

Online Learning for Spectrum Sensing and Reconfigurable Antenna Control

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Abstract—Efficient dynamic spectrum access (DSA) policies rely on accurate spectrum sensing information to exploit spectrum white space optimally. Obtaining accurate channel state information (CSI) from local spectrum measurements is made difficult by wireless signal fading and the presence of thermal noise which distorts measured signals and leads to uncertainty regarding the occupancy of spectrum resources. Electrically reconfigurable antenna systems (ERAS) offer the system designer an additional degree of freedom to exploit pattern and polarization diversity to improve the accuracy of local spectrum sensing decisions. We propose a learning technique to exploit pattern and polarization diversity offered by ERAS to improve spectrum sensing accuracy. While the proposed approach is designed to work within the unique design constraints of reconfigurable antennas, the approach is not antenna specific and will work with a wide variety of reconfigurable antenna designs. Validation of system performance is provided from measurements taken using the wireless open access research platform (WARP) software defined radio (SDR) platform in an indoor office environment.

I. INTRODUCTION

Research in spectrum sensing for dynamic spectrum access (DSA) applications has been broad and varied e.g., [1]. Different techniques have been proposed to limit the occurrence of spectrum sensing errors that can result from wireless signal fading and thermal noise which distort measured signals and leads to uncertainty about the state of the wireless spectrum resource. Since wireless signal fading is environment specific, the performance of spectrum sensors is link dependent. This link dependence can result in sub-optimal performance when sensors are located in the presence of a deep fade or shadowed.

To address this problem, it has previously been proposed to use antenna array techniques like synthetic apertures and diversity algorithms to improve spectrum sensing reliability [2]–[8]. These techniques are capable of improving system performance and reliability by exploiting the spatial diversity available to multi-antenna arrays.

Electrically reconfigurable antenna systems (ERAS) are an emerging class of antenna systems that are capable of altering antenna pattern and polarization through the use of a control network of electric switches (e.g., solid state switches, or micro-electromechanical systems (MEMS)), or tunable resonant elements (e.g., varactors) embedded in the antenna structure [9]–[12]. These control networks allow for rapid reconfiguration of antennas making it possible to adapt to changing environmental conditions on a transmission by transmission basis. Reconfigurable antennas offer the potential to extend the capabilities of both single antenna systems and

Unlike conventional diversity techniques, the challenge of reconfigurable antenna configuration control is uniquely constrained. ERAS can exist in only one antenna configuration at any given time, making it impossible to have complete channel state information (CSI) over all antennas states at any given time. As a result, ERAS configuration control algorithms must balance the benefit of exploring the antenna state space with the cost of using the system in suboptimal antenna configurations. This constraint has been identified in previous applications of ERAS for point-to-point multiple-input-multiple-output (MIMO) wireless links [13]–[15].

The proposed control algorithm provides a method for the optimized selection of ERAS states for the purposes of minimizing spectrum sensing errors. This algorithm balances the trade-off between exploration of the performance of each ERAS configuration, and the exploitation of the ERAS configuration which is best suited for meeting the design objective. Experimental validation of the proposed technique was carried out using empirical wireless channel data from an indoor office environment.

II. SYSTEM MODEL

Typical spectrum sensing decisions are made from local measurement of the wireless spectrum resource that the opportunistic user is hoping to exploit. An overview of a typical spectrum sensing architecture can be seen in Figure 1. After the signal has been captured and sampled, a transform on the signal is done to produce a statistic, $r[n]$, which is used to make a decision, \hat{h} , on the state of the wireless spectrum, (0 = vacant, 1 = occupied). This decision is made by partitioning the domain of $r[n]$ into regions that are most likely to represent that the channel is either occupied or vacant. These decision regions are dictated by the distribution of the test statistic under either the null hypothesis (vacant) or the alternative hypothesis (occupied). Since the test statistic is derived from a transformation of the measured spectrum, the corresponding distribution is determined by environment specific fading and subject to change based on variation in the propagation environment.

To mitigate the variability in performance that comes from the changing wireless environment, a modification of the existing sensor structure is proposed that incorporates the use of a reconfigurable antenna and the algorithm which is capable of configuring the various antenna states. An overview of this system can be seen in Figure 2. This two stage algorithm first estimates the performance of the spectrum sensor using

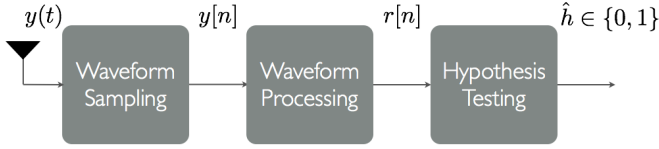


Fig. 1: System diagram of conventional spectrum sensor.

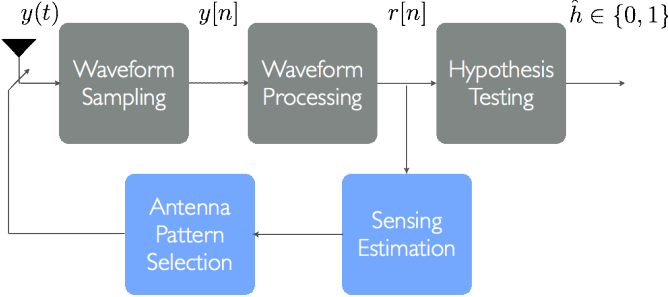


Fig. 2: System diagram of spectrum sensor incorporating ERAS.

the given antenna configuration. This estimate is obtained through characterization of the test statistic, $r[n]$. The estimate of sensor performance is then used to compute a weighted reward which indicates the utility of using the current state for spectrum sensing which is used to select the appropriate antenna state for the next sensing phase. We provide more details in Sec II-B

A. Sensing Performance Estimation

Assuming that the spectrum sensor is an energy detector, the distribution of the test statistic, $r[n]$, is determined by the distribution of the noise energy under the null hypothesis, and the energy of the signal and noise present under the alternative hypothesis. There have been a wide variety of statistical models used to represent the energy of signals with wireless fading, and while noise energy is typically assumed to be additive white Gaussian noise (AWGN), it is desirable to be able to account for non-Gaussian sources of noise as well. For this reason, we model the test statistic as a mixture of Gamma random variables. Gamma random variables are a useful model since i) they generalize a variety of common fading models in the exponential family such as Rayleigh, Nakagami, and χ^2 , and ii) they can approximate other models such as lognormal random variables with reasonable accuracy [16], [17]. A mixture model is used to represent the fact that the statistic is distributed differently depending on the true state of the channel, and this true state is obscured by the non-zero probability of error in spectrum sensing decisions.

The expectation maximization (EM) algorithm provides a robust mechanism for estimating the parameter of various mixture components in a mixture model despite this uncertainty about the state of the channel at the time the energy samples are collected [18]. Additionally, the EM algorithm has the added benefit of converging efficiently when the mixture

components have distributions in the exponential family [19]. This procedure iteratively converges on maximum likelihood estimates (MLE) for the given parameters by:

- 1) Selecting initial parameter estimates
- 2) Computing a conditional likelihood based on initial parameter estimates
- 3) Obtaining revised parameter estimates as the parameter values that maximize the computed conditional likelihood function
- 4) Repeating steps 2) and 3) until convergence

B. Antenna Configuration Selection Technique

Once the estimates are obtained for the parameters of both the noise and signal distributions, a reward function, μ_i is computed to establish how well a given antenna configuration, i , is optimized for spectrum sensing. There are two types of error that can occur from this hypothesis test. A false positive occurs when the spectrum sensor decides the wireless spectrum is occupied when it is in fact vacant. A false negative occurs when the spectrum sensor decides the wireless spectrum is vacant when it is in fact occupied. Since a test can be devised that produces arbitrarily small probability of false positive or false negative at the cost of increasing the other, a more effective method is minimizing the probability of one type of error under a constraint on the other. More specifically, the reward for a given configuration will be minimizing the probability of false negative given a constraint on the probability of false positive, or selecting the antenna configuration which provides the optimal constant false alarm rate (CFAR) detector [20].

If an energy detector is being used, then the image of the test statistic, $r[n]$ is the non-negative real line. Since the likelihood of a signal being present increases with an increase in the value of $r[n]$, the decision rule is straightforward. If the $r[n] \geq \gamma$ then the detector decides that the spectrum is occupied. The critical value, γ , is then determined to meet the false positive constraint such that

$$P_+ = \int_0^\gamma \frac{1}{\Gamma(\alpha_0)} x^{\alpha_0-1} e^{-\beta_0 x} dx, \quad (1)$$

where, α_0, β_0 are the parameters of the noise power estimated from the EM algorithm. This quantity represents the probability that the vacant spectrum noise power exceeds the critical value, γ . The reward, μ , can then be calculated as

$$\mu = 1 - P_- \quad (2)$$

$$= \int_\gamma^\infty \frac{1}{\Gamma(\alpha_1)} x^{\alpha_1-1} e^{-\beta_1 x} dx, \quad (3)$$

where, α_1, β_1 are the statistic, $r[n]$, from the alternative hypothesis, computed from the EM algorithm, and γ is determined from (1). The reward, μ_i is calculated based on different estimates for each test statistic using each antenna configuration.

In this way, the reward $0 < \mu_i < 1$ is a calculation of the probability of correctly identifying an occupied channel

given a constraint on the rate of false positives. We define the optimal reward, $\mu^* = \max_i \mu_i$, as the highest probability of correct detection achievable from any of the antenna configurations. The strategy for selecting an antenna configuration is designed to balance the benefit of learning this reward for each configuration with the cost of using antenna configurations that have sub-optimal reward. The strategy for balancing this exploration and exploitation is based on a variant of a solution to the multi-armed bandit problem [21]. We define the regret, $R(n)$, of an antenna configuration selection strategy as

$$R(n) = \mu^* n - \mu_i \sum_{i=1}^K \mathbf{E}[T_i(n)] \quad (4)$$

where, $\mathbf{E}[T_i(n)]$ is the expected number of times antenna configuration i is played after n total plays. Regret identifies the loss in reward that comes from using sub-optimal antenna configurations. It has previously been shown that any antenna selection strategy whose reward, $\mu \rightarrow \mu^*$ will have regret, $R(n) = O(\ln(n))$. One such strategy for achieving optimal rate of regret growth is the upper confidence bound (UCB) algorithm derived in [21], which assigns each antenna configuration a priority

$$V_i = \mu_i + \sqrt{\frac{2 \ln(n)}{n_i}}, \quad (5)$$

where n is the total number of sensing decisions made, and n_i is the number of times antenna configuration, i , has been used for spectrum sensing. The proposed strategy uses the antenna configuration which maximizes this priority, V_i , for the next spectrum sensing decision. This strategy includes both an estimate for the true reward, μ_i , of using antenna configuration, i , as well as a weight that increases if configuration i is being used infrequently.

III. EXPERIMENT OVERVIEW

To verify the performance of this approach, wireless measurement were taken in the Bossone Research Enterprise Center on the campus of Drexel University. The environment was selected to be representative of an indoor office environment. An overview of the physical link topology can be seen in Figure 3. In this experiment, a single transmitter broadcast bursty traffic on a WIFI channel located at 2.462 GHz, and one receiver used an energy detector to attempt to detect the occupancy of the channel.

In this experiment, traffic was synchronized with spectrum sensing decisions. At each time slot, the transmitter would flip a coin to determine whether to transmit or not. At that same time, the receiver would measure the wireless spectrum and make a decision on whether to transmit or not. Relevant parameters for the experiment can be seen in Table I.

The receiver was deployed with a reconfigurable Alford loop antenna (RALA) [22], a pattern reconfigurable antenna with 4 directional antenna patterns that range 360° in azimuth and an omni-directional antenna pattern. The antenna pattern of each antenna configuration can be seen in Figure 4.

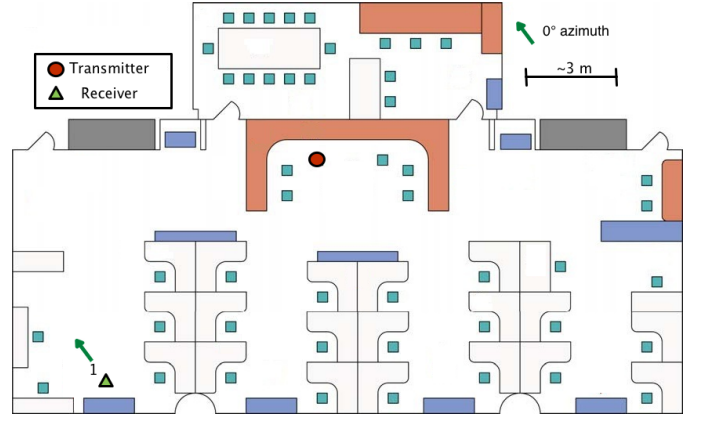


Fig. 3: Overview of radio locations for spectrum sensing experiment.

Parameter	Values
Probability of vacancy, π_0	0.5
Transmit power, P_{Tx}	1 dBm
Total samples, N	3000
EM iterations, T	15

TABLE I: Table of relevant experiment parameters.

Each radio used is a 3rd generation version of the wireless open access research platform (WARP) software defined radio (SDR) platform developed by Rice University [23]. Coordination and control of the radios was managed over a hardwired ethernet interface. All data generation for the transmitters and data analysis for the receivers was performed in MATLAB.

IV. MEASURED RESULTS

The measured configuration reward, P_{CD} , with false alarm rate constrained to 10% can be seen in Table II. As the results indicate, spectrum sensing performance for this radio is poor, most likely due to low transmit power of the broadcasting

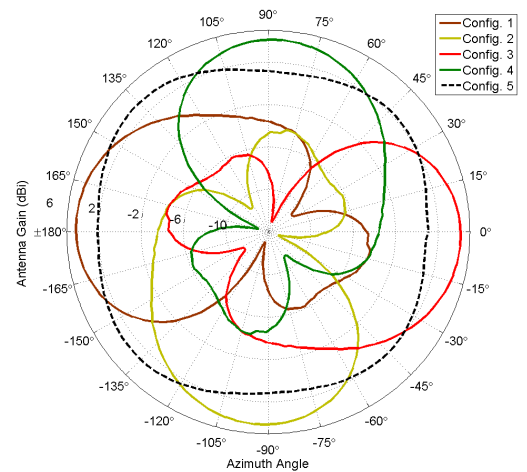


Fig. 4: Antenna patterns for each of the antenna configuration of the RALA.

Antenna Configuration	P_{CD}
1	.11
2	.41
3	.17
4	.12
5	.55

TABLE II: P_{CD} for each antenna configuration with constrained $P_{FA} = 0.1$.

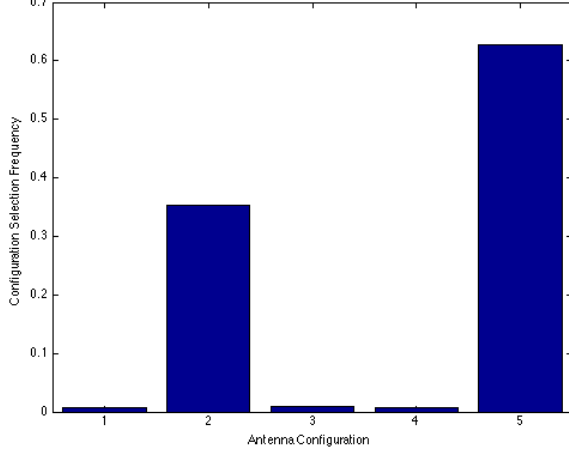


Fig. 5: Frequency of selection for each antenna configuration.

node. The omni-directional antenna configuration offers the best performance and with antenna configuration 2 offering better performance than any other directional pattern.

The UCB algorithm is designed to make use of antenna configurations proportional to their relative performance. This behavior is clear from observing the frequency of accessing each antenna configuration shown in Figure 5. Antenna configuration 5 (omni-directional) is selected for the large majority of spectrum sensing samples over the duration of the 3000 samples collected. Antenna configuration 2 is selected for approximately 30% of spectrum samples owing to its relatively strong performance. Antenna configurations with poor performance are largely unused. The results indicate a strong dependence on the environment in which the measurements are taken. It is reasonable to assume that an omni-directional antenna would offer the best performance in a rich scattering environment. In outdoor environments or in environments with sparse multi-path or at low signal-to-noise ratio (SNR), ERAS with directional patterns have demonstrated performance gains over conventional antennas [24], [25]. Therefore, we believe the sensing performance of directional modes of the antenna may outperform omni-directional mode in such cases. Further, in dense cognitive radio networks, it is possible to use omni-directional antenna modes to maximize sensing accuracy and then use one of the directional modes for spectrum access to reduce interference and maximize network capacity.

Figure 6 shows the effective performance of the ERAS spectrum sensor after 3000 spectrum samples. The optimal ROC shows the performance of a controller with perfect CSI. The

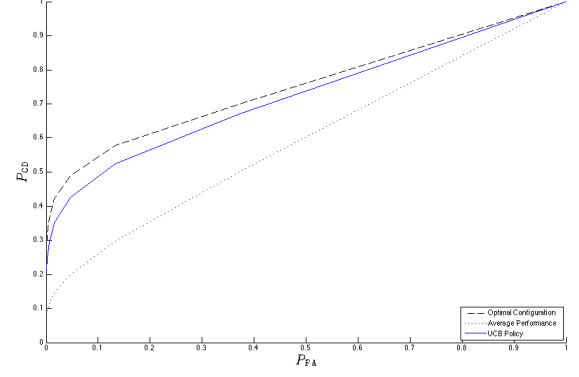


Fig. 6: Effective ROC of ERAS spectrum sensor.

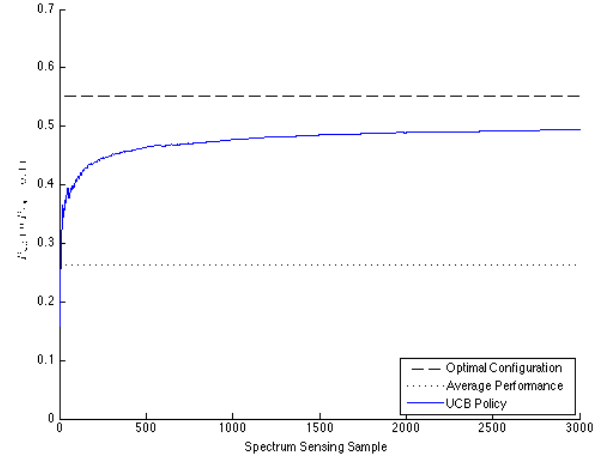


Fig. 7: Effective P_{CD} of ERAS spectrum sensor over time.

UCB approach performs to 89.7% of optimal performance, and 188% of average performance that would be seen with a random selection strategy.

Figure 7 shows the rate of convergence of P_{CD} ($@P_{FA} = 0.1$) UCB algorithm in time. This convergence in performance is quite rapid, converging to 80% of optimal performance within 200 samples. As discussed previously, the UCB algorithm will converge asymptotically on the optimal performance as the number of captured samples tends towards infinity.

The experimental regret is shown in Figure 8. Visual inspection of the regret growth shows that it appears to be sub-linear, however it is difficult to verify whether growth is truly logarithmic from experimental results. Achieving sub-linear regret will result in better performance than an exhaustive training or randomized approach that will have linear growth in regret. Further improvements in regret growth may be obtained from modification of the antenna control algorithm.

Table III shows the MLE of α and β parameters of the signal distribution that can only be obtained with perfect information about the occupancy of the channel. Additionally, the EM estimates of the α and β parameters obtained from the ERAS with state uncertainty are shown. As the results show, the

Antenna Configuration	α	β
config. 1		
-MLE parameters	.97	.0006
-EM parameters	.94	.0006
config. 2		
-MLE parameters	.66	.002
-EM parameters	.66	.003
config. 3		
-MLE parameters	.93	.0008
-EM parameters	.86	.0009
config. 4		
-MLE parameters	.96	.0006
-EM parameters	.84	.0008
config. 5		
-MLE parameters	.50	.008
-EM parameters	.50	.008

TABLE III: α and β parameters for signal distribution estimated from experiment

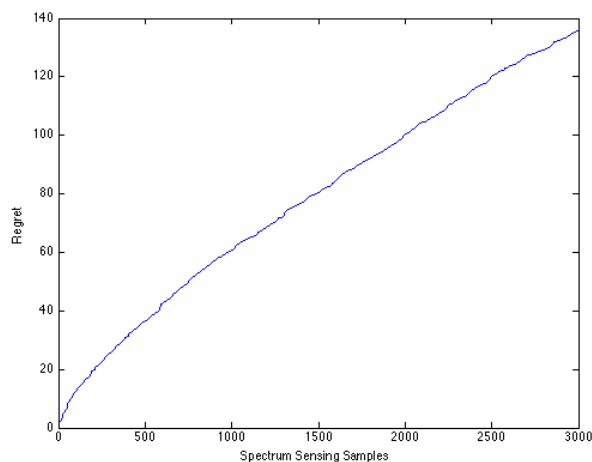


Fig. 8: Experimental Regret of the UCB policy.

ERAS estimates of the signal parameters are quite accurate despite low SNR. Results indicate that the UCB algorithm is capable of accurately estimating sensor performance. This result is notable particularly for antenna configurations that are rarely used. Reliable parameter estimates for rarely used antenna configurations can be obtained from minimal samples, making it possible to improve the effective performance from the frequent use of optimal antenna configurations.

V. CONCLUSION

An online algorithm is proposed for the optimal selection of ERAS antenna configurations for the purpose of spectrum sensing. This configuration selection algorithm minimizes the overhead required to estimate the performance of each antenna configuration, allowing optimal exploitation of the optimal antenna states for the desired objective. In addition, the use of the EM algorithm for Gamma mixtures enables the reliable estimation of spectrum sensing performance without directly observing whether spectrum sensing decisions are made correctly.

Experimental measurements taken in an indoor office environment validate the approach, and demonstrate that the

ERAS selection algorithm achieves rapid convergence in time and provides robust performance estimates even in low SNR scenarios. Measured results show that the ERAS can be leveraged to exploit pattern and polarization diversity to improve spectrum sensing performance. Computationally simple algorithms make it possible for DSA networks to improve CSI reliability locally with little overhead. Additionally, the approach provided is flexible enough to work with any type of reconfigurable antenna with minimal modification.

While results indicate that an omni-directional antenna pattern works well for energy detectors in office environments with rich multi-path scattering, the use of directional antenna patterns has been proven to provide gains in environments where multi-path is sparse. Additionally, spectrum sensors that rely on more complex analysis of the spectrum may benefit from the spatial filtering offered by directional patterns. In both cases, the added diversity of pattern and polarization can demonstrate improvements in both sensing and access performance through the use of intelligent antenna configuration control.

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