Flow-based computing on memristor crossbars is a departure from the traditional computing paradigm and uses device-level parallelism to enable in-memory computations that can overcome the memory bottleneck of the von Neumann architecture. A memristor crossbar implementing a one-bit adder using flow-based computing was presented in [14]. Subsequent work on flow-based computing used binary decision diagrams (BDD) [15], free binary decision diagrams (FBDD) [16], automated synthesis via satisfiability modulo theory [5], and AI-based search procedures [17] to synthesize memristor crossbar circuits. Alamgir and others presented a full adder crossbar implementation on a ReRAM device [5].

The design of edge detection circuits using flow-based memristor crossbars has been presented in [4]. However, they did not demonstrate that their designs can be fabricated on real-world memristors. Our designs are both more compact and more energy-efficient than the designs reported in [4]. Pajouhi and Roy designed a memristor circuit based on ant colony optimization to perform edge detections [18]. Our approach is both space and energy-efficient when compared to the input-aware flow-based approach, while it is faster than the ant colony optimization based approach. A comparison of our approach to these approaches is presented in Table I. The area of our custom 1T1R crossbars is considerably increased due to the integrated transistors in each cell as compared to a 1R crossbar.

III. APPROACH

The design of our edge detection circuit is based on flow-based computing [5], where an electric pulse is applied to one nanowire, and the output current is observed from another

TABLE I
COMPARISON OF THE PERFORMANCE OF OUR APPROACH TO INPUT-AWARE FLOW-BASED [4] AND SWARM-BASED APPROACHES [18]. THE 8×8 TO 5×5 CROSSBARS HAVE BEEN GENERATED USING OUR APPROACH. OUR DESIGN GENERATES EDGES THAT ARE SIMILAR TO THE GROUND TRUTH, AS DEMONSTRATED BY THE HIGH PEAK SIGNAL-TO-NOISE RATIO (PSNR) VALUES AND LOWER PERCEPTUAL DIFFERENCE SCORES [19]. OUR DESIGN USES LESS POWER THAN [4], AND IT IS FASTER THAN [18]

Designs	PSNR	Perceptual Diff Score	Switching Power	Time	Area
Input-aware [4]	8.9	20458	1.188mW	-	-
Swarm-based [19]	-	-	0.220mW	68 μ S	37.22 μ m ²
8×8 crossbar	13.7	6563	0.858mW	16 μ S	756 μ m ²
7×7 crossbar	14.6	5359	0.528mW	16 μ S	578.81 μ m ²
6×6 crossbar	14.5	5401	0.440mW	16μS	425.25 μ m ²
5×5 crossbar	14.4	5818	0.352mW	16 μ S	295.31 μ m ²

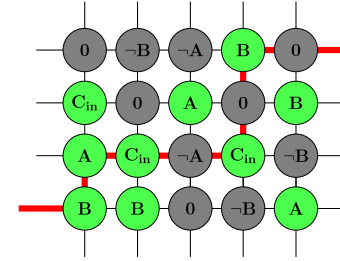


Fig. 2. A crossbar design that implements a full adder. Each memristor is labeled with an input. The states of the memristors after receiving an input is updated according to the input. For inputs $A = 1$, $B = 1$ and $C = 1$ the memristors labeled A , B and C are set to ON, shown by green, and memristors labeled $-A$, $-B$ and $-C$ are set to OFF, shown by grey. The red path shows the current flow from the input wire (bottom row) to the output wire (top row).

nanowire. The crossbar array consists of programmable and non-programmable memristors. Programmable memristor contains the current input value and can change depending upon the input, whereas non-programmable memristors stay constant and do not change. An illustration of this approach implementing a full adder on a 4×5 crossbar is shown in Figure 2. The value of the input determines the programming of the memristors. Memristors in green have been set to ON, and memristors in gray have been set to OFF. Current flow to calculate the sum when the values of $A = 1$, $B = 1$ and $C = 1$ is shown by the red path. Current flow in the top row implies that the sum is 1, whereas no flow implies that the sum is 0.

More complex boolean functions that perform edge detection can be implemented on larger crossbars. Figure 3 shows a crossbar synthesised by our approach to perform edge detection in images. Each crossbar accepts a pixel pair and a flow in the rightmost column indicates an edge between a pixel-pair. The problem of finding the memristor design that can effectively find edges between the pixel pairs in the image is solved in two steps: (1) finding the ternary function which performs edge detection, and then (2) finding a crossbar design that implements the ternary function found in step (1).

A. Ternary Value Function for Edge Detection

The ternary function maps pixel pairs to true, false, and don't-care values. True value denotes the presence of an edge between the pixel, false value denotes the absence of an edge,

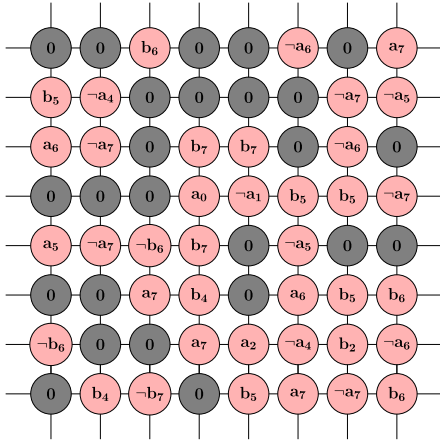


Fig. 3. A 8×8 memristor crossbar design for edge detection. The crossbar is initialized using pixel pair (A, B) , each pixel is 8-bit in size. Pixel $A = \{a_7, a_6, \dots, a_0\}$ and $B = \{b_7, b_6, \dots, b_1\}$. Each memristor is labeled with a_i or b_i for $i \in \{0, 1, \dots, 7\}$. A memristor labeled $a_i = 1$ or $b_i = 1$ is set to On and $a_i = 0$ or $b_i = 0$ is set to OFF.

and the don't-care values denote pixel pairs where the human-labeled data set does not indicate a consistent response. Edge information obtained from the human-annotated BSDS500 dataset is used to calculate the pixel pair to value mapping using the following equations:

$$V(x, y) = \begin{cases} T, & \text{if } f_e(x, y)/f_p(x, y) \geq \theta^E \text{ and } f_e(x, y) \geq \theta^P \\ F, & \text{if } f_e(x, y)/f_p(x, y) < \theta^E \text{ and } f_e(x, y) \geq \theta^P \\ \text{Don't-care}, & \text{otherwise} \end{cases}$$

Here, x, y are values of the pixel pair, $f_p(x, y)$ is the frequency of occurrence of pixel having values x and y in the image dataset, and $f_e(x, y)$ is the frequency of observing an edge between pixel pairs x and y in the human-annotated dataset. The parameter θ^P and θ^E are the values of the threshold of f_p and f_e , respectively, that determines the mapping from pixel pair to the ternary value.

B. Ternary Function to Crossbar Design

A massively parallel implementation of the simulated annealing algorithm is used to search for crossbar designs that implement the approximated ternary function. The cost function of the simulated annealing algorithm is designed to penalize disagreement between the ternary function and the crossbar output, and the disagreement between the generated edges and the human-annotated edges. Flow-based crossbar designs produce Boolean values as output and generate a true or false result on a given input, whereas the ternary function produces a true, false, and don't-care as output. The disagreement D_T between the crossbar output $C(x, y)$ and the ternary function $V(x, y)$ is calculated using the following function:

$$d(x, y) = \begin{cases} 0, & C(x, y) = V(x, y) \\ 0, & \text{if } V(x, y) = \text{Don't Care} \\ 2, & \text{if } V(x, y) = T \text{ and } C(x, y) \neq T \\ 1, & \text{if } V(x, y) = F \text{ and } C(x, y) \neq F \end{cases}$$

$$D_T = \sum_{x, y} d(x, y)$$

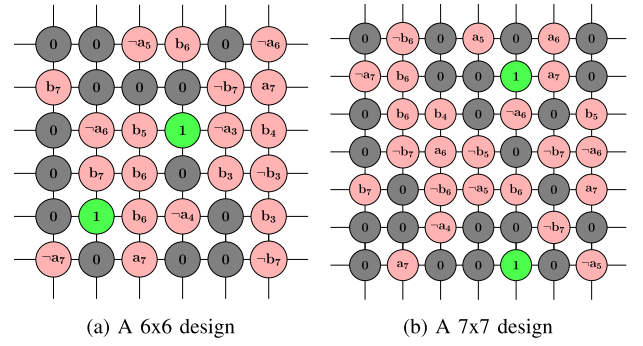


Fig. 4. Edge detection memristor crossbar designs implemented using 7×7 and 6×6 crossbars.

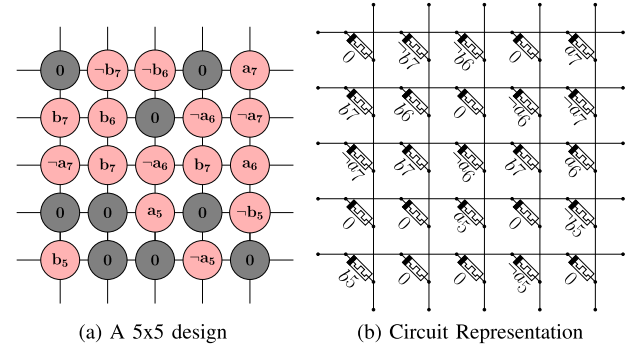


Fig. 5. Edge detection memristor crossbar design implemented using 5×5 and the corresponding circuit representation.

Here, x, y are values of the pixel pair, $V(x, y)$ is the ternary value function output for a given pixel pair, and $C(x, y)$ is the crossbar output. The disagreement D_T is then combined with an image similarity score that can capture the perception of a human observer to find the final disagreement. In our method, we have used the inverse of the perceptual difference (PerceptualDiff) score to obtain the total disagreement D .

Obtaining the crossbar output $C(x, y)$ to calculate the disagreement between the ternary function and the crossbar output in simulation can be time-consuming and can lead to long search time. Evaluating the truth table entries from a crossbar naively has a time complexity of $O(2^b RC)$, where R and C are the numbers of rows and columns respectively in the crossbar, and b is the size of the input. To quickly calculate the output of crossbars circuits, we model the circuit as a directed acyclic graph (DAG) and then only compute incremental changes in the total disagreement D as the design evolves during our search process.

Modelling crossbar circuit as a DAG: For a crossbar of size $R \times C$ with R rows and C columns, a Directed Acyclic Graph (DAG) G can be constructed to emulate the dynamics of the crossbar. The DAG G is divided into components G^k arranged temporally, where each component emulates the dynamics happening within the time it takes for current to cross one memristor in the crossbar. Each component G^k consists of nodes $r_i^{(k)}$ and $c_j^{(k)}$, which represent i th row and j th column wires respectively, and directed edges $e(r_i^{(k)}, c_j^{(k)})$ that can capture the flow of current through a memristor from wire

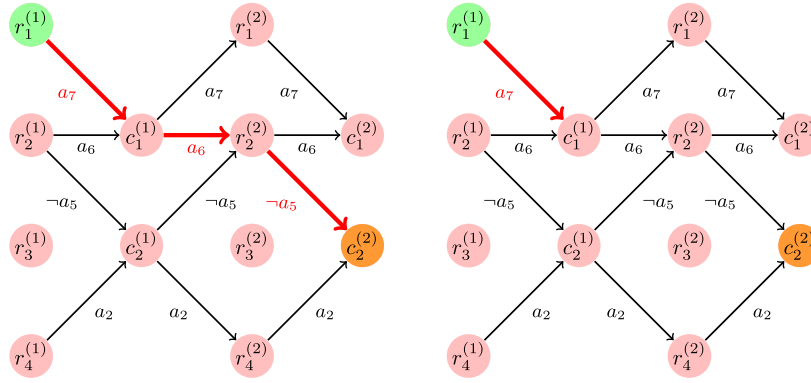


Fig. 6. The input and output nodes of the DAG are represented by $r_1^{(1)}$ and $c_2^{(2)}$ respectively. **(left)** Evidence $E = \{a_7, a_6, \neg a_5\}$ of true in the truth table. Knowing the values of three variables tells us that output of the crossbar is true. **(right)** Evidence $E' = \{\neg a_7\}$ of false in the truth table. Knowing the value of one variable tells us that the output of the crossbar is false.

r_i to c_j at time step k . The components G^{k-1} and G^k are connected by directed edges $e(c_i^{(k-1)}, r_j^{(k)})$ which capture the flow of current through a memristor from wire c_i to r_j at time step $k-1$. An edge between two wires exists if memristor connecting the wires is a programmable turned-on memristor or a non-programmable memristor that is always turned-on. The total number of components K in the graph depends upon the size of the crossbar and is equal to RC . The input pulse is applied to the bottom row and is denoted by the node $r_1^{(1)}$ in the DAG. The output node is denoted by $c_C^{(K)}$.

Given a pair of input pixel, an edge exists between the pixels if there is a flow from the input node to the output node. Such a path is called evidence E for a true truth table entry for the input. We avoid the explicit calculation of flow from the input to the output for a given crossbar by utilizing a special case of the max-flow min-cut problem, where the flow of each edge is equal to unity, and the source and sink nodes are the input and output nodes respectively. The set of edges that form the min-cut of DAG is the evidence E' for a false entry of the truth table. There exists a set of evidences $\mathcal{E} = \{E_1, E_2, \dots, E_t\}$ and $\mathcal{E}' = \{E'_1, E'_2, \dots, E'_t\}$ which account for all the output of the crossbar. Knowing the input value of only a few variables that satisfy an evidence can allow us to know the output of the whole crossbar and removes the need for expensive calculations. An example of this approach is shown in Figure 6. For an 8-bit input, observing a false value of input a_7 allows us to know that the output of the whole crossbar is false irrespective of other values of the input.

IV. RESULTS

We synthesized memristor crossbar designs of sizes 5×5 , 6×6 , 7×7 and 8×8 using our approach. Our approach of using a ternary function with a don't-care condition allowed us to skip 54% of the input pixel pairs while searching for crossbar designs. The crossbar designs are presented in Figures 3, 4 and 5. We tested our design on the BSDS500 dataset and show that the edges computed by our design have lower power to signal noise ratio (PSNR) and higher perceptual difference (PerceptualDiff) score when compared to earlier results [4]. We use the memristor programming circuits

TABLE II
OUTPUTS ON AN 8×8 1T1R DEVICE USING RANDOMLY SELECTED
PIXEL PAIRS. LOGICAL 0 CORRESPONDS TO ABOUT 3.5V
AND LOGICAL 1 CORRESPONDS TO 1.5V – 1.7V

Pixel A	Pixel B	Expected logical output	Observed output
01100000	01101001	0	3.5V
00110110	00100110	0	3.5V
01101111	10010100	1	1.7V
10011110	01111100	1	1.6V
10010000	01111100	1	1.7V
01001100	01110111	1	1.5V

provided in [18] and [5] to compare the performance of the fabricated devices and designs. We compare the speed of edge detection by our crossbar-based memristive computing design to the swarm-based approach presented in [18], and show that our memristor crossbar design takes less time to compute edges than the swarm-based approach.

In Table I, we observe that the best PSNR and PerceptualDiff score between the ground truth and the computed edge are 14.6 and 5359 respectively, whereas the PSNR value of the input-aware method is lower at 8.9, while the PerceptualDiff score is higher at 20458. A higher PSNR and a lower perceptual difference (PerceptualDiff) score denote higher conformance of the ground truth with the computed edge. Similarly, we observe in Figure 7 that the edges computed by our method have less noise and are closer to the ground truth when compared to the input-aware crossbar design. Our 5×5 design is 2.5 times smaller than the designs produced by the input-aware method, and requires 0.418mW of power, which is 3.3 times less than the power required by the input-aware crossbar design.

The synthesized designs have been verified experimentally on an 8×8 1T1R device array fabricated on a 300mm wafer. We chose input pixel pairs, programmed them on the ReRAM device, applied a current of $2\mu A$ to the input, and observed the output voltages difference across the input and the output. A high resistance state (HRS), leading to an observed high voltage implies no edge, while a low resistance state (LRS) leading to a low voltage implies the existence of an edge. The result of the experiments is presented in Table II. We observe

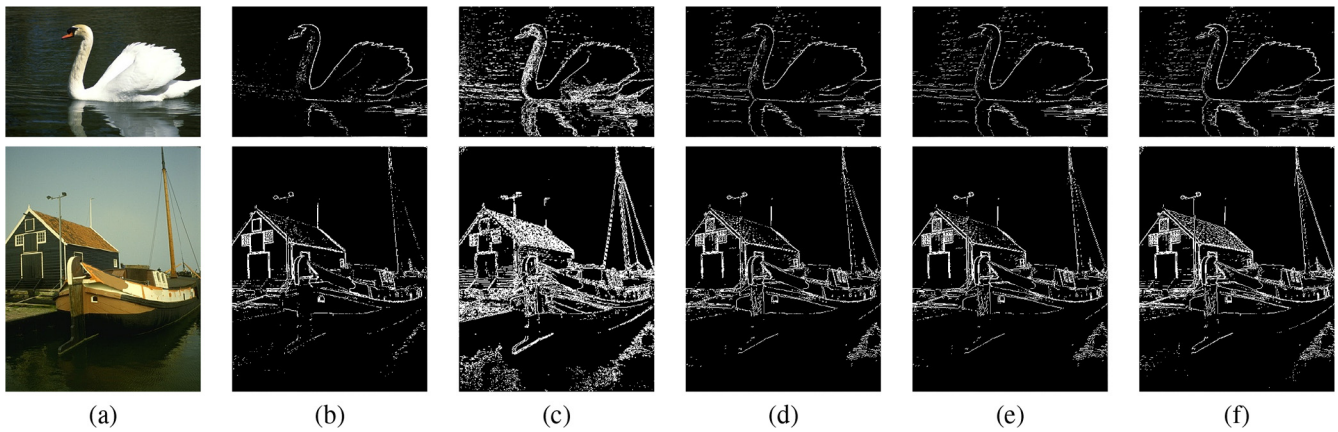


Fig. 7. (a) Original image, (b) Human annotated ground truth, (c) Edges generated using input-aware crossbar design [4], (d), (e) and (f) Edges generated using our 5x5, 6x6 and 8x8 crossbar designs respectively. The number of pixels contributing to noise in (c) is greater when compared to our designs. For example, in (c), there is a lot of noise on the roof of the house; our designs generate edges that are similar to the ground truth.

a voltage difference of 1.9V between high and low resistance states. These results experimentally verify the correctness of the first flow-based edge detection design implemented on a ReRAM device.

V. CONCLUSION AND FUTURE RESEARCH

We present the design and fabrication of an edge detection circuit using flow-based computing in nanoscale memristor crossbars. Our work is the first to experimentally fabricate a flow-based computing for a practical application and produces designs that are 2.78x better in the perceptual difference score, 2.5x smaller and 3.3x more energy-efficient than the state-of-the-art [4].

Crossbar design for a broader distribution of images is an important direction we are going to pursue in the immediate future. We will focus on crossbar synthesis for other relevant applications like convolution, clustering, regression and pattern-matching.

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