



packet loss condition is proposed in [13]. In fusion center, the local estimation results are combined by designed weighting parameters, which are obtained through optimization process. For real-time implementation point of view, it is an unrealistic process as it involves two-stages (adaptive and combine) and requires more time to properly estimate grid states. Finally, computing the optimal weight parameters and convergence analysis are challenging tasks.

From filtering point of view, the Kalman filter (KF), extended KF (EKF), H-infinity EKF, unscented KF and cubature KF algorithms are used for power system state estimations [14], [15]. Moreover, the forecasted-aided KF algorithm considering cyber attack is explored in [16], where Euclidean distance metric is used to detect cyber attack. The observer-based anomaly detection scheme is presented in [17]. In addition, the wavelet transform-based mixed Kalman particle filter algorithm under FDIA is presented in [18]. The scenario based unsupervised learning algorithm for cyber physical power system is developed in [19]. All the aforementioned algorithms are designed for centralized state estimation, which requires all measurements and prone to vulnerable and single point failure. Due to deregulation of power systems, the distributed state estimation is gaining more attention in industrial and research communities.

In order to estimate the discrete time-varying cyber-physical system states, an iterative finite impulse response filter is designed [20]. It can effectively estimate the hidden system states without using any specific initialization scheme. For improvement of estimation accuracy, the robust type chandrasekhar-based maximum correntropy KF algorithm for cyber-physical system is proposed in [21]. The idea is extended in [22], where attack-resilient remote state estimation scheme is proposed and verified. The attackers are manipulated sensory measurements and the fusion center combines them for state estimations. Using residual prewhitening method, the cyber attack detection method is proposed in [23]. Technically, when the covariance matrix of the residual error is not full-rank, this method is used to solve the cyber attack detection and estimation problem.

Moreover, the distributed H- $\infty$  algorithm for a virtual system is presented in [24]. The designed gains of the distributed filter are obtained by offline, and they are computed after solving several linear matrix inequalities (LMIs). The computational complexity of this approach is very high and cannot apply in real-time power systems. Moreover, the power system state estimation is jointly developed with an innovation-based cyber attack detection method to limit communication overhead in [25]. The alternating direction method of multipliers-based distributed state estimation algorithm for power system is presented in [26]. Moreover, the distributed optimal estimation algorithm for sensor networks are developed in [27]. The joint cyber attack detection and state estimation are presented in [28]. Furthermore, the distributed cyber attack detection and state estimation for a cyber-physical system under physical and cyber attacks are presented in [29]. After solving several LMIs and convex optimization, the joint estimation is performed, which incurs significant computational complexity.

Furthermore, the resource constraint based optimal state estimation algorithm for cyber-physical system is presented in [30]. Besides, the mixed integer linear programming based cyber

attack protection scheme for power system is developed in [31]. The computational complexity is very high, and it requires significant amount of time as it is a bilevel optimization problem. In order to guarantee the cyber and operational security, a command authentication approach is proposed to detect intrusion [32], [33]. In addition, the mean squared error (MSE) based smart grid state estimation algorithm is presented in [34]. Due to the suboptimal nature of the computed gain, it cannot accurately estimate the grid states under cyber attacks [34].

Distributed state estimations face real environments, where cyber attacks and noisy measurements are present. Differentiated from prior literature, this article is the first of its kind to solve distributed state estimation problem for smart grid under cyber attacks using the optimal filter theory and Bayesian learning process.

## B. Main Contributions

The main contributions of this article are as follows.

- 1) The power distribution system incorporating multiple synchronous generators is modeled as a state-space framework. The sensors are used to obtain measurement, which is corrupted by noise and cyber attacks.
- 2) Based on the optimal filter and graph theory, the distributed smart grid state estimation algorithm is proposed. Specifically, the local gain of the distributed scheme is obtained using the optimal filter theory, whereas the neighboring gain is determined through convex optimization process and graph theory. For local gain computation, the attack parameter is obtained using the Bayesian learning process.
- 3) The convergence condition of the proposed approach is derived.
- 4) Numerical simulations show that the proposed method can properly tolerate the cyber attack and noises leads to an accurate estimation result. The research issue and potential solution identified in this article can open up several avenues for future research in this area.

*Organizations:* The organization of this article is as follows: In Section II, power system model is described, which follows measurement and cyber attack frameworks in Section III. The proposed algorithm is derived in Section IV. Analysis of the computer simulations is given in Section V and Section VI concludes this article.

*Notations:* This article employs some standard notations. Bold face upper and lower case letters are used to represent matrices and vectors, respectively. Superscripts  $\mathbf{x}'$  denotes the transpose of  $\mathbf{x}$ ,  $E(\cdot)$  denotes the expectation operator,  $\text{diag}(\mathbf{x})$  denotes the diagonal matrix,  $\lambda(\mathbf{X})$  is the spectral radius of  $\mathbf{X}$ , and  $\mathbf{I}$  denotes the identity matrix with appropriate dimension.

## II. STATE-SPACE REPRESENTATION OF POWER NETWORK INCORPORATING SYNCHRONOUS GENERATORS

The smart grid is the backbone of a nation economy and is crucial to the homeland security. Generally, there are many synchronous generators and loads are connected to the distribution power networks. To illustrate, Fig. 2 shows the typical synchronous generators and loads that are connected to the 8-bus distribution lines [35]–[37]. Basically, the  $n$ th-synchronous

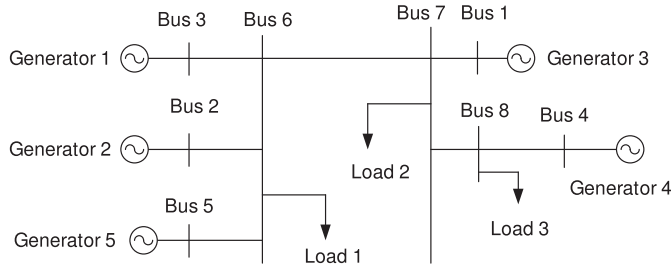


Fig. 2. Distributed power network incorporating synchronous generators.

generators can be represented by the following third-order differential equations as follows [36]–[38]:

$$\Delta \dot{\delta}_n = \Delta \omega_n \quad (1)$$

$$\Delta \dot{\omega}_n = -\frac{D_n}{H_n} \Delta \omega_n - \frac{\Delta P_{en}}{H_n} \quad (2)$$

$$\Delta \dot{E}'_{qn} = -\frac{\Delta E'_{qn}}{T'_{don}} + \frac{\Delta E_{fn}}{T'_{don}} + \frac{X_{dn}}{T'_{don}} \Delta I_{dn} - \frac{X'_{dn}}{T'_{don}} \Delta I_{dn} \quad (3)$$

here,  $\delta_n$  is the rotor angle,  $\omega_n$  is the rotor speed,  $H_n$  is the inertia constant,  $D_n$  is the damping constant,  $P_e$  is the active power delivered at the terminal,  $E'_{qn}$  is the quadrature-axis transient voltage,  $E_{fn}$  is the exciter output voltage,  $T'_{don}$  is the direct-axis open-circuit transient time constant,  $X_{dn}$  is the direct-axis synchronous reactance,  $X'_{dn}$  is the direct-axis transient reactance, and  $I_{dn}$  is the direct-axis current [35].

Generally, an automatic voltage regulator (AVR) is used to control the excitation current that leads to control the terminal voltage [36], [39]. A second-order transfer function is used to represent the AVR as follows [36]:

$$\Delta E_{fn} = b_{0n} z_{1n} + b_{1n} z_{2n} \quad (4)$$

$$z_{1n} = z_{2n} \quad (5)$$

$$\dot{z}_{2n} = -c_{1n} z_{2n} - c_{0n} z_{1n} + \Delta v_n \quad (6)$$

here,  $z_{1n}$  and  $z_{2n}$  are the AVR internal states,  $b_{0n}$  and  $b_{1n}$  are transfer function coefficients of the AVR,  $c_{0n}$  and  $c_{1n}$  are the transfer function coefficients of the excitation system and  $\Delta v_n$  is the control input signal.

Considering  $N$  generators in the power network, the  $d$ -axis current  $I_{di}$  and electrical power  $P_{ei}$  are represented as [39]

$$I_{dn} = \sum_{m=1}^N \Delta E'_{qn} [B_{nm} \cos(\delta_n - \delta_m) - G_{nm} \sin(\delta_n - \delta_m)]. \quad (7)$$

$$P_{en} = \Delta E'_{qn} \sum_{m=1}^N [B_{nm} \sin(\delta_n - \delta_m) + G_{nm} \cos(\delta_n - \delta_m)] \Delta E'_{qm} \quad (8)$$

here,  $n, m \in \{1, \dots, N\}$ ,  $G_{nm}$  and  $B_{nm}$  are the real and imaginary part of the admittance  $\mathbf{Y} \in \mathbb{R}^{N \times N}$ , which is described in the Appendix A.

After linearizing (7) and (8),  $\Delta P_{en}$  and  $\Delta I_{dn}$  are written as follows [36], [40], [13]:

$$\Delta P_{en} = \left[ \frac{\partial P_{en}}{\partial \delta} \frac{\partial P_{en}}{\partial E'_{qn}} \right] [\Delta \delta \Delta E'_{qn}]' \quad (9)$$

$$\Delta I_{dn} = \left[ \frac{\partial I_{dn}}{\partial \delta} \frac{\partial I_{dn}}{\partial E'_{qn}} \right] [\Delta \delta \Delta E'_{qn}]' \quad (10)$$

here,  $\Delta E'_{qn}$  and  $\Delta \delta$  are the transient voltage deviations and rotor angle deviations. By combining (1)–(6) and (9)–(10), it can be written as follows:

$$\dot{\mathbf{s}}_n = \mathbf{A}_n \mathbf{s}_n + \mathbf{B}_n u_n + \sum_{m \in N_n} \mathbf{A}_{nm} \mathbf{s}_m \quad (11)$$

here, the generator state  $\mathbf{s}_n = [\Delta \delta_n \Delta \omega_n \Delta E'_{qn} z_{2n} z_{1n}]'$ , the control input signal  $u_n = \Delta v_n$ ,  $N_n$  indicates a set of connected generators, the system matrices  $\mathbf{A}_n \in \mathbb{R}^{5 \times 5}$ ,  $\mathbf{B}_n \in \mathbb{R}^{5 \times 1}$  and  $\mathbf{A}_{nm} \in \mathbb{R}^{5 \times 5}$  are

$$\mathbf{A}_n =$$

$$\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ -\frac{1}{H_n} \frac{\partial P_{en}}{\partial \delta_n} & -\frac{D_n}{H_n} & -\frac{1}{H_n} \frac{\partial P_{en}}{\partial E'_{qn}} & 0 & 0 \\ X_n \frac{\partial I_{dn}}{\partial \delta_n} & 0 & -\frac{1}{T'_{don}} + X_n \frac{\partial I_{dn}}{\partial E'_{qn}} & \frac{b_{1n}}{T'_{don}} & \frac{b_{0n}}{T'_{don}} \\ 0 & 0 & 0 & -c_{1n} & -c_{0n} \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix}$$

$$\mathbf{A}_{nm} = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ -\frac{1}{H_n} \frac{\partial P_{en}}{\partial \delta_m} & 0 & -\frac{1}{H_n} \frac{\partial P_{en}}{\partial E'_{qm}} & 0 & 0 \\ X_n \frac{\partial I_{dn}}{\partial \delta_m} & 0 & -\frac{1}{T'_{don}} + X_n \frac{\partial I_{dn}}{\partial E'_{qm}} & \frac{b_{1n}}{T'_{don}} & \frac{b_{0n}}{T'_{don}} \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

$$\mathbf{B}_n = [0 \ 0 \ 0 \ 1 \ 0]' [36] \text{ and } X_n = \frac{X_{dn} - X'_{dn}}{T'_{don}}.$$

The aforementioned system can be written in continuous-time form

$$\dot{\mathbf{s}} = \mathbf{A}^c \mathbf{s} + \mathbf{B}^c \mathbf{u} + \mathbf{w} \quad (12)$$

here,  $\mathbf{s} \in \mathbb{R}^{5N \times 1}$ ,  $\mathbf{u} \in \mathbb{R}^{N \times 1}$ ,  $\mathbf{w} \in \mathbb{R}^{5N \times 1}$  is the process noise that can follow the Gaussian distribution incorporating zero mean and  $\mathbf{Q}$  covariance, i.e.,  $\mathbf{N}(\mathbf{0}, \mathbf{Q})$ ,  $\mathbf{A}^c \in \mathbb{R}^{5N \times 5N}$  and  $\mathbf{B}^c \in \mathbb{R}^{5N \times N}$  are given by:

$$\mathbf{A}^c = \begin{bmatrix} \mathbf{A}_1 & \mathbf{A}_{12} & \cdots & \mathbf{A}_{1N} \\ \mathbf{A}_{21} & \mathbf{A}_2 & \cdots & \mathbf{A}_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ \mathbf{A}_{N1} & \mathbf{A}_{N2} & \cdots & \mathbf{A}_N \end{bmatrix}$$

$$\mathbf{B}^c = \text{diag}(\mathbf{B}_1 \cdots \mathbf{B}_N).$$

Now, it can be written as a discrete-time form as follows:

$$\mathbf{s}(t+1) = \mathbf{A} \mathbf{s}(t) + \mathbf{B} \mathbf{u}(t) + \mathbf{w}(t) \quad (13)$$

where  $\mathbf{A} = \mathbf{I} + \mathbf{A}^c \Delta t$ ,  $\Delta t$  is the sampling time, and  $\mathbf{B} = \mathbf{B}^c \Delta t$ .

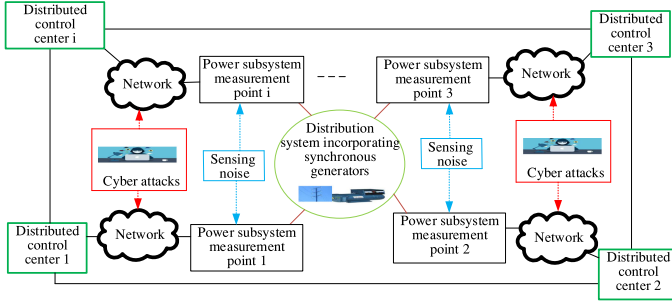


Fig. 3. Interconnected distribution power subsystems incorporating synchronous generators.

### III. MEASUREMENT AND CYBER ATTACK FRAMEWORKS

The distributed control centers are interconnected through communication links as shown in Fig. 3. In this figure, there are  $i$ th distribution subsystems, which are connected to the neighbors units. These control centers can share information with their neighbors in a distributed way. The sensors are installed into subsystem units to obtain distributed measurements. The sensing information is telemetered to the control centres to estimate system states such as rotor angle. The measurements are obtained as follows:

$$\mathbf{z}_i(t) = \mathbf{C}_i \mathbf{s}(t) + \mathbf{v}_i(t) \quad (14)$$

here,  $\mathbf{z}_i(t) \in \mathbb{R}_i^p$  is the measurement, and  $\mathbf{v}_i \sim N(\mathbf{0}, \mathbf{R}_i)$  is the measurement noise, and  $\mathbf{C}_i$  is the sensing matrix.

When sensing information is transmitted to the control center, the attacker can hack communication network and manipulate measurements. There are different kinds of attacks such as FDIA and replay attack [41]. For FDIA, an attacker is added intended information to the actual measurement over time, then report it to the control center for misleading. In latter case, the adversary records the normal measurements over time [9]. During attack, the actual measurements are replaced to be recoded one and thereby moving the system into an incorrect state [42]. The detailed attack templates are described in [43]. Mathematically, when there is attack then the system measurement can be written as follows:

$$\mathbf{z}_i^a(t) = \mathbf{C}_i \mathbf{s}(t) + \mathbf{v}_i(t) + \mathbf{a}_i(t) \quad (15)$$

where,  $\mathbf{a}_i(t)$  is the cyber attack. We consider that the attack vector  $\mathbf{a}_i$  is a Gaussian distribution with mean  $\mu_i$  and covariance  $\bar{\mathbf{R}}_i^a$ , i.e.,  $\mathbf{a}_i \sim N(\mu_i, \bar{\mathbf{R}}_i^a)$  [25], [28]. It assumes that the attack sequence is uncorrelated to each measurement [44], [43]. Let define the system model parameters  $\varphi_i = (\mu_i, \hat{\mathbf{R}}_i)$ , where  $\hat{\mathbf{R}}_i = \mathbf{R}_i + \mathbf{R}_i^a$  is the combined covariance of noise and cyber attack. Based on this noisy and corrupted version of measurements, the proposed state estimation algorithm is developed in the following section.

### IV. PROPOSED DISTRIBUTED SMART GRID STATE ESTIMATION ALGORITHM AND CONVERGENCE CONDITION

Using MSE principle, the distributed dynamic state estimation algorithm is designed where local and neighboring gains are

obtained using optimal filter and graph theory. Afterward, the convergence condition of the developed approach is derived.

#### A. Proposed Algorithm

The proposed distributed state estimation algorithm is obtained using the optimal filter and Bayesian learning approaches [43]. Based on the interconnected structure in Fig. 3, the designed scheme is mathematically written as follows:

$$\begin{aligned} \hat{\mathbf{s}}_i(t+1) = & \mathbf{A} \hat{\mathbf{s}}_i(t) + \mathbf{B} \mathbf{u}(t) + \mathbf{G}_i(t) [\mathbf{z}_i^a(t) - \mathbf{C}_i \hat{\mathbf{s}}_i(t)] \\ & + \mathbf{L}_i(t) \sum_{j \in N_i} [\hat{\mathbf{s}}_j(t) - \hat{\mathbf{s}}_i(t)] \end{aligned} \quad (16)$$

here,  $\hat{\mathbf{s}}_i(t+1)$  is the posterior estimated system state,  $\hat{\mathbf{s}}_i(t)$  is the previous estimated state,  $\mathbf{G}_i(t)$  and  $\mathbf{L}_i(t)$  are the local and consensus gains that can minimize the residual error dynamic,  $\mathbf{z}_i^a(t) - \mathbf{C}_i \hat{\mathbf{s}}_i(t)$ , and neighboring estimation mismatch,  $\hat{\mathbf{s}}_j(t) - \hat{\mathbf{s}}_i(t)$ , over time. Basically, the last term of the distributed scheme (16) is used for neighboring connections in Fig. 3, while the third term is included for self-estimation unit. The following theorem is used to compute these gains for distributed smart grid state estimation.

*Theorem 1:* After defining the error,  $\boldsymbol{\eta}_i(t) = \mathbf{s}(t) - \hat{\mathbf{s}}_i(t)$ , between the true and estimated system states and using the optimal filter as well as graph theory, the designed gains are obtained as follows [43]:

$$\begin{aligned} \mathbf{G}_i(t) = & [\mathbf{A} \mathbf{P}_i(t) \mathbf{C}_i' + \mathbf{L}_i(t) \sum_{r \in N_i} \{\mathbf{P}^{ri}(t) - \mathbf{P}_i(t)\} \mathbf{C}_i'] \\ & \times [\mathbf{C}_i' \mathbf{P}_i(t) \mathbf{C}_i' + \hat{\mathbf{R}}_i]^{-1}. \end{aligned} \quad (17)$$

Using MSE principle, the estimation error covariance  $\mathbf{P}_i(t+1) = E[\boldsymbol{\eta}_i(t+1) \boldsymbol{\eta}_i'(t+1)]$  is determined by

$$\begin{aligned} \mathbf{P}_i(t+1) = & \mathbf{A} \mathbf{P}_i(t) \mathbf{A}' - \mathbf{A} \mathbf{P}_i(t) \mathbf{C}_i' [\mathbf{C}_i \mathbf{P}_i(t) \mathbf{C}_i' \\ & + \hat{\mathbf{R}}_i]^{-1} \mathbf{C}_i \mathbf{P}_i(t) \mathbf{A}' + \mathbf{Q} \end{aligned} \quad (18)$$

here,  $\mathbf{P}_i(t)$  is the prior estimation error covariance. Using the Bayesian learning formula, the covariance  $\hat{\mathbf{R}}_i$  is computed as follows [45]:

$$\begin{aligned} \hat{\mathbf{R}}_i = & (\alpha_i \bar{\mathbf{R}}_i + \rho_i [\text{diag}(\hat{\boldsymbol{\mu}}_i)]^2 - (\rho_i + 1) [\text{diag}(\hat{\boldsymbol{\mu}}_i)]^2 \\ & + [\text{diag}(\mathbf{z}_i^a - \mathbf{C}_i \hat{\mathbf{s}}_i)]^2) / (\alpha_i + 1) \end{aligned} \quad (19)$$

$$\hat{\boldsymbol{\mu}}_i = (\rho_i \bar{\boldsymbol{\mu}}_i + \mathbf{z}_i^a - \mathbf{C}_i \hat{\mathbf{s}}_i) / (\rho_i + 1) \quad (20)$$

where,  $\bar{\mathbf{R}}_i$  and  $\bar{\boldsymbol{\mu}}_i^a$  are the initial values,  $\alpha_i$  and  $\rho_i$  are the hyperparameters.

For mathematical simplicity, we assume that neighboring gain  $\mathbf{L}_i(t) = v \mathbf{I}$ , where  $v$  is the designed gain coefficient. Under a steady-state condition, it can be computed through the following convex optimization process [43]:

$$v = \underset{v}{\text{argmax}} \begin{bmatrix} -\mathbf{I} & \mathbf{\Gamma} \\ \mathbf{\Gamma}' & -\mathbf{I} \end{bmatrix} < \mathbf{0} \quad (21)$$

here,  $\mathbf{\Gamma} = \mathbf{I} \otimes \mathbf{A} - b \text{diag}\{\mathbf{G}_i \mathbf{C}_i\} - b \text{diag}\{\mathbf{L}_i\} (\mathbf{L}_p \otimes \mathbf{I})$  is the augmented state dynamic,  $\mathbf{L}_p$  is the Laplacian operator which is obtained through the graph theory after combining all error dynamics in a compact form as shown in (32),  $\otimes$  indicates the

Kronecker product. From (32), it can be seen that the overall system dynamic is written into two items: augmented state dynamic,  $\mathbf{\Gamma}$ , and augmented error,  $\mathbf{W} = (\mathbf{1} \otimes \mathbf{I})\mathbf{w} - b\text{diag}\{\mathbf{G}_i\}\mathbf{v}$ .

The Proof is derived in Appendix B.

### B. Convergence Condition

When the estimated grid state converges to the actual state over time then the convergence of the developed approach is guaranteed asymptotically [46]. In doing this, the designed local and neighbouring gains ( $\mathbf{G}$  and  $\mathbf{L}$ ) can play key rule to analyze the convergence of the algorithm. With these gains, the system is stable when the spectral radius of augmented state matrix is less than one, i. e.,  $\lambda(\mathbf{\Gamma}) < 1$ . In this case, the term  $E[\mathbf{W}(t)\mathbf{W}'(t)] = \tilde{\mathbf{Q}}_R(t)$  converges to

$$\tilde{\mathbf{Q}}_R(t) = \{\mathbf{I} \otimes \mathbf{G}_i(t)\}\text{diag}(\mathbf{R}_1, \dots, \mathbf{R}_i)\{\mathbf{I} \otimes \mathbf{G}_i(t)\}' + \mathbf{1}\mathbf{1}'\mathbf{Q}. \quad (22)$$

The entry of vector  $\mathbf{1}$  is one with appropriate dimension.

**Lemma 1:** If the eigenvalues of a matrix  $\mathbf{H} \in \mathbb{R}^{r \times r}$  are increasing order, then the following inequality holds [47, p. 235], [46]:

$$\min_{\mathbf{b} \neq \mathbf{0}} \left| \frac{\mathbf{b}'\mathbf{H}\mathbf{b}}{\mathbf{b}'\mathbf{b}} \right| \leq |\lambda_l(\mathbf{H})| \leq \max_{\mathbf{b} \neq \mathbf{0}} \left| \frac{\mathbf{b}'\mathbf{H}\mathbf{b}}{\mathbf{b}'\mathbf{b}} \right|. \quad (23)$$

Here,  $\lambda(\mathbf{H})$  denotes the eigenvalues of  $\mathbf{H}$ ,  $\lambda_l(\mathbf{H})$  is the spectral radius of  $\mathbf{H}$  with  $l = 1, 2, \dots, r$  and  $\mathbf{b} \neq \mathbf{0}$  is the nonzero vector. The smallest and largest eigenvalues of  $\mathbf{H}$  can be characterized as solution of the min-max problems involving the *Rayleigh quotient*  $\mathbf{b}'\mathbf{H}\mathbf{b}/\mathbf{b}'\mathbf{b}$ . The Rayleigh quotient is used in the min-max theorem to get the actual values of all eigenvalues.

In order to apply the Lemma 1, one can ascending order the eigenvalues of  $\mathbf{\Gamma}(t)$  as follows [46]:

$$\lambda_{\min} = \lambda_1(\mathbf{\Gamma}(t)) \leq \lambda_2(\mathbf{\Gamma}(t)), \dots, \leq \lambda_l(\mathbf{\Gamma}(t)) = \lambda_{\max}. \quad (24)$$

Therefore, the maximum absolute eigenvalue of  $\mathbf{\Gamma}(t)$  is

$$\rho(E[\mathbf{\Gamma}(t)]) \leq \max_l |\lambda_l[\mathbf{I} \otimes \mathbf{A} - b\text{diag}\{\mathbf{G}_i(t)\mathbf{C}_i\} - b\text{diag}\{\mathbf{L}_i(t)\}(\mathbf{L}_p \otimes \mathbf{I})]|. \quad (25)$$

When the estimated states converge to the true states, the state error covariance  $\mathbf{P}_i(t)$  converges to  $\mathbf{P}_i$  [48].

## V. NUMERICAL SIMULATION RESULTS AND ANALYSIS

To estimate the system state, the proposed algorithm is applied to the distribution power network as shown in Fig. 2. For simplicity, we assume that there are  $i = 4$  interconnected distributed controllers as shown in Fig. 3. In this network, the step by step procedure of the developed algorithm is demonstrated in Fig. 4. First of all, the power system and measurement are modeled by (13) and (15), respectively. It can be seen that the sensing measurement is corrupted by noise and cyber attacks. For instance, the attacker is injected the malicious false data into the measurements based on the attack templates presented in Section III. Afterward, the distributed estimation structure (16) is formulated considering unknown filtering parameters such as local and consensus gains, which are obtained by (17) and (21).

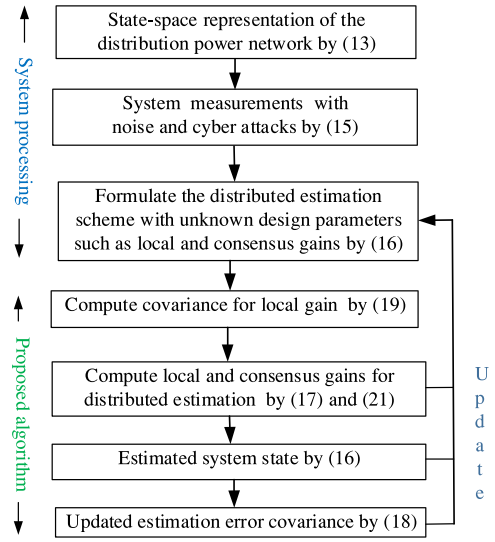


Fig. 4. Step-by-step procedure of the proposed algorithm.

TABLE I  
CONSIDERED GENERATOR PARAMETERS

| Parameters | G1     | G2     | G3     | G4     | G5     |
|------------|--------|--------|--------|--------|--------|
| $D_i$      | 3.04   | 3.67   | 3.35   | 3.98   | 3.4    |
| $H_i$      | 4.5    | 4.65   | 4.43   | 3.94   | 4.9    |
| $X_{di}'$  | 0.0329 | 0.0329 | 0.3290 | 0.0239 | 0.2390 |
| $X_{di}$   | 0.1016 | 0.1016 | 1.016  | 0.1016 | 1.016  |
| $T_{doi}'$ | 5.57   | 5.57   | 5.57   | 5.57   | 5.57   |
| $b_{oi}$   | 656    | 656    | 656    | 656    | 656    |
| $b_{1i}$   | 1232   | 1232   | 1232   | 1232   | 1232   |
| $c_{oi}$   | 3.23   | 3.23   | 3.23   | 3.23   | 3.23   |
| $c_{1i}$   | 32.3   | 32.3   | 32.3   | 32.3   | 32.3   |
| $V$        | 1.03   | 1.03   | 1.023  | 1.03   | 1.023  |
| $\theta$   | 0      | 0.1041 | 0.0933 | 0.0351 | 0.0607 |
| $Q$        | 2.9141 | 1.3821 | 0.4187 | 2.1802 | 0.3469 |
| $P$        | 3.1521 | 4.1016 | 0.4608 | 4.0578 | 0.1547 |

TABLE II  
CONSIDERED LINE PARAMETERS

| Node i | Node j | $R_{ij}$ | $X_{ij}$ | $B_{ij}$ |
|--------|--------|----------|----------|----------|
| 1      | 7      | 0.00335  | 0.01057  | 0.01436  |
| 2      | 6      | 0.00313  | 0.00368  | 0.00304  |
| 3      | 6      | 0.03004  | 0.05242  | 0.06305  |
| 4      | 8      | 0.00514  | 0.01074  | 0.01654  |
| 5      | 6      | 0.00701  | 0.02231  | 0.02632  |
| 6      | 7      | 0.04022  | 0.12685  | 0.15848  |
| 7      | 8      | 0.01714  | 0.04143  | 0.05013  |

For local gain, the covariance is computed by (18). Technically, the local and consensus gains can minimize the residual error and neighboring estimation mismatch over time leads to an accurate estimation result. Finally, the estimated system state is updated by (16), and it error covariance is determined by (18).

The simulation is conducted through MATLAB and YALMIP environments [49]. The simulation parameters are described in Tables I and II [36], [39]. Basically, the process noise covariance is followed by Gaussian distribution with covariance is  $10^{-8}\mathbf{I}$ . In addition, the measurement noise covariance for four estimators are followed by Gaussian distributions with respective covariances are  $1 * 10^{-7}\mathbf{I}$ ,  $2 * 10^{-7}\mathbf{I}$ ,  $3 * 10^{-7}\mathbf{I}$ , and  $4 * 10^{-7}\mathbf{I}$ .

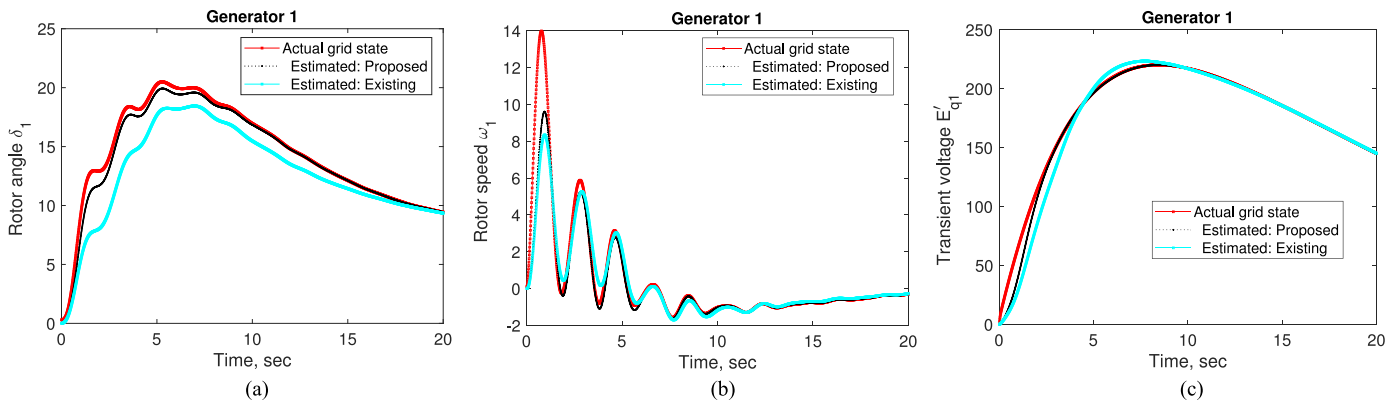


Fig. 5. Generator 1. (a) Actual rotor angle and its estimated state with FDIA. (b) Actual rotor speed and its estimated state with FDIA. (c) Actual transient voltage and its estimated state with FDIA.

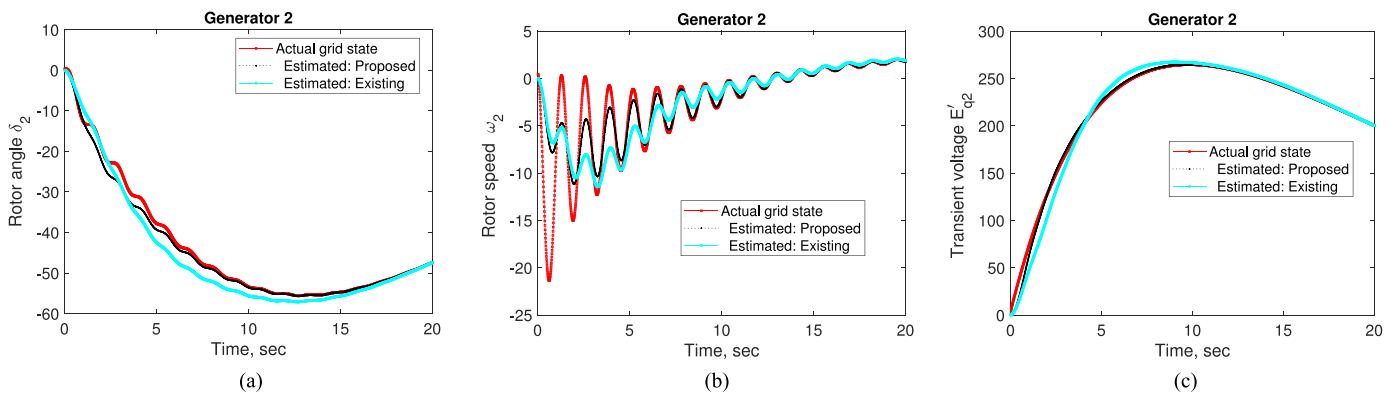


Fig. 6. Generator 2. (a) Actual rotor angle and its estimated state with FDIA. (b) Actual rotor speed and its estimated state with FDIA. (c) Actual transient voltage and its estimated state with FDIA.

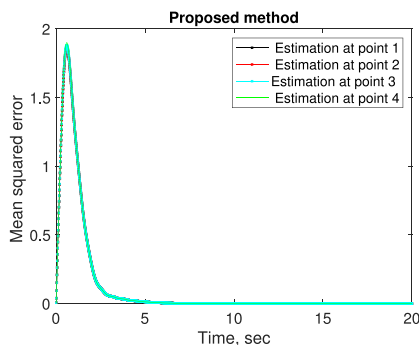


Fig. 7. MSE between true and estimated states with FDIA.

Moreover, the sampling period is 0.01 s, and there are five synchronous generators connected to the 8-bus distribution network as shown in Fig. 2. The simulation is conducted considering FDIA and replay attacks.

First of all, it assumes that the attacker is added FDIA into measurement during 0.05–0.25 s. In this case, the simulation results are illustrated in Figs. 5–8. Basically, Fig. 5(a)–(c) shows the generator 1 true states and their estimation results. The proposed algorithm can properly estimate the system states

within 10 s, whereas the existing method requires 20 s [34]. Similarly, high estimation accuracy is illustrated in Fig. 6(a)–(c) for generator 2. This is due to the fact that the proposed algorithm can find the optimal gains, so the estimated states converge to the actual states within a short period of time. On the other hand, the existing method cannot determine the desired gains as they are suboptimal in nature. In this case [34], we assume the neighboring gain  $\mathbf{L}$  is  $0.026 \times \mathbf{I}$ . The MSE between true and estimated system state is illustrated in Fig. 7. It can be seen that all estimators reach consensus on estimation. In summary, it can be observed that the proposed technique requires almost half time to estimate the system states compared with the existing method.

When there is replay attack during 0.05–0.25 s, the simulation results are presented in Figs. 8–10. It can be seen that the proposed algorithm requires more time to estimate the system states with this attack. Most importantly, the algorithm provides consistent results in all cases. The proposed algorithm can properly estimate the system states within 11 s while the existing method requires 20 s [34]. Fig. 8(a)–(c) shows the generator 4 true states and their estimation results. The developed scheme can accurately estimate the grid states within 11 s, whereas the existing method requires 20 s [34]. Similarly, high estimation accuracy is described in Fig. 9(a)–(c) for generator 5. This is

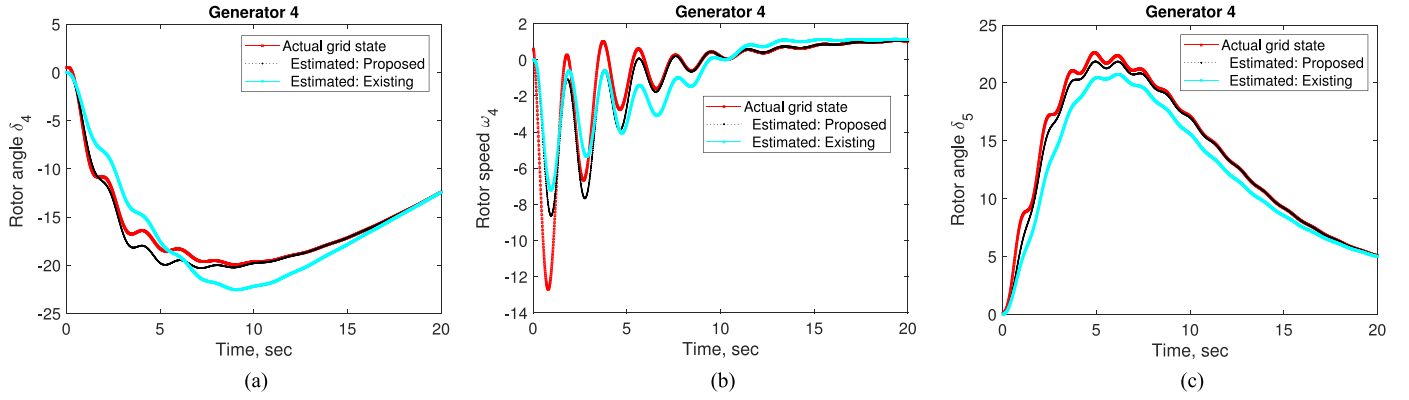


Fig. 8. Generator 4. (a) Actual rotor angle and its estimated state with replay attack. (b) Actual rotor speed and its estimated state with replay attack. (c) Actual transient voltage and its estimated state with replay attack.

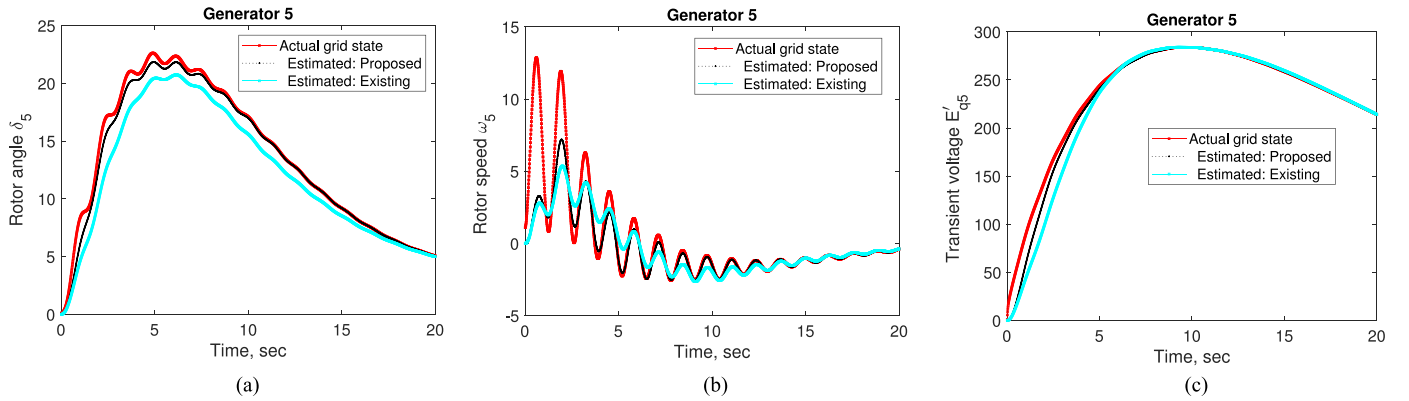


Fig. 9. Generator 5. (a) Actual rotor angle and its estimated state with replay attack. (b) Actual rotor speed and its estimated state with replay attack. (c) Actual transient voltage and its estimated state with replay attack.

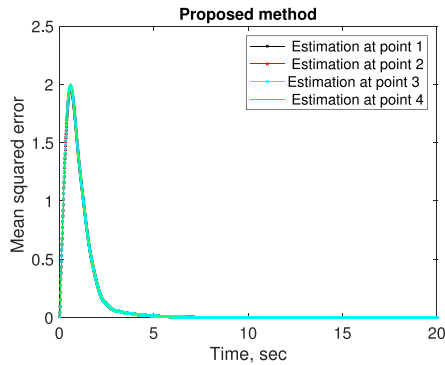


Fig. 10. MSE between true and estimated states with replay attack.

due to the fact that the designed scheme can find the optimal gains, so the estimated grid states converge to the true states within 11 s. However, the comparative scheme cannot compute the optimal gains as they are suboptimal in nature [34]. The MSE under this attack is illustrated in Fig. 10. It can be observed that all estimators reach consensus on estimation. Overall, it can be seen that the proposed method requires almost half time to estimate the grid states compared with the traditional approach.

## VI. CONCLUSION

State estimation is the key task for power system operation and maintain stability as well as observability. However, the smart grid infrastructure is prone to cyber threats. In order to protect power network from cyber attacks, this article proposes a distributed state estimation algorithm. Specifically, we have made three main contributions to enhance the cyber security and resiliency of smart grids. First, the 8-bus distribution grid incorporating synchronous generators is modeled as a state-space framework, where measurement is obtained by a set of sensors. The measurement data are manipulated by cyber attacks such as FDIA. Second, we proposed an attack-resilient distributed state estimation algorithm based on the optimal filter and graph theory. Finally, the convergence condition of the proposed approach is derived. Extensive, simulation results show that the proposed algorithm can able to estimate system state within a short time. We will try to develop a data-driven distributed state estimation algorithm considering cyber attacks.

## APPENDIX A

The network admittance is written by [39], [37]

$$\mathbf{Y} = \mathbf{Y}_{rr} - \mathbf{Y}_{re} \mathbf{Y}_{ee}^{-1} \mathbf{Y}_{re}' \quad (26)$$

here,  $\mathbf{Y}_{rr} = \text{diag}[Y_{17} + jB_{17}, Y_{26} + jB_{26}, Y_{36} + jB_{36}, Y_{46} + jB_{46}, Y_{56} + jB_{56}]$ , where the mutual admittance is computed as follows as an example:  $Y_{17} = 1/(R_{17} + jX_{17})$ ,  $R_{17}$  is the resistance between node 1 and 7,  $X_{17}$  and  $B_{17}$  are its reactance and susceptance, respectively. The term  $\mathbf{Y}_{re}$  is written as

$$\mathbf{Y}_{re} = \begin{bmatrix} 0 & -Y_{17} & 0 \\ -Y_{26} & 0 & 0 \\ -Y_{36} & 0 & 0 \\ 0 & 0 & -Y_{48} \\ -Y_{56} & 0 & 0 \end{bmatrix}. \quad (27)$$

The term  $\mathbf{Y}_{ee}$  is written as

$$\mathbf{Y}_{ee} = \begin{bmatrix} Y_{66} & -Y_{67} & 0 \\ -Y_{67} & Y_{77} & -Y_{78} \\ 0 & -Y_{78} & Y_{88} \end{bmatrix} \quad (28)$$

where  $Y_{ii}$  is the self-admittance.

#### APPENDIX B (PROOF OF THEOREM 1)

The estimation error  $\boldsymbol{\eta}_i(t+1) = \mathbf{s}(t+1) - \hat{\mathbf{s}}_i(t+1)$  without first considering cyber attacks can be expressed as [43]

$$\begin{aligned} \boldsymbol{\eta}_i(t+1) &= [\mathbf{A} - \mathbf{G}_i(t)\mathbf{C}_i]\boldsymbol{\eta}_i(t) + \mathbf{L}_i(t) \sum_{r \in N_i} [\boldsymbol{\eta}_r(t) \\ &\quad - \boldsymbol{\eta}_i(t)] + \mathbf{w}(t) - \mathbf{G}_i(t)\mathbf{v}_i(t). \end{aligned} \quad (29)$$

The estimation error covariance  $\mathbf{P}_i(t+1) = E[\boldsymbol{\eta}_i(t+1)\boldsymbol{\eta}_i'(t+1)]$  is expressed as follows:

$$\begin{aligned} \mathbf{P}_i(t+1) &= [\mathbf{A} - \mathbf{G}_i(t)\mathbf{C}_i]\mathbf{P}_i(t)[\mathbf{A} - \mathbf{G}_i(t)\mathbf{C}_i]' \\ &\quad + [\mathbf{A} - \mathbf{G}_i(t)\mathbf{C}_i] \sum_{s \in N_i} [\mathbf{P}_{is}(t) - \mathbf{P}_i(t)]\mathbf{L}_i'(t) + \mathbf{L}_i(t) \sum_{r \in N_i} \\ &\quad \times [\mathbf{P}_{ri}(t) - \mathbf{P}_i(t)][\mathbf{A} - \mathbf{G}_i(t)\mathbf{C}_i]' + \mathbf{L}_i(t) \sum_{r, s \in N_i} \\ &\quad \times [\mathbf{P}_{rs}(t) - \mathbf{P}_{ri}(t) - \mathbf{P}_{is}(t) + \mathbf{P}_i(t)]\mathbf{L}_i'(t) \\ &\quad + \mathbf{G}_i(t)\mathbf{R}_i\mathbf{G}_i'(t) + \mathbf{Q}. \end{aligned} \quad (30)$$

After partial derivative of  $\mathbf{P}_i(t+1)$  with respect to local gain  $\mathbf{G}_i(t)$ , and setting its derivative equal to zero, i.e.,  $\frac{\partial \{tr[\mathbf{P}_i(t+1)]\}}{\partial \mathbf{G}_i(t)} = 0$ , the optimal local gain is expressed as follows [43]:

$$\begin{aligned} \mathbf{G}_i(t) &= [\mathbf{A}\mathbf{P}_i(t)\mathbf{C}_i' + \mathbf{L}_i(t) \sum_{r \in N_i} \{\mathbf{P}_{ri}(t) - \mathbf{P}_i(t)\}\mathbf{C}_i'] \\ &\quad \times [\mathbf{C}_i\mathbf{P}_i(t)\mathbf{C}_i' + \mathbf{R}_i]^{-1}. \end{aligned} \quad (31)$$

Using the heuristic approach,  $\mathbf{R}_i$  in (31) is replaced by  $\hat{\mathbf{R}}_i$  and  $\mathbf{R}_i^a = \hat{\mathbf{R}}_i$ , which leads to the local gain expression in (17).

Using (29), the overall error dynamic  $\boldsymbol{\eta}(t) = [\boldsymbol{\eta}_1(t) \dots \boldsymbol{\eta}_n(t)]'$  can be written as follows:

$$\begin{aligned} \boldsymbol{\eta}(t+1) &= [\mathbf{I} \otimes \mathbf{A} - \text{bdiag}\{\mathbf{G}_i(t)\mathbf{C}_i\} - \text{bdiag}\{\mathbf{L}_i(t)\} \\ &\quad \times (\mathbf{L}_p \otimes \mathbf{I})]\boldsymbol{\eta}(t) + (\mathbf{1} \otimes \mathbf{I})\mathbf{w} - \text{bdiag}\{\mathbf{G}_i(t)\}\mathbf{v}(t) \\ &= \boldsymbol{\Gamma}(t)\boldsymbol{\eta}(t) + \mathbf{W}(t) \end{aligned} \quad (32)$$

where,  $\mathbf{L}_p$  is the Laplacian operator [13],  $\boldsymbol{\Gamma}(t) = \mathbf{I} \otimes \mathbf{A} - \text{bdiag}\{\mathbf{G}_i(t)\mathbf{C}_i\} - \text{bdiag}\{\mathbf{L}_i(t)\}(\mathbf{L}_p \otimes \mathbf{I})$  and  $\mathbf{W}(t) = (\mathbf{1} \otimes \mathbf{I})\mathbf{w} - \text{bdiag}\{\mathbf{G}_i(t)\}\mathbf{v}(t)$ . The overall error covariance  $\mathbf{P}(t+1) = E[\boldsymbol{\eta}(t+1)\boldsymbol{\eta}(t+1)'(t+1)]$  can be expressed as follows:

$$\mathbf{P}(t+1) = E[\boldsymbol{\Gamma}(t)\mathbf{P}(t)\boldsymbol{\Gamma}'(t)] + E[\mathbf{V}(t)\mathbf{V}'(t)]. \quad (33)$$

According to the matrix property of the Kronecker product  $\text{vec}(\mathbf{X}\mathbf{Y}\mathbf{X}) = (\mathbf{X}' \otimes \mathbf{X})\text{vec}(\mathbf{Y})$ , (33) can be written as a vector form

$$\begin{aligned} \text{vec}[\mathbf{P}(t+1)] &= E[\boldsymbol{\Gamma}(t) \otimes \boldsymbol{\Gamma}'(t)]\text{vec}[E\{\mathbf{P}(k)\}] \\ &\quad + \text{vec}[E\{\mathbf{V}(k)\mathbf{V}'(k)\}]. \end{aligned} \quad (34)$$

It can be seen that  $\mathbf{G}_i(t)$  and  $\mathbf{L}_i(t)$  are unknown but function of each other that makes difficult to get any one of their estimated values. To overcome this, setting  $\mathbf{L}_i(t) = \mathbf{0}$ , the local gain  $\mathbf{G}_i(t)$  and error covariance (18) is derived. For mathematical simplicity, we assume that neighboring gain  $\mathbf{L}_i = v\mathbf{I}$ , where  $v$  is the designed gain coefficient. This is because, the consensus gain is generally small. With the given steady-state local gain and corresponding error covariance, the system is stable when the spectral radius of  $\lambda[\boldsymbol{\Gamma}]$  is less than one [43], i.e.,

$$v = \underset{v}{\text{argmax}} \lambda(\boldsymbol{\Gamma}) < 1 \Rightarrow \boldsymbol{\Gamma}\boldsymbol{\Gamma}' < \mathbf{I} \Rightarrow \begin{bmatrix} -\mathbf{I} & \boldsymbol{\Gamma} \\ \boldsymbol{\Gamma}' & -\mathbf{I} \end{bmatrix} < \mathbf{0}. \quad (35)$$

The aforementioned LMI is obtained using the Schur complement.

Inspired by the Bayesian learning approach, the covariance  $\hat{\mathbf{R}}_i$  is computed for local gain computation. Let define the unknown model parameter  $\boldsymbol{\varphi}_i = (\boldsymbol{\mu}_i, \hat{\mathbf{R}}_i)$ . For simplicity, we are omitted time index. Using the variational Bayesian learning method [45], the likelihood function is written as follows:

$$f(\mathbf{z}_i^a) = \int f(\mathbf{z}_i^a|\mathbf{s}, \boldsymbol{\varphi}_i)f(\mathbf{s})f(\boldsymbol{\varphi}_i)d\mathbf{s}d\boldsymbol{\varphi}_i. \quad (36)$$

Based on the state-space model (13), and measurement (15),  $f(\mathbf{s}) = N(\mathbf{A}\mathbf{s} + \mathbf{B}\mathbf{u}, \mathbf{Q})$  and  $f(\mathbf{z}_i^a|\mathbf{x}, \boldsymbol{\varphi}_i) = N(\mathbf{C}_i\mathbf{s} + \boldsymbol{\mu}_i, \hat{\mathbf{R}}_i)$ . It assumes that the model parameters  $(\boldsymbol{\mu}_i, \hat{\mathbf{R}}_i)$  have conjugate priors, so the posterior model parameter  $\hat{\boldsymbol{\varphi}}_i$  is almost same as the prior probability distribution. Mathematically, it can be written as follows:

$$\boldsymbol{\mu}_i | (\boldsymbol{\mu}_i, \rho_i, \hat{\mathbf{R}}_i) \sim N(\boldsymbol{\mu}_i, \hat{\mathbf{R}}_i/\rho_i) \quad (37)$$

$$\mathbf{R}_i^{-1} | (\alpha_i, \zeta_i) \sim W(\alpha_i, \zeta_i^{-1}/\alpha_i) \quad (38)$$

here,  $W(\bullet)$  is the Wishart distribution,  $\boldsymbol{\mu}_i, \rho_i, \alpha_i$ , and  $\zeta_i$  are the hyperparameters. It can be seen that all distribution functions are belong to the exponential family, and using [45], [50], we can obtain the unknown estimation parameter  $\hat{\boldsymbol{\mu}}_i = E(\boldsymbol{\mu}_i)$  and  $\hat{\mathbf{R}}_i = E(\mathbf{R}_i)$  as shown in (19) and (20). This completes the derivation.

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