

Stochastic Geometry Analysis of Spatial-Temporal Performance in Wireless Networks: A Tutorial

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Abstract—The performance of wireless networks is fundamentally limited by the aggregate interference, which depends on the spatial distributions of the interferers, channel conditions, and user traffic patterns (or queueing dynamics). These factors usually exhibit spatial and temporal correlations and thus make the performance of large-scale networks environment-dependent (i.e., dependent on network topology, locations of the blockages, etc.). The correlation can be exploited in protocol designs (e.g., spectrum-, load-, location-, energy-aware resource allocations) to provide efficient wireless services. For this, accurate system-level performance characterization and evaluation with spatial-temporal correlation are required. In this context, stochastic geometry models and random graph techniques have been used to develop analytical frameworks to capture the spatial-temporal interference correlation in large-scale wireless networks. The objective of this article is to provide a tutorial on the stochastic geometry analysis of large-scale wireless networks that captures the spatial-temporal interference correlation (and hence the signal-to-interference ratio (SIR) correlation). We first discuss the importance of spatial-temporal performance analysis, different parameters affecting the spatial-temporal correlation in the SIR, and the different performance metrics for spatial-temporal analysis. Then we describe the methodologies to characterize spatial-temporal SIR correlations for different network configurations (independent, attractive, repulsive configurations), shadowing scenarios, user locations, queueing behavior, relaying, retransmission, and mobility. We conclude by outlining future research directions in the context of spatial-temporal analysis of emerging wireless communications scenarios.

Index Terms—Large-scale wireless access networks, signal-to-interference ratio (SIR), spatial-temporal correlation, point process modeling, stochastic geometry.

I. INTRODUCTION

Wireless communications systems are evolving toward a heterogeneous architecture (e.g., multi-tier and cell-free) with the dense deployment of different types of access points (e.g., smallcells and hotspots) to enable pervasive wireless Internet access [1]. The evolving wireless networks are expected to provide seamless connectivity to ubiquitous and/or high-mobility devices and users with millisecond delay and gigabits per second data rate [2]. The ever-increasing demand for low-latency high-reliability services from pervasive terminals will

lead to an explosive increase in mobile traffic. To accommodate the massive traffic volume, high network densification and aggressive spatial frequency reuse will be required, which will result in high levels of interference in the network.

A. Background of Stochastic Geometry and Objective

Signal propagation over a radio link is impaired by large-scale path loss and shadowing, small-scale fading, as well as co-channel interference from concurrent transmissions. Since all of these effects are heavily location-dependent, the network spatial configurations become a dominant factor that determines the system-level performance. Hence, developing tractable approaches for modeling large-scale wireless systems and analyzing their statistical performance taking into account the randomness (due to the above-mentioned factors) have become compelling.

In this context, stochastic geometry [3] (also referred to as geometric probability), a probabilistic analytical approach to study (random) point configurations, has become a necessary theoretical tool for the analysis and characterization of large-scale wireless systems, including heterogeneous cellular networks [4]–[6], dynamic spectrum access systems [7], [8], wireless ad hoc networks [9]–[12], drone networks [13]–[15], vehicular networks [16]–[18], and low earth orbit satellite networks [19], [20]. Spatial-temporal aspects of mobile communications, including the spatial distribution of network nodes, wireless channels and traffic patterns have to be considered for system development, resource allocation, performance evaluation and optimization. The purpose of this paper is to provide a tutorial on how to quantitatively analyze the effects of spatial and temporal fluctuations of interference (resulting from the factors above) on the system-level network performance.

B. Importance of Characterization of Signal-to-Interference-plus-Noise Ratio Correlation

Wireless communications systems need to preserve the quality of radio links in time-varying environments. The signal-to-interference-plus-noise ratio (SINR) statistics is commonly used as the main measure of the quality of links [21], and most of the performance metrics for system-level evaluation (to be introduced in Section II-C) are based on the SINR. The variation of the wireless environment (and hence the SINR) is mainly attributed to two causes. On the one hand, the

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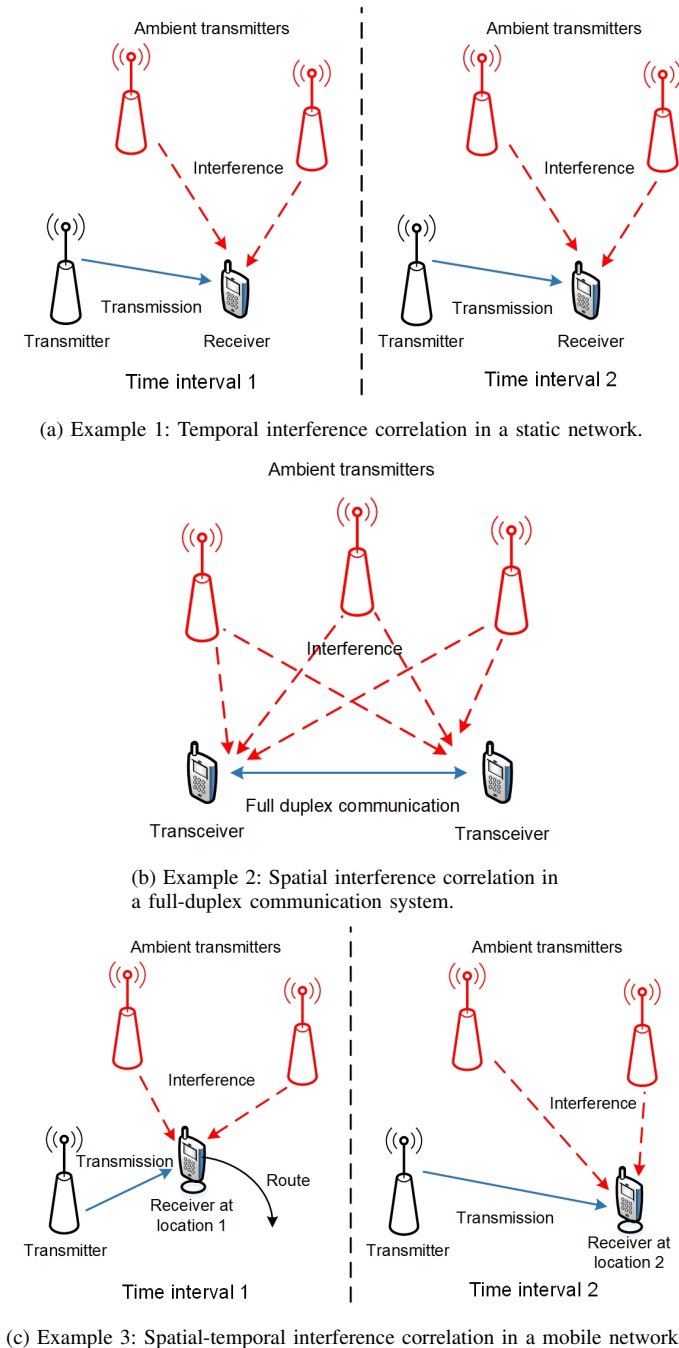


Fig. 1: Examples of interference correlation.

changes of the relative positions of communication devices and surrounding obstructions affect the multipath propagation and thus the received power of both desirable and interfering signals. Also, variations in traffic patterns cause fluctuations in the interference and hence the SINR.

Owing to the spatial-temporal fluctuations of network distributions, wireless channels and traffic patterns, the interferences and hence SINRs (at different locations and time instants) are correlated. Although rapid channel fluctuations due to small-scale fading can result in reduced interference correlation [22],

the correlation still exists due to large-scale path-loss and shadowing [23]. For example, in a static network, as shown in Fig. 1(a), if the ambient transmitters have data packets to send during two time intervals, the interferences at the receiver are temporally correlated. Besides, in a full-duplex communication system as shown in Fig. 1(b), the interferences at the transceiver pair are spatially correlated at any time instance. In a mobile network, as shown in Fig. 1(c), the interference at the receiver is spatially and temporally correlated. The causes of the temporal and/or spatial interference correlation in the above examples all arise from the fact that the interference comes from the same group of transmitters. The interference correlation results in the spatial-temporal correlation of transmission outage/success [24]–[26], throughput [27], [28], mean local delay [29], [30], etc., thus needs to be treated carefully in the designs of mobile systems.

Since SINR correlation affects the performance at different transmission attempts (e.g., when using an error recovery method) and/or locations (e.g., in a relay-based system), an accurate characterization of it is essential to the understanding of wireless network performance. Information about SINR correlation can be exploited to optimize the performance and design of the system accordingly.

C. Related Work

Several survey and tutorial papers have focused on the stochastic geometry analysis of wireless communication networks. In particular, reference [31] provides a survey of point process models and stochastic geometry tools that have been used to analyze static wired, wireless, ad hoc and cellular networks prior to 2009. Reference [32] overviews the impact of spatial modeling on the SINR-based performance metrics, i.e., connectivity, coverage area, and capacity of different types of systems, including ad hoc, cellular and cognitive networks. The survey in [33] comprehensively reviews the works on stochastic geometry analysis of multi-tier and cognitive cellular systems prior to 2013.

In addition to the above survey papers, tutorial papers on the mathematical tools used for stochastic geometry analysis of large-scale systems have also been written. Reference [34] is the first tutorial on point process theory, random geometry graphs, and percolation theory for interference characterizations in ad hoc networks. Targeting cellular networks, [35] provides a tutorial on stochastic geometry analysis of both downlink and uplink networks based on Poisson point process (PPP) modeling. The focus is on characterizing interference in different scenarios by exploiting the properties of the PPP under Rayleigh fading assumption. As PPP modeling fails to capture the spatial correlation among the random points, the authors in [36] emphasize the use of repulsive point processes to model cellular networks, where the base station locations are usually planned with a moderate degree of irregularity due to different development issues. To this end, the authors present a tutorial on the SINR distribution analysis of downlink cellular networks based on the β -GPP (Ginibre point process),

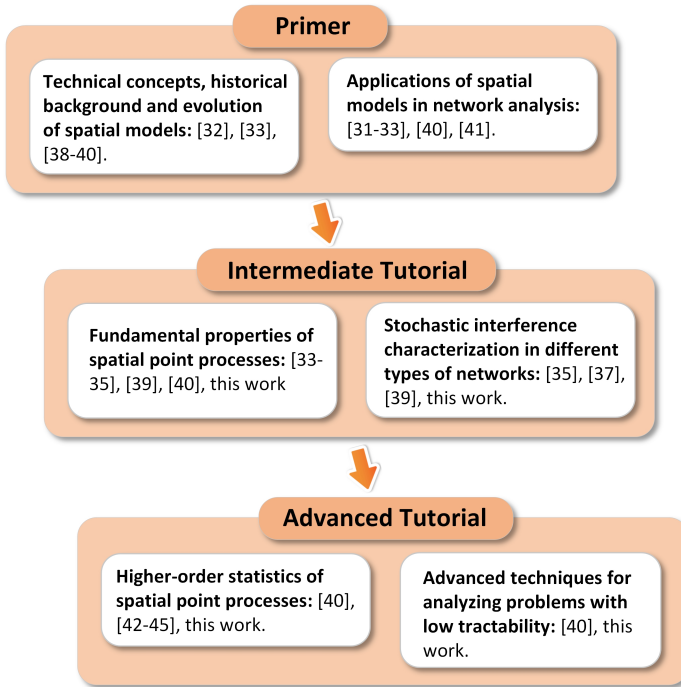


Fig. 2: The suggested learning path for stochastic geometry analysis in wireless networks.

which is a fairly tractable model for random points with spatial repulsion. The tutorial in [37] focuses on a unifying analysis of bit/symbol error probability, coverage outage probability, and ergodic capacity in cellular networks. More recently, reference [40] systematically tutors the analytical techniques to characterize interference, success probability and capacity in Poisson networks. **However, none of the existing survey and tutorial papers focus on the stochastic geometry techniques to characterize the spatial and temporal correlations in their considered systems.** Moreover, this is the first article to include the refined-grained analysis tool of the signal-to-interference ratio (SIR) meta distribution.

For a better understanding of the contents of this tutorial, we recommend taking some prerequisite knowledge of stochastic geometry regarding network analysis which has been comprehensively reviewed in the literature. The suggested learning path and the recommended readings corresponding to the prerequisite knowledge are shown in Fig. 2, which includes the following.

- Primers [31]–[34], [38]–[41] that cover the technical concepts of spatial models and their applications in modeling wireless networks;
- Intermediate tutorials [33]–[35], [37], [39], [40] that tutor the fundamental properties of spatial point processes (e.g., counting measure, superposition, thinning, and transforms) and analytical methods to characterize the interference distribution in different types of wireless networks (e.g., cellular and ad hoc networks);
- Advanced tutorials [40], [42]–[45] that lays out the higher-

order statistics of point processes for fine-grained analysis (e.g., product/joint meta distribution) and up-to-date analytical techniques (e.g., approximation and bounding methods) to analyze complicated scenarios (e.g., with spatial-temporal correlation) where the exact performance characterization is of low tractability or unavailable.

D. Contributions and Organization

This tutorial aims to concisely present the analytical approaches for performance evaluation of large-scale wireless networks taking into account various correlation causes and effects, such as shadowing, traffic queueing, and spatial distribution of network nodes. We focus on the characterization of the interference distribution and signal-to-interference ratio (SIR)-based performance metrics¹, e.g., *success probability*, *joint success probability*, and *moments of conditional success probability (CSP) given the point process*. For spatial point process models, we additionally derive the *asymptotic SIR gain*, which is the horizontal gap between a target SIR distribution and a reference SIR distribution. This metric directly reflects the variation of SIR due to the changes in the network model with respect to (w.r.t.) the reference model. Furthermore, this metric can be utilized to simplify the analysis of non-Poisson networks based on the PPP [47] (to be introduced in Section III-B).

This tutorial considers the following correlation effects in wireless networks.

- The spatial correlation (i.e., attraction and repulsion) among the locations of the transmitters;
- The spatially-correlated shadowing experienced by the links that traverse common obstacles (e.g., buildings);
- The spatially and temporally correlated queue status among the transmitters due to the cross-interference imposed on each other over space and time;
- The spatial correlation experienced by users located in the cell-center and cell-boundary regions;
- The temporal interference correlation between multiple transmission attempts due to the correlation in the locations of the interferers over time;
- The spatial-temporal interference correlation among relaying nodes in a multihop network due to the correlation in the locations of the interferers over space and time;
- The spatial-temporal interference correlation in mobile systems (i.e., where the users and/or base stations (BSs) are mobile) due to the correlation among interferers' locations over space and time.

This tutorial presents the methodologies of analyzing the interference (and SIR) correlation and the effects of spatial-temporal SIR correlations. For a thorough exposition of the analytical techniques, we demonstrate the step-by-step derivations

¹Since in large-scale cellular networks, the impact of aggregate interference typically dominates that of noise [6], [46], this tutorial focuses on analyzing the interference-limited cases with the noise ignored. However, without loss of generality, the same analytical approaches can also be applied to characterize SINR-based performance metrics.

TABLE I: Main differences between other survey and tutorial papers

Reference (year)	Type of Review	Examined systems				Spatial models		Target performance metrics	Spatial-temporal correlation analysis
		Ad hoc	Cellular	Multihop	Mobile	Poisson	Non-Poisson		
[31] (2009)	Survey	✓	✓			✓		Success probability, paging, handover	No
[34] (2009)	Tutorial	✓				✓		Interference characterization, outage probability, capacity, and area spectral efficiency	No
[32] (2010)	Survey	✓	✓			✓		Success probability, coverage area, capacity	No
[33] (2013)	Survey	✓	✓					Success probability, capacity	No
[35] (2016)	Tutorial		✓			✓		Success probability	No
[36] (2016)	Tutorial		✓				✓	Success probability	No
[37] (2017)	Tutorial		✓	✓		✓		Interference characterization, error probability, error rate, outage probability, capacity, and handover	No
[40] (2021)	Tutorial		✓			✓	✓	Interference characterization, success probability, capacity	No
This work	Tutorial	✓	✓	✓	✓	✓	✓	Success probability, joint success probabilities, conditional success probabilities, moments of conditional success probability given the point process, SIR meta distribution, SIR gain, interference correlation coefficient	Yes

as well as numerical results². The main differences between our tutorial and the state-of-the-art discussed in the previous subsection are summarized in Table I³. Herein, Poisson spatial models refer to the PPP and binomial point process, while non-Poisson spatial models refer to any point process whose points are not independently distributed.

The organization of this tutorial and the relations among different sections are shown in Fig. 3. We restrict the point processes to the Euclidean spaces \mathbb{R} and \mathbb{R}^2 . However, the same methodologies can be applied to analyze point processes in higher dimensions without loss of generality. Section I introduces the role of stochastic geometry analysis and highlights the importance of spatial-temporal correlation characterization. Section II explains the correlation effects of different network factors that may affect the SIR-based network performance. Sections III-VI present exact methodologies to analyze network performance with correlations in node distribution, link distance distribution, shadowing, and queueing, respectively. Moreover, Sections VII and VIII present a performance characterization with multiple transmission attempts (i.e., retransmission) and multihop relaying, respectively, for correlated and independent interference. Section IX characterizes the spatial-temporal performance with mobility. Future directions and research

challenges are then discussed in Section X followed by the conclusion in Section XI. Additionally, for convenience, we list the abbreviations used in Table II.

Notations: The notations defined in Table III are used throughout this tutorial.

II. OVERVIEW OF SPATIAL-TEMPORAL PERFORMANCE CORRELATION

This section first discusses the impact of SIR components and then elaborates on the SIR correlation effects in wireless networks and defines the performance metrics for spatial-temporal performance analysis.

A. Parameters Impacting SIR Correlation

Let o denote the origin of \mathbb{R}^d . The SIR at a target receiver located at o expressed as

$$\text{SIR} = \frac{P_y h_y S_y \ell(\|y\|)}{\sum_{x \in \Phi^1} \iota_x P_x h_x S_x \ell(\|x\|)}, \quad (1)$$

where Φ^1 denotes the set of all the interfering transmitters, y the location of the transmitter associated with the target receiver, P_x the transmit power of transmitter at x , h_x (S_x) the fading (shadowing) coefficient between the transmitter located at x and receiver located at o , ℓ the path loss function, and ι_x the state indicator of the transmitter located at x which equals 1 and 0

²Note that the analytical methodologies introduced in this tutorial come from the referenced literature. The analytical results are either well-established or direct extensions of the ones derived in the references.

³Herein, success probability refers to the complementary cumulative distribution function (CDF) of the SINR/SIR.

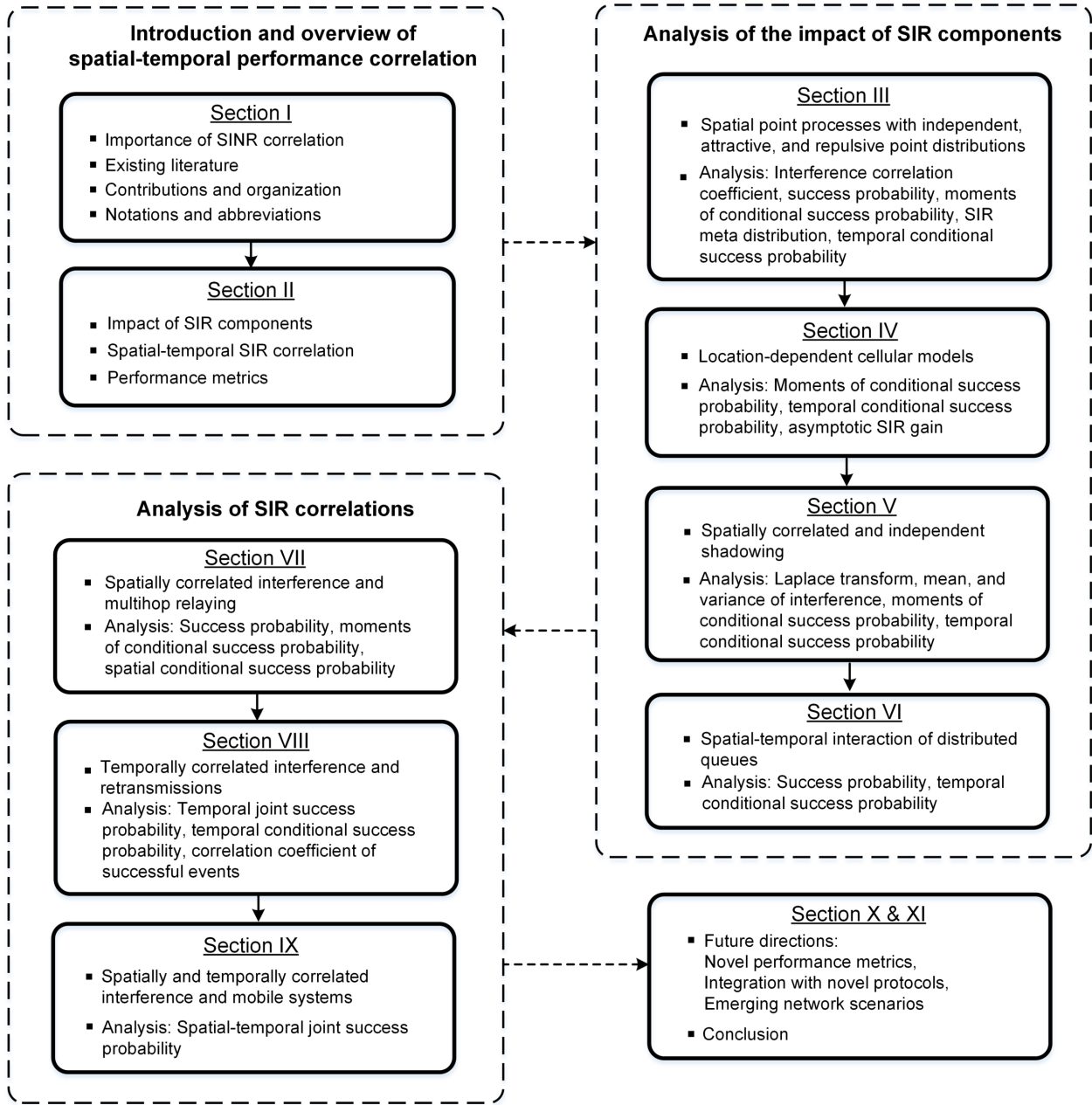


Fig. 3: Organization of this paper.

when the transmitter is on and off, respectively. The physical implications of the SIR components are shown in Table IV.

The SIR varies due to the fading and shadowing effects and system factors such as timing-varying traffic loads and positions of transmitters and receivers. Their impacts are discussed below.

1) *Spatial Distribution*: The spatial distribution of the network nodes can be categorized into three types: independent, repulsive, and attractive.

- With independent distribution, the locations of the transmitters are independent of each other. Spatial point process models to characterize independent distributions include PPP and binomial point process (BPP) [48] models.
- With repulsive distribution, the wireless transmitters trans-

mitting simultaneously are not too close to each other. The repulsion may arise from planned deployment, physical restrictions (e.g., geographic exclusion and terrain occlusion) and channel access control (e.g., carrier sense multiple access (CSMA) [49]–[52] and licensed-user activity [7], [53]) as shown in Fig. 4. Lattice processes [39], Ginibre point processes [36], [54], [55], and Matérn hardcore point processes [49], [56] are some well-known examples of point process models to characterize this repulsive behavior.

- Attractive distributions can be observed when wireless transmitters are only clustered in certain regions, i.e., not identically distributed over the entire plane, as illustrated

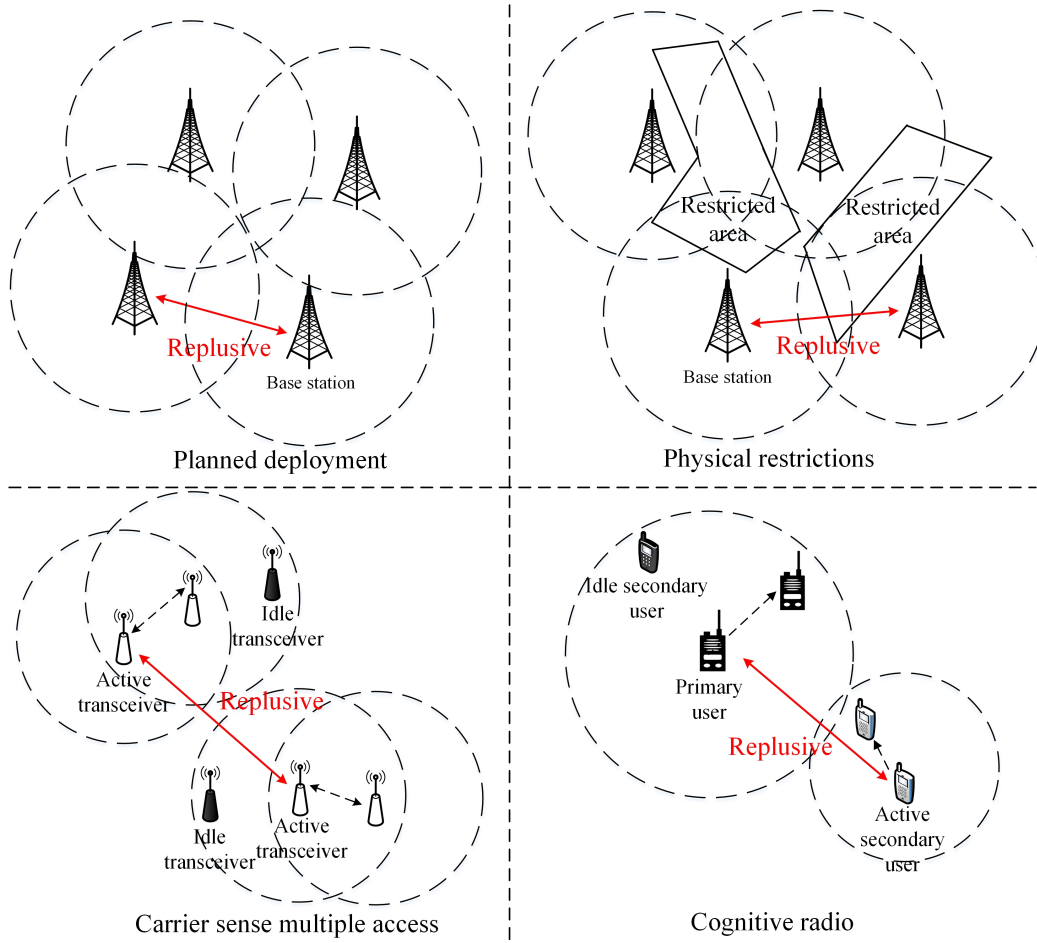


Fig. 4: Examples of wireless systems with spatially repulsive node locations.

in Fig. 5. This can be caused by the base station (BS)-centric user gathering [57], (e.g., around open-access WiFi spots) or due to user-centric BS deployment [58]. The Matérn cluster process [59], [60], Thomas cluster process [57], Gauss-Poisson process [61], [62], Cox process [16], [63], [64], and the Poisson hole process [7], [53] are some representative point processes to model attractive spatial distributions.

Additionally, a point process can be a mix of the above three types. Whether a point process is of any type may depend on the distance between the locations considered. For example, the type I user point process introduced in [65] (for modeling user distribution in cellular networks) is repulsive at short distances, attractive at intermediate distances and eventually approaches independence at larger distances.

For network performance analysis, a significant body of literature adopts the independent distribution assumption for simplicity and tractability (see [33] and references therein). However, this assumption is too idealized to hold in practice as most of the cellular networks are deployed under system-level planning [66]. Due to factors such as geographical restriction, access control and resource allocation, active network nodes

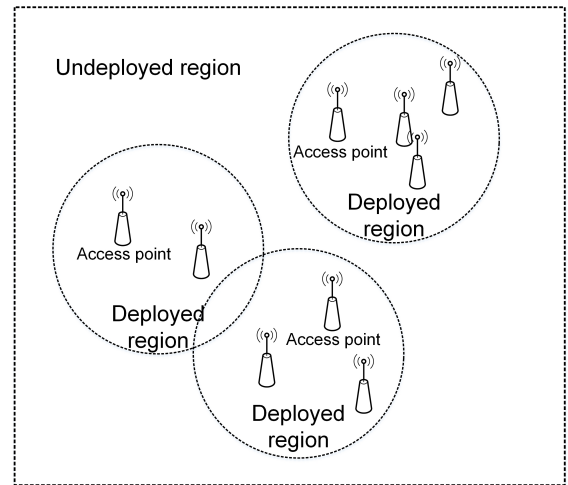
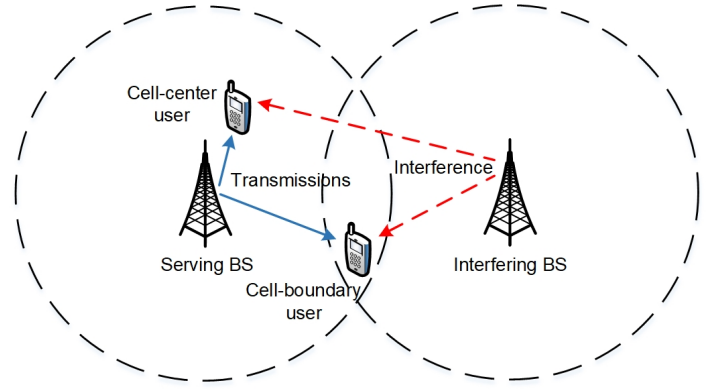
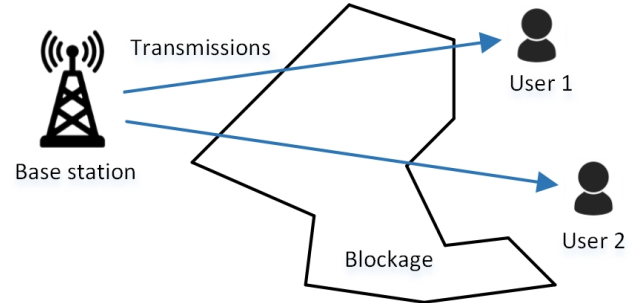


Fig. 5: Illustration of a wireless system with spatially attractive deployment.

can exhibit a spatial pattern with a degree of correlation (i.e., repulsion or attraction) which cannot be ignored in performance characterization.

TABLE II: List of abbreviations

Abbreviation	Description
PPP	Poisson point process
PLP	Poisson line process
BPP	Binomial point process
MCP	Matérn cluster process
GPP	Ginibre point process
RDP	Relative distance process
SINR	Signal-to-interference-plus-noise ratio
SIR	Signal-to-interference ratio
CSP	Conditional success probability
JSP	Joint success probability
MISR	Mean interference-to-signal ratio
LSU	Location-specific user
ASAPPP	Approximate SIR analysis based on the PPP
MIMO	Multiple-input and multiple-output
PDF	Probability density function
CDF	Cumulative distribution function
PCF	Pair correlation function
PGFL	Probability generating functional
MAC	Medium access control
CSMA	Carrier-sense multiple access
BS	Base station
DF	Decode-and-forward
HARQ	Hybrid automatic repeat request
QSI	Quasi-static interference
FVI	Fast-varying interference
IoT	Internet of Things
IoNT	Internet of Nanothings
IoST	Internet of Space Things
ML	Machine learning

**Fig. 6:** Illustration of location-specific users.**Fig. 7:** Illustration of correlated shadowing.

2) *Locations of Receivers:* In cellular networks, the transmission performance of a mobile user is highly dependent on its location w.r.t. the serving BS and interfering BSs [67]. Specifically, for a cell-edge user, the desired signal is weaker and the interference signal is stronger compared to those for a cell-center user (Fig. 6). Different techniques have been developed to strengthen the desirable signals and/or to mitigate the interference of cell-edge users, e.g., through coordinated beamforming [68] and intercell interference coordination [69], [70]. With these techniques, resource allocation strongly depends on the spatial variation of mobile users. Therefore, location-dependent modeling of user performance is fundamental to the understanding of spatial-temporal performance of these systems.

3) *Small-Scale Fading:* Owing to small-scale fading, over only a fraction of a wavelength, the signal power variation may reach up to 40 dB [71]. The correlation of the signal strength can be both temporal and spatial and is frequency dependent [21]. The temporal correlation occurs due to movement of environmental scatters. The spatial correlation is caused by correlated multipath components due to the common propagation environment. However, small-scale fading is only correlated over a short time duration and a small distance (e.g., several wavelengths) in environments with moving scattering objects that change the multipath propagation. Hence it is reasonable to adopt independent and identically distributed (i.i.d.) small-scale fading models for receiving antennas with wavelength separation. This assumption is widely used in the existing

literature (see [33], [37] and references therein).

4) *Shadowing:* Compared to small-scale fading, shadowing is correlated at a much larger time and space scale [72]. For example, in an urban environment, shadowing caused by obstacles in communication paths can be geographically correlated on a scale from 50 to 200 meters [21]. Moreover, temporal shadow fades with variations below 1 dB (i.e., highly correlated) over an IEEE 802.11ad channel can be commonly measured in urban streets [73]. Fig. 7 demonstrates an example of correlated shadowing. As the two links between the BS and users experience a similar propagation environment, their shadowing attenuations tend to be correlated. As evidenced by experimental studies in [74] and [75], path attenuations in a region are correlated due to shadowing effects. Therefore, correlated shadowing should be carefully treated when assessing system performance that is heavily impacted by environmental blockages.

5) *Transmission Buffer Status:* The buffer status (i.e., queuing status) of each transmitter depends on the packet arrival and service processes, and it impacts the activation of the transmitters, and therefore, the mutual interference. In view of this, the buffer statuses of the transmitters are interdependent, leading to *interacting queues*. The queues are spatially coupled since the mutual interference directly affects the transmission success probability (i.e., service processes of the queues). Also, the queues are temporally coupled since the current buffer status is affected by the previous departure process. Due to the random nature of channel fading and aggregate interference, the service

TABLE III: Notations

Symbol	Definition
$j = \sqrt{-1}$	The imaginary unit
\equiv	Equivalence relation
$\mathbb{N}, \mathbb{R}, \mathbb{R}^+$	Natural numbers, real numbers, and positive real numbers, respectively
\mathbb{R}^d	d -dimensional Euclidean space
o	The origin of \mathbb{R}^d
$\mathbb{B}(x, R)$	Disk of radius R centered at x
Φ	Point process representing the nodes in the network
Φ^I	The set of interferers
$\Phi(A)$	The number of elements in $\Phi \cap A$
\triangleq	The definition operator
$\stackrel{(d)}{=}$	Equivalence in distribution
$\mathbb{1}_{\{\cdot\}}$	Indicator function which equals 1 and 0 if the statement $\{\cdot\}$ is true and false, respectively
$\mathbb{E}[\cdot]$	Expectation operator
$\mathbb{E}_x^t[\cdot]$	Expectation operator w.r.t. the reduced Palm measure at x
$\mathbb{P}[\cdot]$	Probability measure
$\mathbb{P}_x^t[\cdot]$	Reduced Palm probability measure at x
$\mathbb{V}[\cdot]$	Variance
$ \cdot $	Modulus operator
$\ \cdot\ $	Euclidean norm
$\Gamma(\cdot)$	Gamma function
$\gamma(s, x)$	Lower incomplete gamma function, i.e., $\gamma(s, x) = \int_0^x u^{s-1} e^{-u} du$
$\mathcal{E}(a)$	Exponential distribution with rate parameter a
$\mathcal{P}(a)$	Poisson distribution with rate parameter a
$\mathcal{G}(a, b)$	Gamma distribution with shape parameter a and scale parameter b
Det	Determinant operator
${}_2F_1(\cdot, \cdot; \cdot; z)$	The Gauss hypergeometric function
$f_X(\cdot), F_X(\cdot), \mathcal{L}_X(\cdot)$	The probability density function (PDF), the cumulative distribution function, and Laplace transform of random variable X .
$\mathcal{W}(\cdot)$	The Lambert- \mathcal{W} function, i.e., the inverse function of $f(x) = xe^x$

TABLE IV: SIR components

Component	Representation	Characterization
Φ	Spatial distribution of interferers	Spatial point process models (in Section III)
$r_1 = \ x_1\ $	Contact distance distribution	Location-dependent analysis of cellular models (in Section IV)
S_x	Blockage	Shadowing models (in Section V)
ι_x	Buffer status	Queueing models (in Section VI)

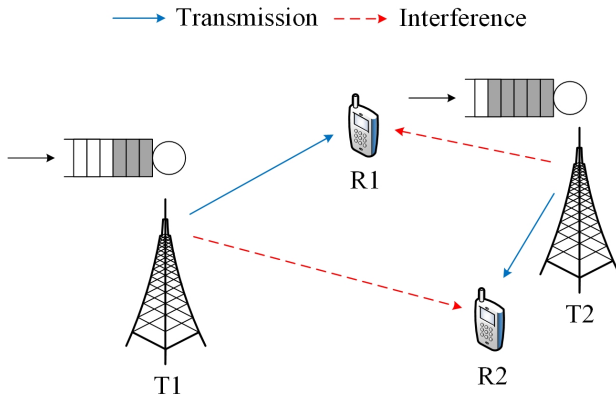


Fig. 8: Illustration of the interacting queues.

processes can be very dynamic.

Fig. 8 depicts an example of the correlation between the queues at two transmitters. Transmitter T1, which serves receiver R1, has a longer transmission link and a shorter interference link compared to those of transmitter T2, which serves receiver R2. On the one hand, the channel disparity

between the two communication links results in different packet service rates. Compared to T2, T1 suffers from more severe path loss and interference. Correspondingly, given a similar traffic load T1 tends to vacate its queue more slowly and thus remain active more frequently than T2. On the other hand, the queue lengths of T1 and T2 determine the activation of T1 and T2. In particular, if both transmitters are busy, their transmissions will cause mutual interference, which slows down the departure process. If one of the transmitters has an empty queue, the other enjoys a speedy departure process. Due to the interacting queues, transmissions in a large-scale system inherently experience spatial and temporal correlation.

Finally, Fig. 9 summarizes the sources of SINR correlation and their spatial-temporal impact discussed in this subsection.

B. Spatial-Temporal SIR Correlation and Network Scenarios

Even over independent fading channels, aggregated interference is spatially correlated when it comes from the same group of interferers. Similarly, interference becomes temporally correlated when common static interferers exist across time. Spatially and temporally-correlated interference causes spatial-temporal SIR correlation. When different transmission attempts

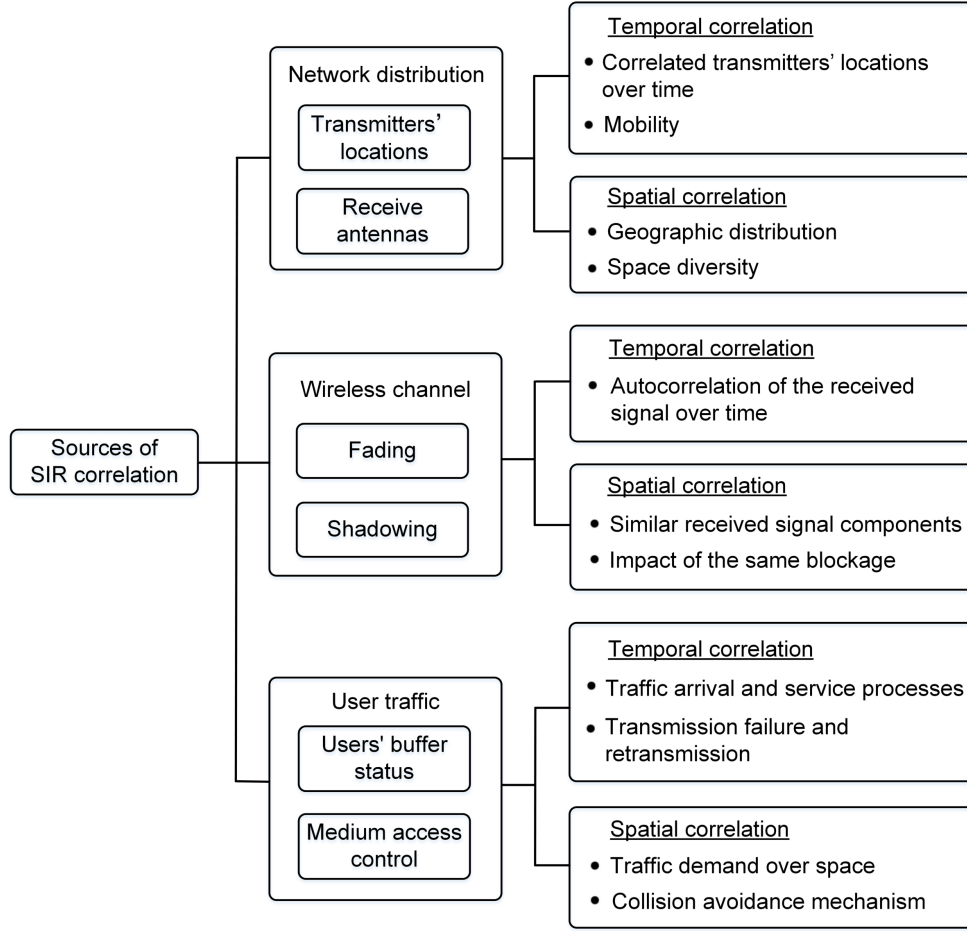


Fig. 9: Sources of SINR correlation.

are affected by the same group of static interferers (e.g., the same point process), it gives rise to a *quasi-static interference* (QSI) scenario. By contrast, a *fast-varying interference* (FVI) [76] scenario results when different transmission attempts are affected by different sets of interferers⁴ (e.g., independent point processes). As two extreme cases, QSI and FVI, respectively, render the highest and lowest interference correlation caused by the interferer locations.

Spatial-temporal SIR correlation affects the performance of many practical systems, where the transmission performance depends on the SIR measured over different space and time spots. Typical examples of these systems include the following:

- *Multihop relaying*: The end-to-end performance of a multihop relaying system depends on the SIRs at the receivers at different hops. The SIRs at different hops can be correlated due to spatial-temporal interference correlation (e.g., common interferers exist from one transmission to another).
- *Retransmission*: The reliability of a retransmission scheme or a multi-packet transmission scheme depends on the SIR

at the receiver during multiple transmissions. The SIRs from different transmissions can be correlated if they are subject to the interference from the same group of transmitters.

- *Mobile networks*: For a mobile user, transmissions could occur at different spatial locations and times. The correlation between the SIRs at the receiver for different transmissions are dependent on the mobility.

The SIR correlation affects the performance metrics such as temporal joint success probability (JSP) [77], end-to-end success probability [78], the local delay [29], and joint SIR meta distribution [43].

C. Metrics for Spatial-Temporal Performance Analysis

Definition 1. (Success Probability): The (average) success probability of a user can be defined as the probability that the user's received SIR is greater than a threshold denoted by θ [66], i.e.,

$$P_s = \bar{F}_{\text{SIR}}(\theta) \triangleq \mathbb{P}[\text{SIR} > \theta], \quad \theta \in \mathbb{R}^+. \quad (2)$$

where \bar{F}_{SIR} represents the complementary CDF of SIR.

⁴The QSI and FVI assumptions can be used for modeling static and highly-mobile network scenarios, where the interferers across different transmissions remain the same and become completely different, respectively.

Definition 2. (Moments of Conditional Success Probability given the Point Process): Let $P_{s|\Phi}(\theta) = \mathbb{P}[\text{SIR} > \theta \mid \Phi]$ represent the CSP given the point process Φ (abbreviated as CSP_Φ), which is averaged over the fading of all the links and the random channel access (if applicable). The b -th moment of $P_{s|\Phi}$ is defined as [79]

$$\mathcal{M}_{P_s}(b) \triangleq \mathbb{E}\left[\left(P_{s|\Phi}(\theta)\right)^b\right], \quad b \in \mathbb{C}. \quad (3)$$

Note that the first moment of the CSP_Φ is the average success probability defined in (2), i.e., $\mathcal{M}_{P_s}(1) \equiv P_s$, and the variance is $\mathbb{V}(P_s) = \mathcal{M}_{P_s}(2) - \mathcal{M}_{P_s}^2(1)$. Besides, the mean local delay, defined as the average number of transmission attempts to accomplish a success [80], is given by $\mathcal{M}_{P_s}(-1)$. It is worth mentioning that for static random networks and $b \in \mathbb{N}$, the b -th moment of the CSP_Φ is equivalent to the JSP of b transmissions of the same link [81] and JSP that the b antennas of a multiple-antenna receiver all succeed in reception (i.e., SIR exceeds θ) [82].

Definition 3. (SIR Meta Distribution): The SIR meta distribution is the complementary CDF of the CSP_Φ , defined as [44], [79]

$$\bar{F}_{P_s}(\theta, s) \triangleq \mathbb{P}[P_{s|\Phi}(\theta) > s], \quad s \in [0, 1], \quad (4)$$

where s represents the target success probability.

In ergodic point processes, the SIR meta distribution indicates the fraction of links that can achieve successful transmission with probabilities greater than x in any realization of Φ . Compared to the average success probability defined in (2), the SIR meta distribution also characterizes the variability of link success probabilities. For example, the inverse function of the SIR meta distribution $\bar{F}_{P_s}^{-1}(p)$, $p \in [0, 1]$, yields the success probability x that a fraction $1 - p$ of the links achieve while the rest do not.

Example: With $\theta = 1$ and $s = 0.9$, $\bar{F}_{P_s}(1, 0.9)$ represents the fraction of links/transmissions that achieve a receive SIR greater than 1 with probability at least 0.9.

Definition 4. (SIR Gain): The SIR gain is the horizontal gap between the complementary CDF of two SIR distributions. Evaluated at the target success probability P_t , the SIR gain is defined as [83, Eq. (1)]

$$G(P_t) \triangleq \frac{\bar{F}_{\text{SIR}_{\text{tm}}}^{-1}(P_t)}{\bar{F}_{\text{SIR}_{\text{rm}}}^{-1}(P_t)}, \quad P_t \in (0, 1), \quad (5)$$

where $\bar{F}_{\text{SIR}}^{-1}$ represents the inverse function of the complementary CDF of the SIR, and SIR_{tm} and SIR_{rm} denote the SIRs of a target model and a reference model, respectively.

The SIR gain can be used to quantify the impact of a target model on the SIR distribution w.r.t. that of a reference model. An SIR gain greater or less than 1 indicates that the target model can achieve the same success probability with a larger or smaller SIR threshold, respectively, than the reference model.

In wireless networks, the SIR gain is usually not sensitive to the target success probability to be evaluated [83], [84]. Therefore, the SIR gain can be approximated by the asymptotic SIR gain evaluated in the high-reliability regime, i.e., $P_t \rightarrow 1$ or $\theta \rightarrow 0$, defined as

$$G_0 \triangleq \lim_{P_t \rightarrow 1} G(P_t), \quad (6)$$

whenever the limit exists.

Note that a necessary and sufficient condition for the asymptotic SIR gain to exist is that the slopes of the two CDFs of the SIR are asymptotically the same as $\theta \rightarrow 0$ [83].

Definition 5. (Joint Success Probability): Let $\mathbf{x} = \{x_k\}_{k \in \{1, \dots, K\}} \in (\mathbb{R}^d)^K$ and $\mathbf{t} = \{t_k\}_{k \in \{1, \dots, K\}} \in \mathbb{N}^K$ denote deterministic vectors of locations and time instances, respectively, and SIR_k denote the SIR measured at location x_k and time t_k . The JSP is defined as the probability that SIR_i is greater than the corresponding SIR threshold θ_k for all $k \in \{1, \dots, K\}$. Given the locations \mathbf{x} , times \mathbf{t} , and the target SIR thresholds $\boldsymbol{\theta} = \{\theta_k\}_{k \in \{1, 2, \dots, K\}} \in (\mathbb{R}^+)^K$, the JSP is defined as

$$\begin{aligned} \mathcal{J}_K(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) \\ \triangleq \mathbb{P}[\text{SIR}_1 > \theta_1, \text{SIR}_2 > \theta_2, \dots, \text{SIR}_K > \theta_K], \end{aligned} \quad (7)$$

where SIR_k denotes the SIR of the k -th transmission.

The SIR corresponding to the different transmissions can exhibit spatial and/or temporal correlation, the effect of which will be reflected in the JSP. The JSP can be for temporal, spatial, or spatial-temporal transmission events. In particular, we have

- *Temporal JSP* of multiple transmission attempts occurring at the same location but different time instances (e.g., multiple transmissions [81] and retransmissions [85] by a transmitter);
- *Spatial JSP* of multiple transmission attempts occurring at the same time but different locations (e.g., joint uplink and downlink transmissions [86], [87]);
- *Spatial-temporal JSP* of multiple transmission events occurring at different space and time intervals (e.g., for the same mobile user [77]).

For a multihop communication scenario, the probability that the transmissions at each hop are successful for a given packet is referred to as the *end-to-end success probability*.

Definition 6. (Conditional Success Probability): The CSP is the probability of achieving a successful transmission given that $K - 1$ ($K \geq 2$) such events have occurred. Given the locations \mathbf{x} , times \mathbf{t} , and the target SIR thresholds $\boldsymbol{\theta}$, the CSP is given by

$$\mathcal{C}_K(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) = \frac{\mathcal{J}_K(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t})}{\mathcal{J}_{K-1}(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t})}, \quad (8)$$

where $\mathcal{J}_K(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t})$ is defined in (7).

The CSP reveals the dependence between two successful transmission events. If the two events are positively

correlated (spatially or temporally), one has $\mathcal{C}_2(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) > \mathcal{J}_1(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t})$ [81]. Moreover, if the events are independent, one has $\mathcal{C}_2(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) = \mathcal{J}_1(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) = \sqrt{\mathcal{J}_2(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t})}$. Similar to the definition of the JSP, the CSP of spatial events, temporal events, and spatial-temporal events are referred to as *spatial CSP*, *temporal CSP*, and *spatial-temporal CSP*, respectively.

Definition 7. (Product SIR Meta Distribution): Let $\mathcal{J}_{K|\Phi}(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) \triangleq \mathbb{P}[\text{SIR}_1 > \theta_1, \dots, \text{SIR}_K > \theta_K \mid \Phi]$ denote the JSP of K transmissions given the point process. The product SIR meta distribution is defined as the complementary CDF of the JSP at locations \mathbf{x} and times \mathbf{t} given the point process Φ [43, Eq. (11)], i.e.,

$$\bar{F}_{\mathcal{J}_K}(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}, s) \triangleq \mathbb{P}[\mathcal{J}_{K|\Phi}(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}) > s], s \in [0, 1]. \quad (9)$$

Example: Let $K = 2$, $\boldsymbol{\theta} = \{\theta_1, \theta_2\}$, $\mathbf{x} = \{x_1, x_2\}$, $\mathbf{t} = \{t_1, t_2\}$. Considering a stationary and ergodic point process $\Psi = \{u_j\}_{j \in \mathbb{N}}$ representing the users, the product SIR meta distribution measures the fraction of users for which the joint probability that the SIR at their location u at time t_1 exceeds θ_1 and the SIR at location $u + (x_2 - x_1)$ at time t_2 exceeds θ_2 is larger than s .

Definition 8. (Joint SIR Meta Distribution): Let $P_{s|\Phi}^{(k)}(\theta_k) = \mathbb{P}[\text{SIR}_k > \theta_k \mid \Phi]$ represent the CSP_Φ at x_k . The joint SIR meta distribution is defined as the joint distribution of CSP_Φ at locations \mathbf{x} and times \mathbf{t} [43, Eq. (10)], i.e.,

$$\bar{F}_{P_s}^{(K)}(\boldsymbol{\theta}, \mathbf{x}, \mathbf{t}, s) \triangleq \mathbb{P}\left[\bigcap_{k=1}^K \{P_{s|\Phi}^{(k)}(\theta_k) > s_k\}\right], \quad (10)$$

where $\mathbf{s} = \{s_1, s_2, \dots, s_K\} \in [0, 1]^K$ is the vector of target success probabilities corresponding to the locations \mathbf{x} .

Example: Let $K = 2$, $\boldsymbol{\theta} = \{\theta_1, \theta_2\}$, $\mathbf{x} = \{x_1, x_2\}$, $\mathbf{t} = \{t_1, t_2\}$, $\mathbf{s} = \{s_1, s_2\}$. Considering a stationary and ergodic point process $\Psi = \{u_j\}_{j \in \mathbb{N}}$ representing the users, the joint SIR meta distribution measures the fraction of users that meet the following conditions: 1) the probability that the SIR at the user location u at time t_1 exceeds θ_1 is larger than s_1 and 2) the probability that the SIR at location $u + (x_2 - x_1)$ at time t_2 exceeds θ_2 is larger than s_2 .

Definition 9. (Interference Correlation Coefficient): Let $I_x^{(t)}$ denote the aggregated interference received at location x at time t . The correlation degree of interference at two locations and times can be quantified by the *Pearson correlation coefficient* defined as [23, Eq. (2)]

$$\zeta_{t_1, t_2}(u_1, u_2) \triangleq \frac{\mathbb{E}[I_{u_1}^{(t_1)} I_{u_2}^{(t_2)}] - \mathbb{E}[I_{u_1}^{(t_1)}] \mathbb{E}[I_{u_2}^{(t_2)}]}{\sqrt{\mathbb{E}[(I_{u_1}^{(t_1)})^2] - \mathbb{E}[I_{u_1}^{(t_1)}]^2} \sqrt{\mathbb{E}[(I_{u_2}^{(t_2)})^2] - \mathbb{E}[I_{u_2}^{(t_2)}]^2}}, \quad (11)$$

where the numerator computes the covariance of $I_{u_1}^{(t_1)}$ and $I_{u_2}^{(t_2)}$ and the denominator is the product of the standard deviations of $I_{u_1}^{(t_1)}$ and $I_{u_2}^{(t_2)}$.

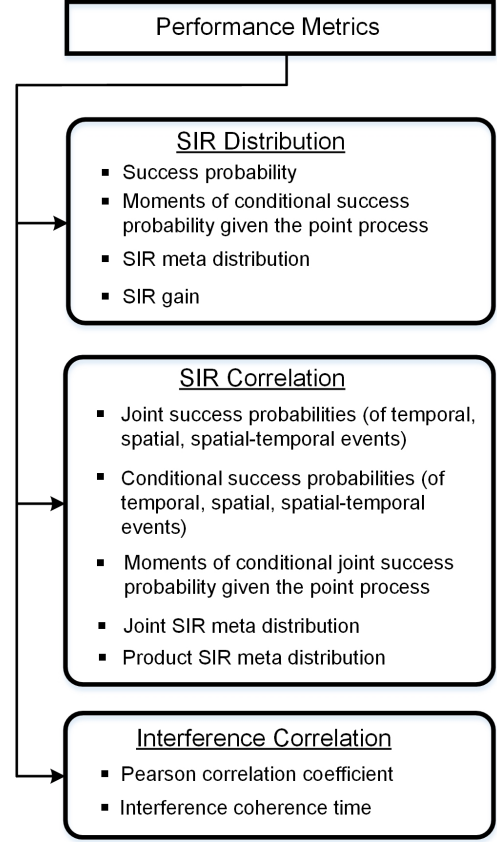


Fig. 10: Metrics for spatial-temporal performance analysis.

The interference correlation coefficient is a statistical measure that quantifies the extent to which the interferences at u_1 and u_2 are associated. The value of the interference correlation coefficient ranges between 0 and 1. The coefficients of 0 and 1, respectively, indicate no linear relationship and full correlation between the interferences at u_1 and u_2 .

Note that when the interferers are motion-invariant, one has $I_o^{(t_1)} \stackrel{(d)}{=} I_u^{(t_1)}$. In this case, $\zeta(\|u\|)$ can be simplified to [22, Eq. (12)]

$$\zeta_{t_1, t_2}(\|u_1 - u_2\|) \triangleq \frac{\mathbb{E}[I_{u_1}^{(t_1)} I_{u_2}^{(t_2)}] - \mathbb{E}[I_{u_1}^{(t_1)}]^2}{\mathbb{E}[(I_{u_1}^{(t_1)})^2] - \mathbb{E}[I_{u_1}^{(t_1)}]^2}. \quad (12)$$

Definition 10. (Interference Coherence Time): The interference coherence time is defined as the minimum time lag such that the interference correlation coefficient is below a threshold ζ_{th} [88, Eq. (35)], i.e.,

$$\tau_{ct} \triangleq \min \{\tau = t_2 - t_1 \in \mathbb{N} \mid \zeta_{t_1, t_2} \leq \zeta_{th}\}, \zeta_{th} \in \mathbb{R}^+. \quad (13)$$

Fig. 10 summarizes the above performance metrics for spatial-temporal performance analysis.

III. SPATIAL POINT PROCESS MODELS

This section formally defines some common point processes, also referred to as random point fields, and illustrates how the spatial distribution of the random points affects the network performance. To evaluate the impacts of independent, attractive and repulsive spatial configurations, we choose three representative point processes, namely, the Poisson point process (PPP), the Matérn cluster process (MCP), and the β -Ginibre point process (GPP), due to their tractability.

- In a PPP, each point is located independently from the others.
- In an MCP, the locations of the random points have a propensity to be clustering.
- In a β -GPP, the random points are scattered with repulsion.

This section considers both ad hoc and downlink cellular networks modeled based on the above point processes and analyzes the interference correlation coefficient, success probability, SIR meta distribution and SIR gain.

A. System Models

1) *System Configurations*: We focus on analyzing the performance of a target in both ad hoc and downlink cellular networks. The target receiver is considered to be at the origin $o \in \mathbb{R}^2$ and attempts to decode the transmitted signals from the associated transmitter subject to the aggregate interference.

- *Ad hoc networks*: The target receiver at o is served by the transmitter located at $x_t \in \mathbb{R}^2$, where $\|x_t\| = r_t$. The target link is impaired by the interference from a random field of interferers modeled by a stationary point process Φ of intensity λ . Φ does not contain the serving transmitter at x_t .
- *Downlink cellular networks*: The transmitters (i.e., BSs) form a stationary point process Φ with intensity λ . The users are located according to a stationary point process independent of the BS process. Each user is served by the nearest BS in Φ .

In the networks considered, the transmitters stay active with unit transmit power, i.e., there is no MAC scheme. All the transmitters and receivers are equipped with one antenna. The system employs universal frequency reuse. The channels of the links experience i.i.d. block Rayleigh fading and power-law path loss, i.e., $d^{-\alpha}$, where d and $\alpha > 2$ are the link distance and the path-loss exponent, respectively.

For the convenience of notation, the points in Φ are assumed to be ordered from nearest to farthest to the origin, i.e., $\|x_j\| < \|x_{j+1}\|$. Subsequently, the set of interferers in the ad hoc networks and downlink cellular networks is $\Phi^1 = \Phi = \{x_j\}_{j \in \mathbb{N}}$ and $\Phi^1 = \Phi \setminus \{x_1\}$, respectively. The distance from the j -th nearest point in Φ to the origin is denoted as $r_j = \|x_j\|$.

The SIR at the target receiver in an ad hoc and downlink cellular network is given by, respectively, as

$$\eta = \frac{h_t \|x_t\|^{-\alpha}}{\sum_{j \in \mathbb{N}} h_j \|x_j\|^{-\alpha}} \quad (14)$$

and

$$\eta = \frac{h_1 \|x_1\|^{-\alpha}}{\sum_{j=2}^{\infty} h_j \|x_j\|^{-\alpha}}, \quad (15)$$

where h_t and h_j denote the power fading coefficients from the transmitters at x_t, x_j to the target receiver, which are exponential random variables with unit mean, i.e., $h_t, h_j \sim \mathcal{E}(1)$. It is noted that (14) represents the SIR of a specific receiver instead of the typical receiver in ad hoc networks, while (15) is the SIR of a user at the origin that, upon averaging over the base station and user point processes, becomes the typical user in the downlink cellular network, for any stationary point process of users that is independent of the base station point process.

2) *Spatial Configurations*: We consider the homogeneous PPP, MCP, and β -GPP, formally defined as follows.

Definition 11. (Homogeneous Poisson point process): A point process $\Phi = \{x_j\}_{j \in \mathbb{N}} \subset \mathbb{R}^d$ is a homogenous PPP if it satisfies two conditions: i) for any compact set $\mathcal{B} \subset \mathbb{R}^d$, the number of points inside it follows a Poisson distribution with average value being $\lambda|\mathcal{B}|$; and ii) disjoint sets are independent in terms of the number of points inside them.

Properties:

1) **Density**: For a homogeneous PPP, the first moment density (or first-order density) and second moment density (or second-order product density) are given, respectively, as [89]

$$\rho^{(1)}(x) = \lambda, \quad (16)$$

and

$$\rho^{(2)}(x, y) = (\rho^{(1)}(x))^2 = \lambda^2. \quad (17)$$

The n -th moment density is the density pertaining to the n -th order factorial moment measure, which, for $n > 1$, indicate the spatial correlation. For example, $\rho^{(1)}(x)dx$ measures the probability of having a point at x in some infinitesimal region dx .

2) **Contact Distance Distribution**: Contact distance refers to the distance between a reference location and the nearest point in Φ [90]. The probability density function (PDF) and CDF of the contact distance in a homogeneous PPP are given, respectively, as [91]

$$f_{r_1}(r) = 2\pi\lambda \exp(-\lambda\pi r^2) \quad (18)$$

and

$$F_{r_1}(r) = \exp(-\lambda\pi r^2). \quad (19)$$

3) **Joint Distance Distribution**: The joint PDF of the distances to n nearest points is [92, Eq. (30)]

$$f_{r_1, r_2, \dots, r_n}(x_1, x_2, \dots, x_n) = e^{-\lambda\pi x_n^2} (2\lambda\pi)^n x_1 x_2 \dots x_n. \quad (20)$$

4) **Distance Ratio Distribution**: Let $\varrho_j = \frac{r_1}{r_j}$, $j \in \mathbb{N}$, denote the distance ratio of the nearest point to the j -th nearest point.

The CDF and PDF of ϱ_j are given, respectively, by [93, Lemma 3]

$$F_{\varrho_j}(\varrho) = 1 - (1 - \varrho^2)^{j-1}, \quad \varrho \in [0, 1] \quad (21)$$

and

$$f_{\varrho_j}(\varrho) = 2(j-1)\varrho(1 - \varrho^2)^{j-2}, \quad \varrho \in [0, 1]. \quad (22)$$

5) Probability Generating Functional (Product Functional): For $v(x) \in [0, 1]$ and $\int_{\mathbb{R}^2} (1 - v(x))dx < \infty$, the probability generating functional (PGFL) for the PPP is given by [3, Eq. (4.8)]

$$\mathbb{E} \left[\prod_{j \in \mathbb{N}} v(x_j) \right] = \exp \left(-\lambda \int_{\mathbb{R}^d} (1 - v(x))dx \right). \quad (23)$$

According to Slivnyak's theorem [3, Theorem 8.10], the reduced Palm distribution of the PPP is the same as its ordinary distribution. Therefore, the conditional PGFL for the PPP can also be expressed as (23).

Let $\Phi^R = \{x \in \Phi \setminus \{x_1\} : \|x_1\|/\|x\|\} \subset (0, 1)$ denote the *relative distance process* (RDP) of a PPP Φ . The PGFL of Φ^R is given by [84, Lemma 1]

$$\mathbb{E} \left[\prod_{\varrho \in \Phi^R} v(\varrho) \right] = \frac{1}{1 + 2 \int_0^1 (1 - v(r))r^{-3}dr}, \quad (24)$$

for functions $v(r) \in [0, 1]$ such that $\int_{\mathbb{R}^+} (1 - v(r))r^{-3}dr$ is finite.

6) Sum Functional: For any measurable function $g \geq 0$ on \mathbb{R}^d , the sum functional for a stationary point process is given by Campbell's Theorem [3, Theorem 4.1] as

$$\mathbb{E} \left[\sum_{j \in \mathbb{N}} g(x_j) \right] = \lambda \int_{\mathbb{R}^d} g(x)dx. \quad (25)$$

7) Sum-Product Functional: Let $\mathbf{p} \triangleq (p_1, p_2, \dots, p_q) \in \mathbb{N}^q$, $\|\mathbf{p}\|_1 \triangleq \sum_{k=1}^q p_k$ and $\{M_j\}_{j \in \mathbb{N}}$ denote a set of i.i.d. random marks associated with the points in $\Phi = \{x_j\}_{j \in \mathbb{N}}$. If $g_k(x) \in [0, \infty)$, $1 \leq k \leq q$, and $v(x) \in [0, 1]$ are measurable functions for all $x \in \mathbb{R}^d$, the sum-product functional for the PPP with $\|\mathbf{p}\|_1 > 0$ is given by [42, Theorem 1]

$$\begin{aligned} & \mathbb{E} \left[\prod_{k=1}^q \left(\sum_{j \in \mathbb{N}} g(x_j, M_j) \right)^{p_k} \prod_{j \in \mathbb{N}} v(x_j, M_j) \right] \\ &= \exp \left(-\lambda \int_{\mathbb{R}^d} \left(1 - \mathbb{E}_{M_j} [v(x, M_j)] \right) dx \right) \\ & \quad \times \sum_{l=1}^{\|\mathbf{p}\|_1} \sum_{N \in \mathcal{N}_l^p} \frac{D_N}{l!} \prod_{k=1}^l \lambda \int_{\mathbb{R}^d} \mathbb{E}_M [v(x, M_j)] \\ & \quad \times \prod_{i=1}^q f_i^{n_{ki}}(x, M_j) dx, \end{aligned} \quad (26)$$

where $D_N \triangleq \prod_{k=1}^q \frac{p_k!}{\prod_{i=1}^l m_{ik}!}$ and $\mathcal{N}_l^p \subset \mathbb{N}^{q \times l}$ is the class of all $q \times l$ matrices with the columns $\|n_{\cdot i}\|_1 > 0$, $\forall i \in \{1, 2, \dots, l\}$ and the rows $\|n_{k \cdot}\|_1 = p_k$, $\forall k \in \{1, 2, \dots, q\}$.

In the special case $q = 1$ and $p = 1$, the sum-product functional for the PPP is given by the Campbell-Mecke Theorem [3, Theorem 8.2] as

$$\begin{aligned} & \mathbb{E} \left[\sum_{j \in \mathbb{N}} g(x_j, M_j) \prod_{j \in \mathbb{N}} v(x_j, M_j) \right] \\ &= \exp \left(-\lambda \int_{\mathbb{R}^d} \left(1 - \mathbb{E}_{M_j} [v(x, M_j)] \right) dx \right) \\ & \quad \times \lambda \int_{\mathbb{R}^d} \mathbb{E}_{M_j} [g(x, M_j) v(x, M_j)] dx. \end{aligned} \quad (27)$$

Definition 12. (Matérn cluster process): The MCP Φ^M is a doubly Poisson cluster process constructed from a parent PPP $\Phi_p = \{x_j\}_{j \in \mathbb{N}}$ with intensity λ_p with each point of Φ_p substituted by a daughter cluster consisting of a PPP with an average number of points \bar{c} within a disk of radius R_d centered at that point.

Properties:

1) **Density:** For an MCP, the first moment and second moment densities are given, respectively, by [3]

$$\rho^{(1)}(x) = \lambda = \lambda_p \bar{c}, \quad (28)$$

$$\rho^{(2)}(x, y) = \lambda_p^2 \bar{c}^2 + \lambda_p \frac{\bar{c}^2}{\pi^2 R_d^4} A_{R_d}(\|x - y\|), \quad (29)$$

A_{R_d} represents the area of the intersection of two disks with radius R_d at a distance $r > 0$ given as [3, Eq. (6.4)]

$$A_{R_d}(r) = \begin{cases} 2R_d^2 \arccos(r/2R_d) - r\sqrt{R_d^2 - r^2/4}, & 0 \leq r \leq 2R_d, \\ 0 & \text{otherwise.} \end{cases} \quad (30)$$

2) **Daughter Point Distribution:** Each daughter point is located uniformly within a disk of radius R around the origin with the PDF given by [9, Eq. (5)]

$$f^M(y) = \begin{cases} \frac{1}{\pi R_d^2}, & \text{if } \|y\| \leq R_d \\ 0, & \text{otherwise.} \end{cases} \quad (31)$$

3) **Contact Distance Distribution:** The PDF and CDF of the contact distance in the MCP are given as (32) and (33), respectively [90].

4) **Probability Generating Functional:** The PGFL for an MCP is given as follows [9]:

$$\begin{aligned} & \mathbb{E} \left[\prod_{j \in \mathbb{N}} v(x_j) \right] \\ &= \exp \left(-\lambda \int_{\mathbb{R}^2} \left[1 - M \left(\int_{\mathbb{R}^2} v(x+y) f^M(y) dy \right) \right] dx \right) dy, \end{aligned} \quad (34)$$

where $M(t) = e^{-\bar{c}(1-t)}$ is the moment generating function of the representative cluster in an MCP.

Let $\mathbb{E}_o^l[\cdot]$ denote the expectation operation based on the reduced Palm measure [89] which takes the expectation for a

$$f_{r_1}^M(r|x) = \begin{cases} \frac{2r}{R_d^2}, & \text{if } 0 \leq r \leq R_d - x \text{ for } \|x\| < R_d \\ \frac{2r}{\pi R_d^2} \cos^{-1}\left(\frac{r^2 + x^2 - R_d^2}{2rx}\right), & \text{if } R_d - \|x\| \leq r \leq R_d + \|x\| \text{ for } \|x\| < R_d, \text{ and if } \|x\| > R_d, \end{cases} \quad (32)$$

$$F_{r_1}^M(r) = 1 - \exp\left(-\lambda_p\left(\int_{\mathbb{B}(0, R_d)} \left[1 - \exp\left(-\bar{c}\left(\int_0^{\min(r, R_d-x)} \frac{2y}{R_d^2} dy\right.\right.\right.\right.\right. \\ \left.\left.\left.\left. + \int_{\min(r, R_d-x)}^{\min(r, R_d+x)} \frac{2y}{\pi R_d^2} \cos^{-1}\left(\frac{y^2 + x^2 - R_d^2}{2yx}\right) dy\right)\right] dx \right.\right. \\ \left.\left. + \int_{\mathbb{R}^2 \setminus \mathbb{B}(0, R_d)} \left[1 - \exp\left(-\bar{c} \int_{\min(r, R_d-x)}^{\min(r, R_d+x)} \frac{2y}{\pi R_d^2} \cos^{-1}\left(\frac{y^2 + x^2 - R_d^2}{2yx}\right) dy\right)\right] dx\right)\right), \quad (33)$$

point process conditioned at a point of the process at o without including the point. The conditional PGFL of an MCP is [9, Lemma 1]

$$\begin{aligned} \mathbb{E}_o\left[\prod_{j \in \mathbb{N}} v(x_j)\right] \\ = \exp\left(-\lambda_p \int_{\mathbb{R}^2} \left[1 - M\left(\int_{\mathbb{R}^2} v(x+y) f(y) dy\right)\right] dx\right) \\ \times \int_{\mathbb{R}^2} G_d^M(v(x-y)) f^M(y) dy, \end{aligned} \quad (35)$$

where G_d^M is the PGFL for the representative cluster given by

$$G_d^M(v) = M\left(\int_{\mathbb{R}^2} v(x) f^M(x) dx\right). \quad (36)$$

Definition 13. (β -Ginibre point process): A β -GPP $\Phi_\beta^G = \{x_j\}_{j \in \mathbb{N}}$ is a determinantal point process with the kernel⁵ given by [54]

$$\mathcal{K}_{\beta, \lambda}(x, y) = \lambda e^{-\frac{\pi \lambda |x-y|^2}{2\beta}}, \quad x, y \in \mathbb{C}, \quad (37)$$

w.r.t. the Lebesgue measure on \mathbb{C} [94].

Properties:

1) **Density:** For a β -GPP, the first moment density and the second moment density of the β -GPP are given, respectively, by [54]

$$\rho^{(1)}(x) = \det[\mathcal{K}_{\beta, \lambda}(x, x)] = \lambda, \quad (38)$$

$$\begin{aligned} \rho^{(2)}(x, y) &= \det \begin{bmatrix} \mathcal{K}_{\beta, \lambda}(x, \bar{x}) & \mathcal{K}_{\beta, \lambda}(x, \bar{y}) \\ \mathcal{K}_{\beta, \lambda}(\bar{x}, y) & \mathcal{K}_{\beta, \lambda}(\bar{y}, y) \end{bmatrix} \\ &= \lambda^2 \left(1 - \exp\left(-\frac{\pi \lambda |x-y|^2}{\beta}\right)\right). \end{aligned} \quad (39)$$

⁵The kernel represents the interaction force among the points of the process.

2) **Link Distance Property:** Let λ represents the intensity of Φ_β^G . Let $\{Q_j\}_{j \in \mathbb{N}}$ be a set of independent gamma random variables with PDF

$$f_{Q_j}(q) = \frac{q^{j-1} e^{-\frac{\pi \lambda}{\beta} q}}{\left(\frac{\beta}{\pi \lambda}\right)^j \Gamma(j)}, \quad (40)$$

i.e., $Q_j \sim \mathcal{G}(j, \beta/\pi\lambda)$. Then the set $\{\|x_j\|^2\}_{j \in \mathbb{N}}$ is equivalent in distribution with the set constructed by retaining each element from $\{Q_j\}_{j \in \mathbb{N}}$ independently with probability β [96, Theorem 4.7.1].

Fig. 11(a), (b), (c), respectively, illustrate realizations of MCP, PPP, and β -GPP. As shown by the realizations of MCP and β -GPP, the point sets are attractive and repulsive, respectively.

The correlation between the spatial points can be measured by the pair correlation function (PCF). For a point process $\Phi \subset \mathbb{R}^d$, the PCF is defined as $g(x, y) \triangleq \frac{\rho^{(2)}(x, y)}{\rho^{(1)}(x)\rho^{(1)}(y)}$. If Φ is motion-invariant, the first moment density is the constant intensity and the second moment density $\rho^{(2)}(x, y)$ only depends on the difference $r = \|x - y\|$. Hence, the PCF can be expressed as $g(r) = \frac{\rho^{(2)}(r)}{\lambda^2}$.

Let $\Phi(A)$ denote the number of points in $\Phi \cap A$. The PCF quantifies the degree of correlation between the random variables $\Phi(A)$ and $\Phi(B)$ in a non-centered way. If $g(x, y) \equiv 1$ then for disjoint A and B the covariance of $\Phi(A)$ and $\Phi(B)$ is zero, which means the two variables are uncorrelated. For spatial point processes, the PCF describes how a point is surrounded by others. The PCF equals one if the points are uncorrelated (e.g., as in the PPP), and is greater (smaller) than 1 if the points are attractive (repulsive).

The PCFs of the MCP, PPP and β -GPP are given, respectively, as [3, Page 153], [54]

$$g(r) = \begin{cases} 1 + \frac{\bar{c} A_{R_d}(r)}{\lambda \pi^2 R_d^4}, & \text{MCP} \\ 1, & \text{PPP} \\ 1 - \frac{\exp(-r^2/\beta)}{\pi \lambda}, & \beta\text{-GPP}, \end{cases} \quad (41)$$

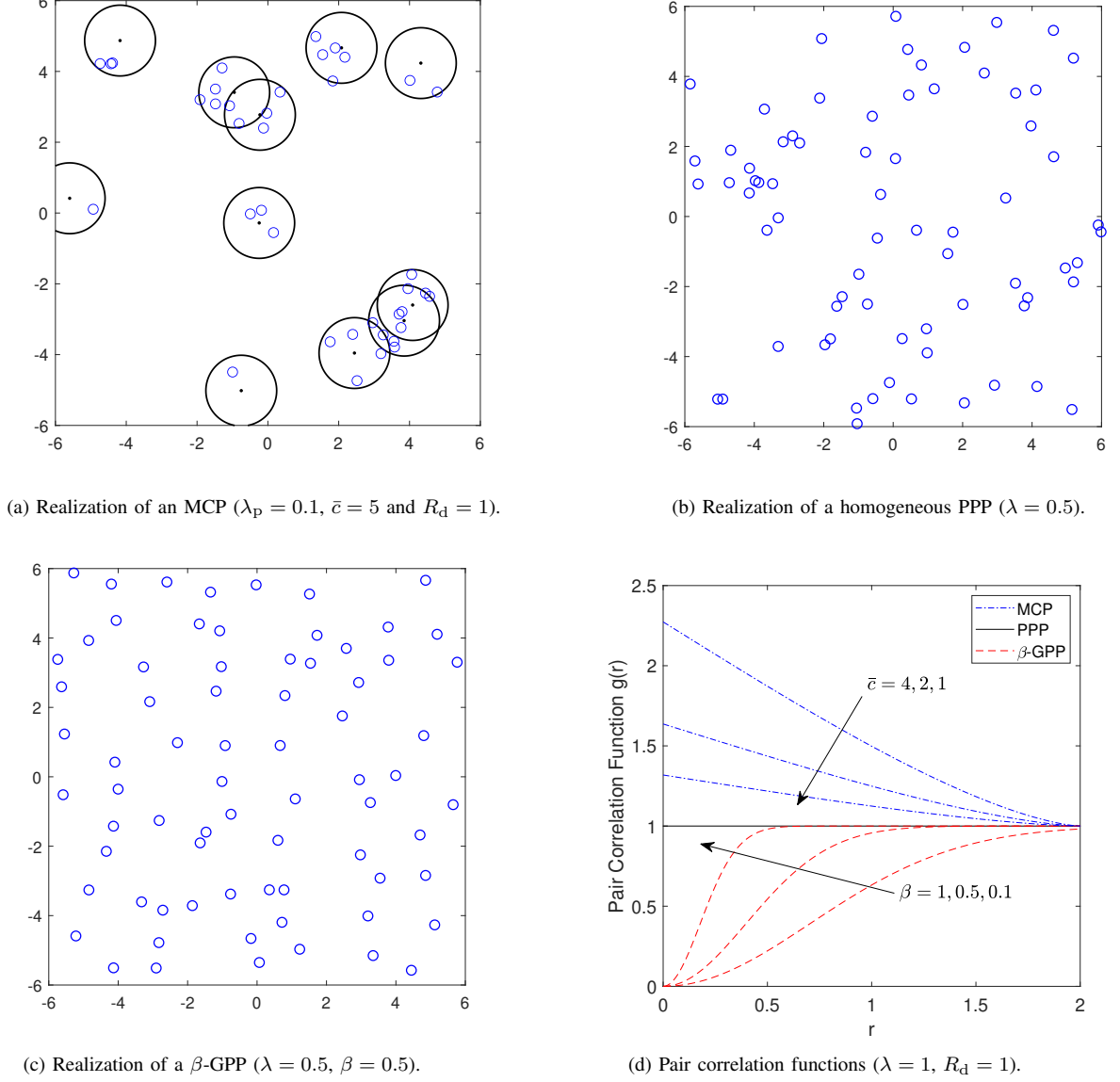


Fig. 11: Illustration of spatial point processes.

where $A_{R_d}(r)$ is given in (30).

Properties:

1) For $r \geq 2R_d$, for an MCP, the PCF is the same as that of the PPP, because two points with a distance greater than $2R_d$ must belong to different clusters and thus are independent.

2) Given the intensity λ , the PCF of an MCP approaches that of the PPP as $\bar{c} \rightarrow 0$, since the points are less likely to belong to the same cluster.

3) For a PPP, the PCF is not affected by the intensity since the points are independently distributed.

4) For a β -GPP, the PCF approaches that of the PPP as $\beta \rightarrow 0$ or $\lambda \rightarrow \infty$.

Fig. 11(d) shows the PCFs of the spatial point processes and demonstrates the properties discussed above.

3) *Validity:* The validity of the point process approximation of real-world cellular networks has been demonstrated in the literature. In particular, an early study in [95] found that the success probability of an actual cellular network (with a degree regularity) is lower-bounded by that of a Poisson network (with complete randomness). More importantly, reference [83] revealed that adding a horizontal SIR threshold shift from 0 dB to 3.4 dB to the success probability in a Poisson network yields a tight approximation of that in any network with a spatial layout from complete randomness to perfect regularity (i.e., triangular lattice). Hence, the analysis of an actual cellular network can be performed based on that of a Poisson network with a horizontally shifted SIR threshold calibrated to the regularity of the topology.

The validity of the β -GPP to model actual cellular networks

has been rigorously explored. The studies in [54], [97] show that the repulsion parameter β that adjusts the degree of repulsion among points can be numerically fitted with the data set from actual network deployments. The simulations demonstrate that the success probabilities from β -GPP analysis closely approximates the ones from simulations with the actual deployment topologies. Hence, it is valid to employ either a horizontally shifted PPP or the fitted β -GPP to analyze actual cellular networks.

B. Performance Analysis

1) *Spatial-Temporal Interference Correlation in Ad Hoc Networks*: We first investigate the correlation of interference observed at two locations o and u in two time slots t_1 and t_2 , respectively, with the interferers distributed as the three types of point processes considered. As PPP, MCP and β -GPP are motion-invariant, the spatial-temporal interference correlation can be measured by the Pearson correlation coefficient defined in (12).

As can be seen from (12), the expectation and the second moments of $I_o^{(t_1)}$ and the mean product of $I_o^{(t_1)}$ and $I_u^{(t_2)}$ are needed to quantify the interference correlation. However, these two quantities do not exist because of the singularity of the path-loss function $\ell(x)$. To cope with this issue, we follow the approach in [9] by defining $\ell_\epsilon(x) = \frac{1}{\epsilon + \|x\|^\alpha}$, $\alpha > 2$, $\epsilon > 0$, such that $\ell(x) = \lim_{\epsilon \rightarrow 0} \ell_\epsilon(x)$.

The expectation of $I_o^{(t_1)}$ can be derived as

$$\begin{aligned} \mathbb{E}[I_o^{(t_1)}] &\stackrel{(a)}{=} \mathbb{E}[I_o] \\ &= \mathbb{E}\left[\sum_{j \in \mathbb{N}} h_j \ell_\epsilon(x_j)\right] \\ &= \mathbb{E}[h] \int_{\mathbb{R}^2} \ell_\epsilon(x) \rho^{(1)}(x) dx \\ &\stackrel{(b)}{=} 2\pi\lambda \int_0^\infty \frac{x}{\epsilon + x^\alpha} dx \\ &\stackrel{(c)}{=} \delta\pi^2 \lambda \epsilon^{\delta-1} \csc(\delta\pi), \end{aligned} \quad (42)$$

where (a) follows since MCP, PPP, and β -GPP are all motion-invariant and the superscript (t_1) is dropped for conciseness, (b) applies the conversion from Cartesian to polar coordinates, and (c) substitutes $\frac{2}{\alpha}$ with δ .

If the densities of the interferers following a PPP, MCP, and β -GPP as given in (16), (28), and (38), respectively, are the same, we can observe from (42) that the three point processes cause the same mean interference at an arbitrary location. This indicates that spatial attraction and repulsion do not affect the first moment of the interference.

We then continue to derive the second moment of $I_o^{(t_1)}$ as

$$\begin{aligned} &\mathbb{E}\left[\left(I_o^{(t_1)}\right)^2\right] \\ &= \mathbb{E}\left[\left(\sum_{j \in \mathbb{N}} h_j \ell_\epsilon(x_j)\right)^2\right] \end{aligned}$$

$$\begin{aligned} &= \mathbb{E}\left[\sum_{j \in \mathbb{N}} h_j^2 \ell_\epsilon^2(x_j) + \sum_{j, i \in \mathbb{N}}^{j \neq i} h_j h_i \ell_\epsilon(x_j) \ell_\epsilon(x_i)\right] \\ &= \mathbb{E}[h^2] \int_{\mathbb{R}^2} \ell_\epsilon^2(x) \rho^{(1)}(x) dx \\ &\quad + \mathbb{E}[h]^2 \int_{\mathbb{R}^2} \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \rho^{(2)}(x, y) dx dy \\ &= 2\delta\lambda\pi^2(1 - \delta)\epsilon^{\delta-2} \csc(\delta\pi) \\ &\quad + \int_{\mathbb{R}^2} \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \rho^{(2)}(x, y) dx dy. \end{aligned} \quad (43)$$

By plugging in the second moment density $\rho^{(2)}(x, y)$ for the MCP, PPP, and β -GPP given, respectively, in (29), (17), (39), we obtain the second moment of the interference in (44), where \bar{I}_{PPP}^2 is

$$\begin{aligned} \bar{I}_{\text{PPP}}^2 &= 2\delta\lambda\pi^2(1 - \delta)\epsilon^{\delta-2} \csc(\delta\pi) \\ &\quad + \delta^2\pi^4\lambda^2\epsilon^{2\delta-2} \csc(\delta\pi)^2. \end{aligned} \quad (45)$$

Table V shows the variance of the interference with different fields of interferers. It can be observed that the MCP and the β -GPP cause larger and smaller interference variance than the PPP, respectively, as illustrated in Fig. 12. Moreover, it can be found that, for an MCP, given the interference density $\lambda = \lambda_p \bar{c}$, the variance increases when the points are more densely clustered (i.e., with smaller cluster density λ_p and a larger average number of points \bar{c} within each cluster).

Similarly, we have the mean product of $I_o^{(t_1)}$ and $I_u^{(t_2)}$, $t_1 \neq t_2$, as

$$\begin{aligned} &\mathbb{E}\left[I_o^{(t_1)} I_u^{(t_2)}\right] \\ &= \mathbb{E}\left[\sum_{j \in \mathbb{N}} h_j^{(t_1)} \ell_\epsilon(x_j - o) \sum_{i \in \mathbb{N}} h_i^{(t_2)} \ell_\epsilon(x_i - u)\right] \\ &= \mathbb{E}\left[\sum_{j \in \mathbb{N}} h_j^{(t_1)} h_j^{(t_2)} \ell_\epsilon(x_j) \ell_\epsilon(x_j - u) \right. \\ &\quad \left. + \sum_{j, i \in \mathbb{N}}^{j \neq i} h_j^{(t_1)} h_i^{(t_2)} \ell_\epsilon(x_j) \ell_\epsilon(x_i - u)\right] \\ &= \mathbb{E}[h]^2 \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x - u) \rho^{(1)}(x) dx \\ &\quad + \mathbb{E}[h]^2 \int_{\mathbb{R}^2} \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \rho^{(2)}(x, y) dx dy, \end{aligned} \quad (46)$$

which is an integral function of the first and second moment densities. Subsequently, (46) can be obtained by following the derivation of the second moment of interference as in (47).

Finally, by inserting the expectation, second moment, and mean interference product given in (42), (43), and (46), respectively, into (12), we have the spatial-temporal interference correlation coefficient in the following theorem.

Theorem 1. The spatial-temporal correlation coefficient with interferers distributed as MCP, PPP, and β -GPP and path-loss function $\ell_\epsilon(x) = \frac{1}{\epsilon + x^\alpha}$ is given by (48), where $\varpi^{\text{MCP}}(R_d, \bar{c})$

$$\mathbb{E} \left[\left(I_o^{(t_1)} \right)^2 \right] \stackrel{(a)}{=} \begin{cases} \bar{I}_{\text{PPP}}^2 + \frac{\lambda \bar{c}}{\pi^2 R_d^4} \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) A_{R_d}(\|x-y\|) dx \right) dy, & \text{MCP} \\ \bar{I}_{\text{PPP}}^2, & \text{PPP} \\ \bar{I}_{\text{PPP}}^2 - \lambda^2 \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \left(\exp \left(-\frac{\pi \lambda \|x-y\|^2}{\beta} \right) \right) dx \right) dy, & \beta\text{-GPP}. \end{cases} \quad (44)$$

TABLE V: Variance of Interference

Point Process	$\mathbb{E}[I_o^2] - \mathbb{E}[I_o]^2$
MCP	$\delta \pi^2 \lambda (1-\delta) \epsilon^{\delta-2} \csc(\delta \pi) + \frac{\lambda \bar{c}}{\pi^2 R_d^4} \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) A_{R_d}(\ x-y\) dx \right) dy$
PPP	$\delta \pi^2 \lambda (1-\delta) \epsilon^{\delta-2} \csc(\delta \pi)$
β -GPP	$\delta \pi^2 \lambda (1-\delta) \epsilon^{\delta-2} \csc(\delta \pi) - \lambda^2 \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \exp \left(-\pi \lambda \ x-y\ ^2 / \beta \right) dx \right) dy$

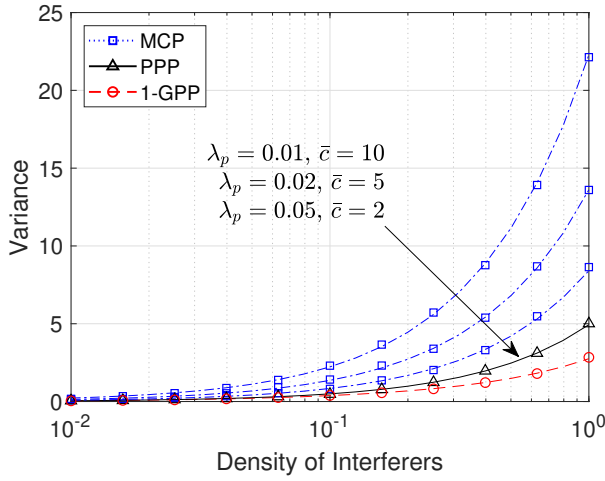


Fig. 12: Interference variance in ad hoc networks with different fields of interferers ($\ell(x) = \frac{1}{1+x^\alpha}$ and $R_d = 1$).

and $\varpi^{\beta\text{-GPP}}(\lambda)$ are given, respectively, by

$$\begin{aligned} & \varpi^{\text{MCP}}(R_d, \bar{c}) \\ &= \frac{\bar{c}}{\pi^2 R_d^4} \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) A_{R_d}(\|x-y\|) dx \right) dy, \end{aligned}$$

and

$$\begin{aligned} & \varpi^{\beta\text{-GPP}}(\lambda) \\ &= \lambda \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \exp \left(-\frac{\pi \lambda \|x-y\|^2}{\beta} \right) dx \right) dy. \end{aligned}$$

Remark 1: The interference correlation coefficient for the MCP is greater than for the PPP, since

$$\zeta^{\text{MCP}}(\|u\|) - \zeta^{\text{PPP}}(\|u\|) = \frac{(C_2 - C_1) \varpi^{\text{MCP}}(R_d, \bar{c})}{C_2(C_2 + \varpi^{\text{MCP}}(R_d, \bar{c}))} \stackrel{(a)}{>} 0,$$

where (a) holds as $\varpi^{\text{MCP}}(R_d, \bar{c}) > 0$ and $C_2 - C_1 \geq 1/2$ [9], and $C_1 = \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx$ and $C_2 = 2\delta \pi^2 (1 - \delta) \epsilon^{\delta-2} \csc(\delta \pi)$.

Remark 2: The interference correlation coefficient for the β -GPP is smaller than for the PPP, since

$$\zeta^{\beta\text{-GPP}}(\|u\|) - \zeta^{\text{PPP}}(\|u\|) = \frac{(C_1 - C_2) \varpi^{\beta\text{-GPP}}(\lambda)}{C_2(C_2 - \varpi^{\beta\text{-GPP}}(\lambda))} \stackrel{(a)}{<} 0,$$

where (a) holds as $C_1 < C_2$ and $C_2 > \varpi^{\beta\text{-GPP}}$ (since $\exp(-\pi \lambda \|x-y\|^2 / \beta) < 1$).

Fig. 13 shows the spatial-temporal correlation coefficient for different fields of interferers, which illustrates the properties discussed above.

2) *Moments of the CSP $_\Phi$ in Ad Hoc Networks:* Next, we derive the moments of the CSP $_\Phi$ in three different random fields of interferers by following the methodology in [79]. For this,

- We start by obtaining the CSP $_\Phi$ by averaging out the randomness of channel gains for a given point process Φ and a link distance.
- We then derive the moments of the CSP $_\Phi$ by averaging over the spatial distributions of the interferers in the different random fields. Specifically,
 - after averaging over the channel gains of the contact link and interfering links based on their distributions, the moments of the CSP $_\Phi$ can be represented by the expectation of the product of a function w.r.t. the locations of the interferers, i.e., $\mathbb{E}[\prod_{x \in \Phi} f(x)]$.
 - the expectation for Matérn cluster and Poisson fields of interferers can be derived based on the PGFLs of MCP and PPP given in (34) and (23), respectively, and that for the β -Ginibre field of interferers can be derived based on the distributions of the distances of the interfering links given in (40).

Theorem 2. The moments of the CSP $_\Phi$ for a Matérn cluster field, Poisson field and β -Ginibre fields of interferers are given by (49), where

$$V_b(x, \theta) = \int_{\mathbb{B}(0, R_d)} \left(\frac{1}{1 + \theta r_t^\alpha \|x-y\|^{-\alpha}} \right)^b dy. \quad (50)$$

Proof. See Appendix A. \square

$$\mathbb{E}[I_o^{(t_1)} I_u^{(t_2)}] = \begin{cases} \lambda \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx + \delta^2 \pi^4 \lambda^2 \epsilon^{2\delta-2} \csc(\delta\pi)^2 + \frac{\lambda \bar{c}}{\pi^2 R_d^4} \\ \quad \times \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) A_{R_d}(\|x-y\|) dx \right) dy, & \text{MCP} \\ \lambda \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx + \delta^2 \pi^4 \lambda^2 \epsilon^{2\delta-2} \csc(\delta\pi)^2, & \text{PPP} \\ \lambda \int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx + \delta^2 \pi^4 \lambda^2 \epsilon^{2\delta-2} \csc(\delta\pi)^2 \\ \quad - \lambda^2 \int_{\mathbb{R}^2} \left(\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(y) \exp\left(-\frac{\pi \lambda \|x-y\|^2}{\beta}\right) dx \right) dy, & \beta\text{-GPP}. \end{cases} \quad (47)$$

$$\zeta(\|u\|) = \begin{cases} \frac{\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx + \varpi^{\text{MCP}}(R_d, \bar{c})}{2\delta\pi^2(1-\delta)\epsilon^{\delta-2} \csc(\delta\pi) + \varpi^{\text{MCP}}(R_d, \bar{c})}, & \text{MCP} \\ \frac{\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx}{2\delta\pi^2(1-\delta)\epsilon^{\delta-2} \csc(\delta\pi)}, & \text{PPP} \\ \frac{\int_{\mathbb{R}^2} \ell_\epsilon(x) \ell_\epsilon(x-u) dx - \varpi^{\beta\text{-GPP}}(\lambda)}{2\delta\pi^2(1-\delta)\epsilon^{\delta-2} \csc(\delta\pi) - \varpi^{\beta\text{-GPP}}(\lambda)}, & \beta\text{-GPP}, \end{cases} \quad (48)$$

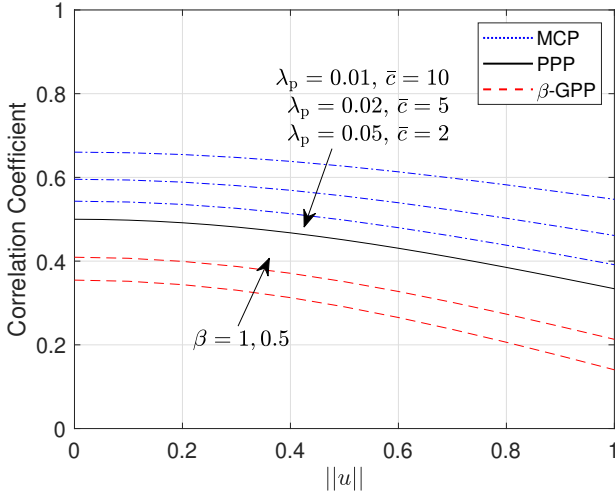


Fig. 13: Correlation coefficient in ad hoc networks with different fields of interferers ($\lambda = 0.1$, $\epsilon = 1$ and $R_d = 1$).

Fig. 14 shows the average success probabilities (i.e., $\mathcal{M}_{P_s}(1)$) in ad hoc networks with different fields of interferers. In order to reveal the entire SIR distribution, we plot the success probabilities in the Möbius homeomorphic (MH) scale. The conversion from linear scale to MH scale is given by the function $x \text{ MH} = \frac{x}{1-x}$ [98], which maps the one-sided infinite support $[0, \infty)$ to the unit interval $[0, 1)$. It can be observed that, in ad hoc networks, spatial repulsion and attraction among the interferers result in lower and higher success probabilities, respectively, compared to the independently located interferers. This can be intuitively understood from the fact that stronger spatial attraction (repulsion) increases the chance that the

interferers are located further away (closer to) the target receiver (as shown in Fig. 11).

Subsequently, we investigate the temporal dependence of successful transmissions by evaluating the temporal CSP, i.e., $\mathcal{M}_{P_s}(2)/\mathcal{M}_{P_s}(1)$, in Fig. 15. It can be found that the CSP increases when the interferers are more clustered (i.e., with smaller R_d or larger \bar{c}) and decreases when the interferers are more scattered (i.e., with larger β). The reason is that spatial attraction and repulsion cause a lower and higher intensity of aggregated interference at the target receiver, and thus a transmission attempt is more likely to succeed given a previous successful transmission.

Furthermore, we investigate the SIR meta distribution for a target link in ad hoc networks. According to the Gil-Pelaez theorem [99], the exact SIR meta distribution can be represented as an integral function of the moments of the CSP_Φ as

$$\bar{F}(\theta, x) = \frac{1}{2} + \frac{1}{\pi} \int_0^\infty \frac{\Im(e^{-ju \log x} M_{ju}(\theta))}{u} du, \quad (51)$$

herein $\Im(z)$ is the imaginary part of z and $j = \sqrt{-1}$ denotes the imaginary unit.

Fig. 16 shows the SIR meta distribution in ad hoc networks with different fields of interferers. Compared with the success probability in Fig. 14, the SIR meta distribution gives the entire distribution of the CSP_Φ . For example, for a 0.5 MH SIR threshold, although the average success probability for PPP is slightly greater than that of β -GPP, the percentage of the links achieving 90% reliability in PPP is higher than twice of that in β -GPP.

3) *Moments of the CSP_Φ in Downlink Cellular Networks:* With Rayleigh fading, the exact moments of the CSP given Φ^M and Φ_β^G can be obtained using the following steps [79]:

$$\mathcal{M}_{P_s}(b) = \begin{cases} \exp\left(-\lambda_p \int_{\mathbb{R}^2} \left(1 - \exp\left(-\bar{c} + \frac{\bar{c}V_b(x, \theta)}{\pi R_d^2}\right)\right) dx\right), & \text{MCP} \\ \exp\left(-\pi\lambda\theta^\delta r_t^2 \frac{\Gamma(1-\delta)\Gamma(b+\delta)}{\Gamma(b)}\right), & \text{PPP} \\ \int_0^\infty e^{-\pi\lambda q/\beta} \left(\frac{\beta}{1+\theta r_t^\alpha q^{-\alpha/2}} + 1 - \beta\right)^b \prod_{j \geq 1} \frac{(\pi\lambda/\beta)^j}{\Gamma(j)} q^{j-1} dq, & \beta\text{-GPP} \end{cases} \quad (49)$$

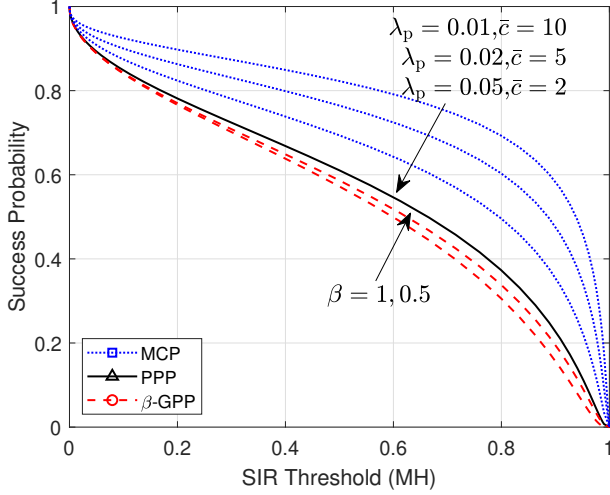


Fig. 14: Success probability in different ad hoc networks ($\lambda = 0.1$, $r_t = 1$, and $R_d = 1$. The curves and markers correspond to the analytical and simulation results, respectively.).

- Deriving the CSP_Φ by averaging over the randomness of the channel gains of the contact link and interfering links based on the PDF of the exponential distribution due to Rayleigh fading;
- Deriving the moments of the CSP_Φ by deconditioning on the spatial distributions of the interferers in the different random fields and contact distance. Specifically,
 - The spatial randomness of the interferers in Matérn cluster and Poisson downlink networks is averaged out based on the reduced PGFL of MCP and PPP, respectively, given in (35) and (23), and that in Ginibre downlink networks is averaged out based on the PDF of the distances of the interfering links given in (40);
 - Based on the nearest-BS associated rule, the contact distances in Matérn cluster, Poisson, and Ginibre downlink networks are averaged out based on their PDFs given in (32), (18), and (40), respectively.

For non-Poisson networks, since deriving the exact downlink success probability is tedious (if not impossible) and the resulting expressions are cumbersome, we use an approximation method, referred to as *Approximate SIR analysis based on the PPP (ASAPPP)* method [47], [100], [101], to simplify the

evaluation of the SIR distribution. ASAPPP provides an approximate SIR distribution in a non-Poisson network, obtained from the SIR distribution in a Poisson network along with a horizontal shift (in the dB scale). This method is based on the insight that when the disparity between the target system model and the Poisson model purely lies in the spatial configuration of the points, shifting the SIR threshold θ of the Poisson model by a coefficient G_0 results in a close approximation of the success probability and the meta distribution of the target system model. The subscript in G_0 indicates that the shift is calculated when θ approaches 0. In other words, the ASAPPP method becomes exact as $\theta \rightarrow 0$. As shown in [84], ASAPPP yields a very good approximation across the whole SIR distribution and is barely susceptible to the fading and path-loss models.

The asymptotic gain G_0 can be obtained by taking the ratio of the mean interference-to-signal ratios (MISRs) of the considered point process to that of the PPP as [83]

$$G_0 = \frac{\text{MISR}_{\text{PPP}}}{\text{MISR}} = \frac{2}{\alpha - 2} \frac{1}{\text{MISR}}, \quad (52)$$

where MISR is defined as

$$\text{MISR} \triangleq \mathbb{E} \left[\frac{\sum_{k=2}^{\infty} \|x_k\|^{-\alpha}}{\|x_1\|^{-\alpha}} \right], \quad (53)$$

and MISR_{PPP} is the MISR in Poisson downlink networks, which can be obtained by utilizing the distance ratio distribution as follows [83].

$$\begin{aligned} \text{MISR}_{\text{PPP}} &= \sum_{j=2}^{\infty} \mathbb{E} \left[\left(\frac{r_1}{r_j} \right)^\alpha \right] \\ &= \sum_{j=2}^{\infty} \mathbb{E} [\varrho_j^\alpha] \\ &\stackrel{(a)}{=} \sum_{j=2}^{\infty} \int_0^1 \varrho^\alpha 2(j-1) \varrho (1-\varrho^2)^{j-2} d\varrho \\ &= \sum_{j=2}^{\infty} \frac{\Gamma(1+\alpha/2)\Gamma(j)}{\Gamma(j+\alpha/2)} \\ &= \frac{2}{\alpha-2}, \end{aligned} \quad (54)$$

where (a) follows the PDF of ϱ_j given in (22).

The numerical value of G_0 can be obtained easily from Monte Carlo simulations for a given network geometry, i.e., λ_p , \bar{c} , and R_d for the MCP and λ and β for the β -GPP. Note

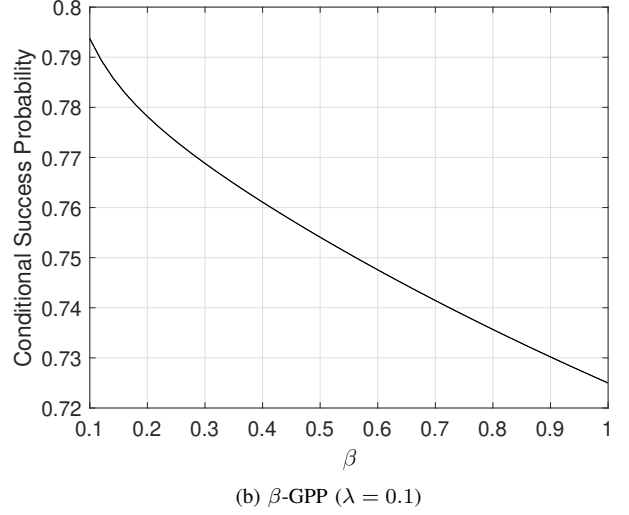
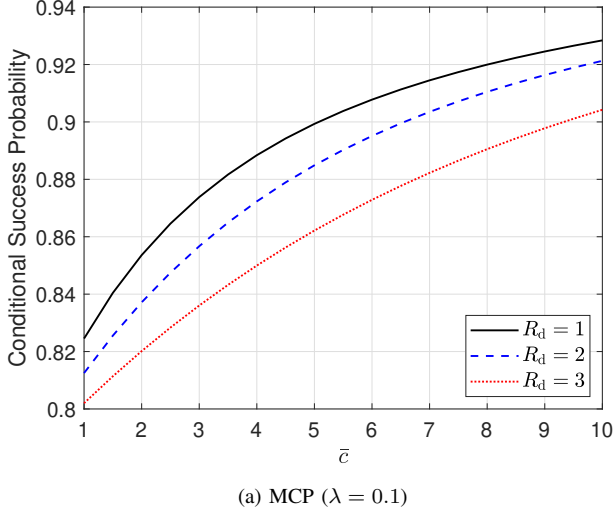


Fig. 15: Temporal CSP in ad hoc networks with non-Poisson fields of interferers ($\alpha = 4$).

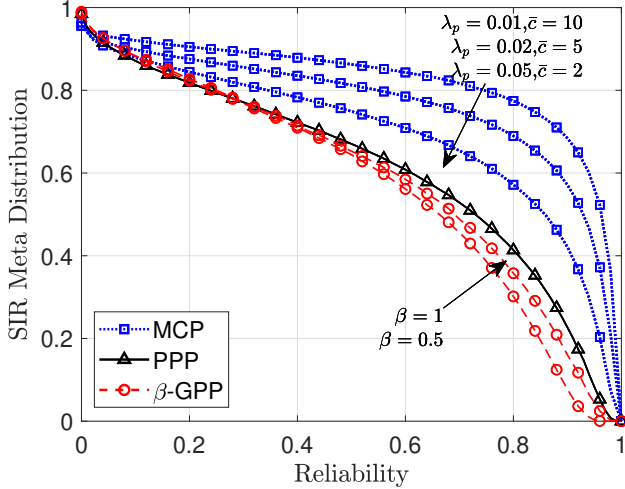


Fig. 16: SIR meta distribution in ad hoc networks with different fields of interferers. ($\theta = 0.5$ MHz, $r_t = 1$, $\lambda = 0.1$, and $R_d = 1$. The curves and markers correspond to the analytical and simulation results, respectively.)

that, as found in [4], the MISR-based gain of the β -GPP can be accurately approximated as

$$G_0 \approx 1 + \beta/2, \quad (55)$$

which is insensitive to the network density and path-loss exponent [84]. Therefore, the success probability of the typical user in a β -GPP network is approximately identical to that in a Poisson network with the SIR threshold scaled from θ to $\theta/(1 + \beta/2)$.

Fig. 17 and Fig. 18, respectively, confirm the effectiveness of the ASAPPP method for approximating the SIR distribution and the SIR meta distribution in MCP and β -GPP downlink networks. As expected, for both types of networks, the ASAPPP method yields a more accurate approximation of SIR distribu-

tion as the SIR threshold decreases. For the MCP model, we can observe that the ASAPPP method is more accurate when Φ^M is less clustered (e.g., with smaller \bar{c} given λ). When the network is more clustered (e.g., when $\bar{c} = 10$), there exists an observable gap between the approximation and the simulation results for both success probability and SIR meta distribution. The reason is that a higher degree of clustering results in a slowly growing rate of the asymptotics [101]. Moreover, for the β -GPP, the insignificant disparity between the simulation and the approximation results can be ascribed to the approximation of the MISR-based gain given in (55).

Next, in Fig. 19(a) and Fig. 19(b), we evaluate the temporal CSPs for MCP and β -GPP downlink networks, respectively. We observe that the CSP decreases slightly when the network becomes more clustered (i.e., with larger \bar{c} and/or smaller R_d with fixed λ) and increases slightly when the network becomes more repulsive (i.e., with larger β). However, the curves are almost flat, i.e., the temporal CSPs in Matérn cluster and Ginibre downlink networks are generally non-sensitive to attractiveness and repulsiveness of the spatial points.

C. Summary and Discussion

We have analyzed the impact of the spatial distribution of transmitters on the interference correlation coefficient and success probability. Specifically, we have presented the derivations of spatial-temporal interference correlation coefficient and success probability and SIR meta distribution of a target link in MCP, PPP, and β -GPP fields of interferers. We have also introduced an approximation of the success probabilities and SIR meta distributions in non-Poisson networks based on the analysis of a Poisson network by scaling the SIR threshold. The main lessons learned are as follows.

- With non-Poisson fields of interferers, the spatial-temporal interference is more and less correlated when the interferers are distributed more attractively and repulsively, respectively.

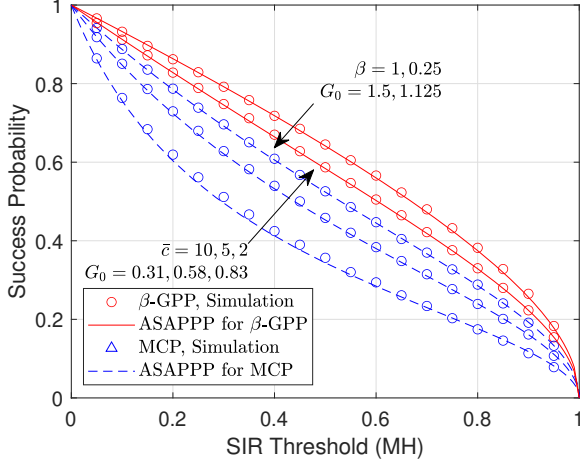


Fig. 17: The ASAPPP approximation of non-Poisson cellular networks ($\alpha = 4$, $\lambda = 0.1$, $R_d = 4$).

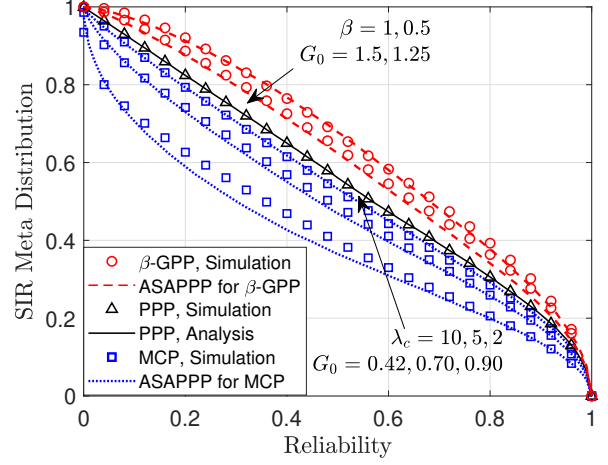


Fig. 18: The ASAPPP approximation of SIR meta distribution in non-Poisson cellular networks ($\theta = 0.5$ MH, $\alpha = 4$, $\lambda = 0.1$, $R_d = 5$).

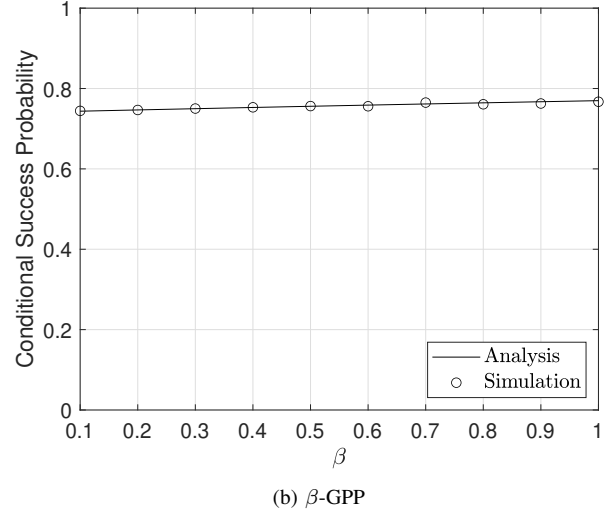
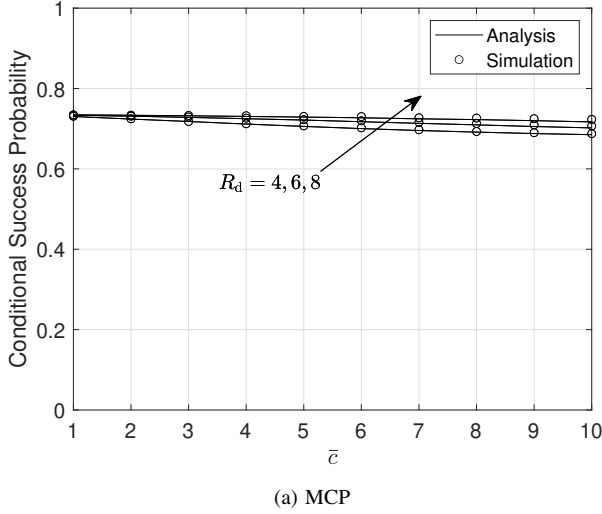


Fig. 19: Temporal CSP in non-Poisson cellular networks ($\alpha = 4$, $\lambda = 0.1$).

- Compared with a Poisson field of interferers, the Matérn cluster and β -Ginibre field of interferers increases and decreases the success probability of a target link, respectively. Moreover, the successful transmission events are more and less temporally-correlated when the locations of the interferers are more clustered and scattered, respectively.
- Compared with a Poisson downlink network, the Matérn cluster and Ginibre downlink networks render a lower and a higher success probability, respectively. Additionally, the successful transmission events are less and more temporally-correlated when the interferers are more attractive and repulsive, respectively.

Open Technical Issues: The point process models presented

in Section III have been adopted to characterize the general spatial features (e.g., independence, repulsion and clustering) of wireless networks. In addition to these generic models, some point process models have been established for some specialized network scenarios. For example, reference [102] proposes an *energized point process* to model the spatial distribution of wireless-powered devices. The model exhibits spatial correlation of the RF-powered nodes that can harvest sufficient energy from a Poisson field of RF sources. Besides, reference [103] introduces a *Poisson rain process* to model cellular networks with spatial-temporal traffic. Specifically, the model employs a space-time PPP to model traffic arrives which are then assigned to PPP-distributed BSs based on different allocation schemes. Future efforts should be dedicated to developing point processes that incorporate spatial-temporal

features (e.g., fading, shadowing, blockage, and renewable energy arrivals) for more specific system models.

IV. LOCATION-DEPENDENT ANALYSIS OF CELLULAR MODELS

The locations of the users play a crucial part in their performance. Conditioning the regions (e.g., cell center and boundary) of the user of interest allows a location-dependent analysis [104], [105] for a location-specific user (LSU) in cellular networks. The objective of this section is to demonstrate the effect of the location of the target user on the SIR distribution. For this, we derive the moments of the CSP_Φ and the SIR gain for an LSU in different regions.

A. System Models

For a stationary point process $\Phi \subset \mathbb{R}^2$ of BSs, let $x_j(u) \in \Phi$ represent the location of the j -th nearest BS to the LSU at u . Under the nearest-BS association principle, $x_1(u)$ and $x_2(u)$ are the locations of the associated (serving) BS and the nearest interfering BS to the LSU, respectively. The region of a user's location can be defined by the relationship between the link distances to the associated BS and to the nearest interfering BS.

For $\rho \in [0, 1]$, the regions of *cell-center user* and *cell-boundary user* are defined, respectively, as

$$\begin{aligned}\mathcal{R}_c &\triangleq \{u \in \mathbb{R}^2 : \|x_1(u) - u\| \leq \rho \|x_2(u) - u\|\}, \\ \mathcal{R}_b &\triangleq \{u \in \mathbb{R}^2 : \|x_1(u) - u\| > \rho \|x_2(u) - u\|\}.\end{aligned}$$

The area fraction of a region can be determined by the probability that an arbitrary location, say o , falls into that region, which solely depends on ρ . For the analysis, we focus on the case where Φ is a PPP.

The area fractions of \mathcal{R}_c and \mathcal{R}_b can be obtained based on the CDF of ϱ_2 (given in (21)) as

$$\mathbb{P}[o \in \mathcal{R}_c] = \mathbb{P}[\varrho_2 \leq \rho] = F_{\varrho_2}(\rho) = \rho^2$$

and

$$\mathbb{P}[o \in \mathcal{R}_b] = 1 - \mathbb{P}[o \in \mathcal{R}_c] = 1 - \rho^2.$$

Moreover, in the special cases when a cell-boundary user has two or three equidistant closest BSs, it is referred to as an *edge user* or a *vertex user*, respectively. The regions of the edge users (one-dimensional) and vertex users (zero-dimensional) are defined, respectively, as

$$\begin{aligned}\mathcal{R}_e &= \{u \in \mathbb{R}^2 : \|x_1(u) - u\| = \|x_2(u) - u\|\}, \\ \mathcal{R}_v &= \{u \in \mathbb{R}^2 : \|x_1(u) - u\| = \|x_2(u) - u\| \\ &\quad = \|x_3(u) - u\|\}.\end{aligned}\quad (56)$$

Note that in a two-dimensional homogeneous PPP there exists no location with more than three nearest BSs almost surely [89].

In the following, we explore the performance at five different types of typical users: (i) The standard typical user, whose performance represents the average of all users; it is

referred to in this section as the typical general user; (ii) the typical cell-center user, whose performance represents the average of all cell-center users; (iii) the typical cell-boundary user, whose performance represents the average of all cell-boundary users; (iv) the typical cell-edge user whose performance represents the average of all locations on the cell edges; and (v) the typical vertex user, whose performance represents the average of all vertex points of the cells.

In Poisson cellular networks, the PDF of the contact distance of the typical vertex user is given by [106], [107]

$$f_{r_1}^v(r) = 2(\lambda\pi)^2 r^3 \exp(-\lambda\pi r^2). \quad (57)$$

The success probability of the typical cell-center and cell-boundary users can be defined, respectively, as

$$\begin{aligned}\bar{F}_\eta^c(\theta) &= \mathbb{P}[\eta > \theta \mid o \in \mathcal{R}_c], \\ \text{and } \bar{F}_\eta^b(\theta) &= \mathbb{P}[\eta > \theta \mid o \in \mathcal{R}_b].\end{aligned}\quad (58)$$

Fig. 20 shows a realization of network partitions of a Poisson network with unit intensity, where the locations of the BSs are represented by the black circles. White and cyan regions represent the cell-center regions \mathcal{R}_c and cell-boundary regions \mathcal{R}_b , respectively. The blue triangle markers represent the region of the vertex users and blue lines represent the region of the edge users.

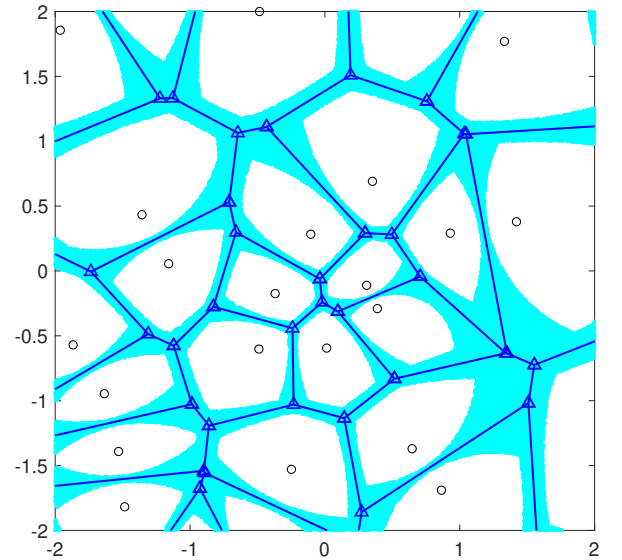


Fig. 20: Illustration of network partitions in a two-dimensional Poisson network for $\rho = 0.8$.

B. Performance Analysis

1) *Cell-Center and Cell-Boundary User:* We start by showing how to obtain the b -th moments of the CSP_Φ for cell-center and cell-boundary users, denoted as $\mathcal{M}_{P_s}^c(b)$ and $\mathcal{M}_{P_s}^b(b)$, respectively. The methodology of obtaining the former is to

derive the SIR meta distribution for the typical general user by conditioning its location to be in the cell-center and cell-boundary regions [104], [105]. Such conditioning restricts the distribution regions of interferers which will be reflected in the lower bounds of the integrals after applying the PGFL of PPP.

Based on this approach, we obtain $\mathcal{M}_{P_s}^c(b)$ in the following theorem.

Theorem 3. The moments of the CSP_Φ for the typical cell-center user, i.e., the typical general user conditioned on \mathcal{R}_c , in a Poisson downlink network with Rayleigh fading are

$$\mathcal{M}_{P_s}^c(b) = \frac{1}{{}_2F_1(b, -\delta; 1 - \delta; -\rho^\alpha \theta)}, \quad (59)$$

where $\delta = 2/\alpha$.

Proof. See Appendix B. \square

Remark 3: The SIR gain of the typical cell-center user compared to the typical general user is $\rho^{-\alpha}$, which is easily obtained by comparing (59) with the moments of the CSP_Φ for the typical general user in a Poisson downlink network given as [79]:

$$\mathcal{M}_{P_s}(b) = \frac{1}{{}_2F_1(b, -\delta; 1 - \delta; -\theta)}. \quad (60)$$

Remark 4: A fraction $x = \rho^2$ of users who have the lowest distance ratio of the associated BS to the interfering BS has an SIR gain of $-5\alpha \log_{10} x$ dB.

Applying the law of total probability, the moments of the CSP_Φ for the typical general user can be expressed as follows:

$$\begin{aligned} \mathcal{M}_{P_s}(b) &= \mathbb{E}[\mathbb{P}[\eta > \theta]^b] \\ &= \mathbb{E}[\mathbb{P}[\eta > \theta]^b \mid o \in \mathcal{R}_c] \mathbb{P}[o \in \mathcal{R}_c] \\ &\quad + \mathbb{E}[\mathbb{P}[\eta > \theta]^b \mid o \in \mathcal{R}_b] \mathbb{P}[o \in \mathcal{R}_b] \\ &= \mathcal{M}_{P_s}^c(b) \mathbb{P}[o \in \mathcal{R}_c] + \mathcal{M}_{P_s}^b(b) (1 - \mathbb{P}[o \in \mathcal{R}_c]). \end{aligned}$$

Then, the moments of the CSP_Φ for the typical cell-boundary user are given by

$$\begin{aligned} \mathcal{M}_{P_s}^b(b) &= \frac{\mathcal{M}_{P_s}(b) - \mathcal{M}_{P_s}^c(b) \mathbb{P}[o \in \mathcal{R}_c]}{1 - \mathbb{P}[o \in \mathcal{R}_c]} \\ &= \frac{\mathcal{M}_{P_s}(b) - \rho^2 \mathcal{M}_{P_s}^c(b)}{1 - \rho^2}. \end{aligned} \quad (61)$$

We then have the following corollary by using (60) and (59) in (62).

Corollary 1. The moments of the CSP_Φ of the typical cell-boundary user, i.e., $o \in \mathcal{R}_b$, in a Poisson downlink network with Rayleigh fading are

$$\begin{aligned} \mathcal{M}_{P_s}^b(b) &= \frac{1}{(1 - \rho^2) {}_2F_1(b, -\delta; 1 - \delta; -\theta)} \\ &\quad - \frac{\rho^2}{(1 - \rho^2) {}_2F_1(b, -\delta; 1 - \delta; -\rho^\alpha \theta)}, \end{aligned} \quad (62)$$

where $\delta = 2/\alpha$.

2) *Edge User:* An edge user can be considered as an asymptotic case of a cell-boundary user when $\rho \rightarrow 1$, i.e., $\mathbb{P}[o \in \mathcal{R}_b] \rightarrow 0$. In this case, the cell-boundary user is equidistant from the serving BS and the nearest interferer, i.e., on the cell edge. Therefore, the performance of the typical cell-edge user can be obtained from the asymptotics of the typical cell-boundary user as $\rho \rightarrow 1$.

By taking the limit $\rho \rightarrow 1$ of $\mathcal{M}_{P_s}^b(b)$ in (62), we have the moments of the CSP_Φ for the typical edge user as follows:

$$\begin{aligned} \mathcal{M}_{P_s}^e(b) &\stackrel{(a)}{=} \lim_{\rho \rightarrow 1} \mathcal{M}_{P_s}^b(b) \\ &= \frac{1}{{}_2F_1(b, -\delta; 1 - \delta; -\theta)} \\ &\quad - \frac{b\theta {}_2F_1(b+1, 1 - \delta; 2 - \delta; -\theta)}{(1 - \delta) {}_2F_1(-b, -\delta; 1 - \delta; -\theta)^2}, \end{aligned} \quad (63)$$

where (a) follows from L'Hôpital's rule.

After some mathematical manipulations, we have $\mathcal{M}_{P_s}^e(b)$ in the following corollary.

Corollary 2. With Rayleigh fading, the moments of the CSP_Φ for the typical edge user in a Poisson downlink network are

$$\mathcal{M}_{P_s}^e(b) = \frac{1}{(1 + \theta)^b {}_2F_1(b, -\delta; 1 - \delta; -\theta)^2}, \quad b \in \mathbb{C} \quad (64)$$

where $\delta = 2/\alpha$.

Remark 5: Compared with $\mathcal{M}_{P_s}(b)$ in (59), $\mathcal{M}_{P_s}^e(b)$ in (64) can be expressed as $\mathcal{M}_{P_s}^e(b) = \frac{\mathcal{M}_{P_s}(b)^2}{(1 + \theta)^b}$, $\forall b \in \mathbb{R}^+$. $\mathcal{M}_{P_s}^e(b)$ is smaller than $\mathcal{M}_{P_s}(b)$ and their gap depends only on θ and b .

3) *Vertex User:* Finally, we compute the performance of the typical vertex user, which represents the worst-case performance. To obtain the moments of the CSP_Φ , denoted as $\mathcal{M}_{P_s}^v(b)$, we follow the derivation steps similar to those for the typical general user with the condition that the user is equidistant from the serving BS and the two nearest interfering BSs [105]. Under this condition, the contact distance is averaged out based on its PDF given in (57). With this methodology, we have $\mathcal{M}_{P_s}^v(b)$ given in the following theorem.

Theorem 4. With Rayleigh fading, the moments of the CSP_Φ for the typical vertex user in a Poisson downlink network are

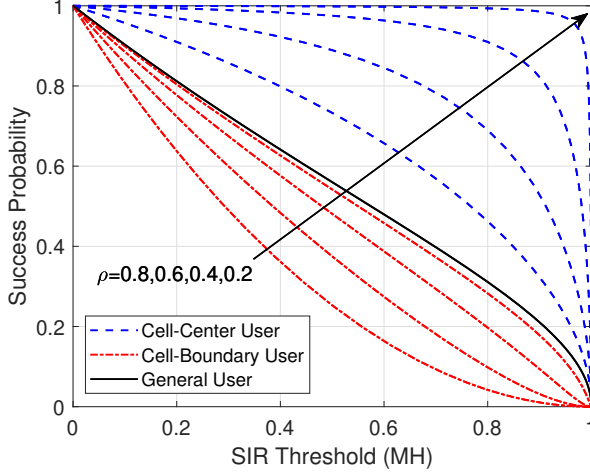
$$\mathcal{M}_{P_s}(b) = \frac{1}{(1 + \theta)^{2b} {}_2F_1(b, -\delta; 1 - \delta; -\theta)^2}, \quad (65)$$

where $\delta = 2/\alpha$.

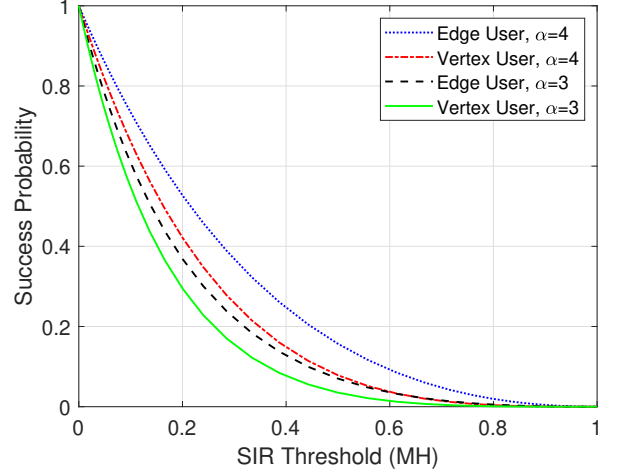
Proof. See Appendix C. \square

Remark 6: Compared with (117), the success probability of the typical vertex user in (65) can be expressed as $\mathcal{M}_{P_s}^v(b) = \frac{\mathcal{M}_{P_s}^c(b)}{(1 + \theta)^b}$. $\mathcal{M}_{P_s}^v(b)$ is smaller than $\mathcal{M}_{P_s}^c(b)$ due to the two equidistant closest interfering BSs.

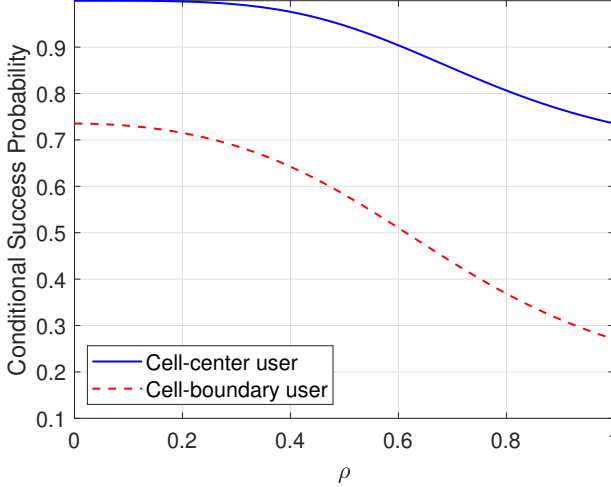
Fig. 21 depicts the average success probabilities (i.e., $\mathcal{M}_{P_s}(1)$) of the LSUs. It is straightforward that the success probability monotonically decreases with ρ . The weighted mean of the success probabilities for the typical cell-center user and the typical cell-boundary user with the same value of ρ are equal to that of the typical general user.



(a) Typical general user and cell-center/boundary user



(b) Typical edge user and vertex user

Fig. 21: Success probability of LSUs in Poisson downlink networks ($\lambda = 1$, $\alpha = 4$).**Fig. 22:** The temporal CSP of LSUs.

4) *Temporal Effect of Location Dependence:* Next, we investigate how location dependence affects the temporal correlation among different successful transmission events by evaluating the CSP $\frac{\mathcal{M}_{F_s}(2)}{\mathcal{M}_{F_s}(1)}$. Fig. 22 shows how the CSP varies with ρ . It is found that the CSP for the typical cell-center user is greater than the typical cell-boundary user. Moreover, for both typical cell-center and cell-boundary users, their CSPs decrease with ρ . The reason is that the successful transmission events are more temporally correlated when the received signal dominates the interference. Such temporal correlation decreases when the interference becomes stronger.

5) *SIR Gain:* Next, we investigate the asymptotic SIR gain to illustrate the SIR improvement over the typical general user due to location dependence. The MISR of the typical cell-center

user can be derived as

$$\begin{aligned} \text{MISR}_{\mathcal{R}_c} &= \sum_{i=2}^{\infty} \mathbb{E} \left[\left(\frac{r_1}{r_i} \right)^{\alpha} \mid o \in \mathcal{R}_c \right] \\ &\stackrel{(b)}{=} \mathbb{E} \left[\left(\frac{r_1}{r_2} \right)^{\alpha} \mid o \in \mathcal{R}_c \right] \sum_{i=2}^{\infty} \mathbb{E} \left[\left(\frac{r_2}{r_i} \right)^{\alpha} \right], \\ &= \frac{\mathbb{E} \left[\left(\frac{r_1}{r_2} \right)^{\alpha}, o \in \mathcal{R}_c \right]}{\rho^2} \sum_{i=2}^{\infty} \mathbb{E} \left[\left(\frac{r_2}{r_i} \right)^{\alpha} \right], \end{aligned} \quad (66)$$

where (b) holds as the condition that the typical general user is in the cell-center region, i.e., $o \in \mathcal{R}_c$, only affects the relative distance between r_1 and r_2 .

The first expectation in (66) can be derived based on the joint distribution of r_1 and r_2 as

$$\mathbb{E} \left[\left(\frac{r_1}{r_2} \right)^{\alpha}, o \in \mathcal{R}_c \right] = \int_0^{\rho} f_{v_2}(v) v^{\alpha} dv = \frac{2\rho^{2+\alpha}}{2+\alpha}. \quad (67)$$

The second expectation in (66) can be obtained by exploiting the distribution of the distance ratios of a PPP as

$$\begin{aligned} \sum_{j=2}^{\infty} \mathbb{E} \left[\left(\frac{r_2}{r_j} \right)^{\alpha} \right] &= \frac{\sum_{j=2}^{\infty} \mathbb{E} \left[\left(\frac{r_1}{r_j} \right)^{\alpha} \right]}{\mathbb{E} \left[\left(\frac{r_1}{r_2} \right)^{\alpha} \right]} \\ &\stackrel{(c)}{=} \frac{2+\alpha}{2} \sum_{j=2}^{\infty} \int_0^1 v_i^{\alpha} f_i(v) dv \\ &= \frac{2+\alpha}{2} \int_0^1 v^{\alpha} 2v^{-3} dv \\ &= \frac{\alpha+2}{\alpha-2}, \end{aligned} \quad (68)$$

where (c) follows the distribution of q_2 .

Plugging (67) and (68) into (66) yields

$$\text{MISR}_{\mathcal{R}_c} = \frac{2\rho^{\alpha}}{\alpha-2}. \quad (69)$$

Subsequently, with MISR_{PPP} and $\text{MISR}_{\mathcal{R}_c}$ given in (54) and (69), respectively, the SIR gain of the typical cell-center user can be obtained as

$$G_{\mathcal{R}_c} = \frac{\text{MISR}_{\text{PPP}}}{\text{MISR}_{\mathcal{R}_c}} = \rho^{-\alpha}. \quad (70)$$

Similarly, the SIR gains of different LSUs can be obtained as in Table VI.

Remark 7: Since $\alpha > 2$ and $\rho \in [0, 1]$, the SIR gain of the typical cell-center user is greater than or equal to one, and the equality holds when $\rho = 1$. Moreover, the SIR gain of the typical cell-boundary user is smaller than or equal to one and the equality holds when $\rho = 0$. It can be found that the influence of ρ on the SIR gain of the typical cell-center user is much more significant than that of the typical cell-boundary user. Moreover, the gap between the SIR gains of the typical edge user and vertex user increases with α .

C. Summary and Discussion

We have discussed the impact of the relative distance of a random user on its performance in Poisson downlink networks. Based on the relationship between the contact distance and the distance to the two nearest interferers, a user can be categorized as cell-center, cell-boundary, cell-edge, and vertex user. We present the derivations of the moments of the CSP_Φ for the typical cell-center, cell-boundary, cell-edge and vertex users in two-dimensional Poisson downlink networks. To show the direct impact of the location on the SIR, we also derive the SIR gain of the four types of users. The major lessons learned are as follows.

- The gaps among the b -th moments of the CSP_Φ for the typical cell-center, cell-edge and vertex users depend only on b and the SIR threshold θ , i.e., $\mathcal{M}_{P_s}^c(b) = (1 + \theta)^b \mathcal{M}_{P_s}^e(b) = (1 + \theta)^{2b} \mathcal{M}_{P_s}^v(b)$, $\forall b \in \mathbb{R}^+$.
- The successful transmission events are more temporally correlated for the typical cell-center user than the typical cell-boundary user.
- The SIR gains for the typical cell-center and cell-boundary user depend on the path-loss exponent and location dependence coefficient ρ while those for the typical cell-edge and vertex user only depend on the path-loss exponent.

Open Technical Issues: In the literature, a location-dependent analysis has been carried out [105], and BS cooperation schemes have been designed based on the location-dependent modeling [104]. However, the location-dependent performance for Poisson uplink networks remains unexplored. Characterizing the location-dependent uplink user is more challenging as the uplink model is not fully tractable [65]. Another interesting direction is to analyze the performance of LSUs in non-Poisson cellular networks, e.g., when the transmitters exhibit repulsive or attractive distribution.

Additionally, cell-free communication systems where densely deployed BSs could simultaneously serve a number of users have emerged as a practical solution for future-generation communication systems. In such an infrastructure, the BS

cooperation schemes need to take into account the user locations. For example, it is more meaningful to assign more resources for cell-boundary users to improve their success probabilities. Thus, location-dependent analysis is the key to the design of BS cooperation schemes in cell-free massive MIMO systems.

V. SPATIALLY CORRELATED AND INDEPENDENT SHADOWING

The main objective of this section is to investigate the effect of correlated and independent shadowing in large-scale systems based on stochastic geometry analysis.

A. System Model

To model the effect of shadowing, we consider that the whole plane is partitioned into a set of deterministic shadowing cells $\mathbf{S} = \{\mathbb{S}_k \subset \mathbb{R}^2\}_{k \in \mathbb{N}}$ in which the transmitters experience similar shadowing. We consider an ad hoc network with a Poisson field of interferers Φ introduced in Section III-A as an example system model for the analysis in this section.

The channel gain from $x_j \in \Phi$ to o is $h_j S_j \ell(\|x_j\|)$, where h_j and S_j are the fading coefficient and shadowing coefficient, respectively, from x_j to o , and $\ell(\|x_j\|) = \|x_j\|^{-\alpha}$ is the path-loss function. With a Poisson field of interferers, the aggregated interference at the receiver of interest is

$$I_o = \sum_{j \in \mathbb{N}} h_j S_j \ell(\|x_j\|), \quad (71)$$

and the SIR follows as

$$\eta = \frac{h_t S_t \ell(\|x_t\|)}{I_o}, \quad (72)$$

where S_t denotes the shadowing coefficient of the link under consideration.

Let $c(j)$ be the cell index that x_j resides in, i.e., $c(j) = k$ if $x_j \in \mathbb{S}_k$. We consider the extreme cases of (fully) correlated and independent intra-cell shadowing. In independent shadowing, the $S_{c(j)}$ are all independent and distributed with CDF $F_{c(j)}$. In correlated shadowing, $S_{c(j)}$ and $S_{c(i)}$ are independent only if $c(j) \neq c(i)$, and $S_{c(j)} = T_k$ for all $x_j \in \mathbb{S}_k$. By choosing F_k , different shadowing properties can be assigned to the individual cells.

B. Performance Analysis

We first characterize the Laplace transform of the interference. Then, based on it, we derive the mean and variance of the interference. Furthermore, we obtain the moments of the CSP_Φ under both correlated and independent shadowing.

1) *Laplace Transform of the Interference:* To analyze the interference distribution, we start by characterizing the Laplace transform of the aggregated interference at the target receiver at the origin with correlated and independent shadowing, denoted by I_o^{Cor} and I_o^{Ind} , respectively, by following the methodology in [108]. Similar to the derivation steps of the moments of the CSP given a Matérn cluster field of interferers presented

TABLE VI: MISR and SIR gain of location-specific users

User Location	MISR	SIR Gain
Typical general user	$2/(\alpha - 2)$	1
Typical cell-center user	$2\rho^\alpha/(\alpha - 2)$	$\rho^{-\alpha}$
Typical cell-boundary user	$2(1 - \rho^{\alpha+2})/(\alpha - 2)(1 - \rho^2)$	$(1 - \rho^2)/(1 - \rho^{\alpha+2})$
Typical edge user	$(\alpha + 2)/(\alpha - 2)$	$2/(\alpha + 2)$
Typical vertex user	$2\alpha/(\alpha - 2)$	$1/\alpha$

in **Appendix A**, we first express the Laplace transform as the expectation of the product of a function of the locations of the interferers given the distribution of the shadowing coefficients. Then, by the PGFLs of MCP given in (34), we convert the expectation into an integral expression as a function of the shadowing coefficients. Subsequently, the randomness of the shadowing can be averaged out by conditioning on that the daughter points in each cluster are associated with the same and independent shadowing coefficients with correlated shadowing and independent shadowing, respectively.

Based on this methodology, the Laplace transforms of I_o^{Cor} and I_o^{Ind} are given in the following theorem.

Theorem 5. In a Rayleigh fading environment with a Poisson field of interferers, the conditional Laplace transform of the interference at the target receiver is

$$\mathcal{L}_{I_o^{\text{Cor}}}(s) = \prod_{k \in \mathbb{N}} \mathbb{E}_{T_k} \left[\exp \left(-\lambda \int_{\mathbb{S}_k} \left(1 - \frac{1}{1 + s\ell(\|x\|)T_k} \right) dx \right) \right], \quad (73)$$

with correlated shadowing, and

$$\mathcal{L}_{I_o^{\text{Ind}}}(s) = \prod_{k \in \mathbb{N}} \exp \left(-\lambda \int_{\mathbb{S}_k} \left(1 - \mathbb{E}_{T_k} \left[\frac{1}{1 + s\ell(\|x\|)T_k} \right] \right) dx \right), \quad (74)$$

with independent shadowing.

Proof. See **Appendix D**. \square

According to Jensen's inequality, i.e., the convex transformation of an expectation equals or exceeds the expectation over the convex transformation, we have the inequality in (75). Furthermore, since $\exp(-x)$ is a completely monotone function, we readily obtain the following observation by comparing (73) and (74).

Remark 8: For all $s > 0$, $\mathcal{L}_{I_o^{\text{Cor}}}(s) > \mathcal{L}_{I_o^{\text{Ind}}}(s)$.

2) *Mean and Variance of Interference:* Due to the singularity of the path-loss function $\ell(x)$, we adopt $\ell_\epsilon(x) = \frac{1}{\epsilon + \|x\|^\alpha}$, $\alpha > 2$, $\epsilon > 0$, to evaluate the mean and variance of interference, similar to Sec. III-B1. As the mean of a random variable can be derived by taking the first-order derivative of its Laplace transform w.r.t. s and setting $s = 0$, i.e., $\mathbb{E}[X] = -\frac{d\mathcal{L}_X(s)}{ds}|_{s=0}$, we can easily obtain the mean interference as

$$\mathbb{E}[I_o^{\text{Ind}}] = \mathbb{E}[I_o^{\text{Cor}}] = \lambda \sum_{k \in \mathbb{N}} \mathbb{E}[T_k] \int_{\mathbb{S}_k} \frac{1}{\epsilon + \|x\|^\alpha} dx,$$

which shows that the mean interference at the typical receiver with independent shadowing and correlated shadowing are identical, i.e., $I_o^{\text{Ind}} = I_o^{\text{Cor}}$.

The same result can also be obtained from Campbell's Theorem [3, Thm. 4.1], which shows that it holds for all stationary point processes with an arbitrary fading model.

As the second moment of a random variable can be derived by evaluating the second derivative of its Laplace transform w.r.t. s at $s = 0$, the variance of I_o^{Ind} can be obtained based on the Laplace transform as [108]

$$\begin{aligned} \mathbb{V}[I_o^{\text{Ind}}] &= \mathbb{E}[(I_o^{\text{Ind}})^2] - \mathbb{E}[I_o^{\text{Ind}}]^2 \\ &= \frac{d^2 \mathcal{L}_{I_o^{\text{Ind}}}(s)}{d^2 s^2} \Big|_{s=0} - \left(\frac{d \mathcal{L}_{I_o^{\text{Ind}}}(s)}{ds} \Big|_{s=0} \right)^2 \\ &= 2\lambda \sum_{k \in \mathbb{N}} \mathbb{E}[T_k^2] \int_{\mathbb{S}_k} \frac{1}{(\epsilon + \|x\|^\alpha)^2} dx \\ &\quad + \lambda^2 \sum_{k \in \mathbb{N}} \mathbb{E}[T_k]^2 \left(\int_{\mathbb{S}_k} \frac{1}{\epsilon + \|x\|^\alpha} dx \right)^2. \end{aligned} \quad (76)$$

In the same manner, the variance of I_o^{Cor} can be derived as

$$\begin{aligned} \mathbb{V}[I_o^{\text{Cor}}] &= 2\lambda \sum_{k \in \mathbb{N}} \int_{\mathbb{S}_k} \frac{\mathbb{E}[T_k^2]}{(\epsilon + \|x\|^\alpha)^2} dx \\ &\quad + \lambda^2 \sum_{k \in \mathbb{N}} \mathbb{E}[T_k]^2 \left(\int_{\mathbb{S}_k} \frac{1}{\epsilon + \|x\|^\alpha} dx \right)^2 \\ &\quad + \lambda^2 \sum_{k \in \mathbb{N}} \mathbb{V}[T_k] \left(\int_{\mathbb{S}_k} \frac{1}{\epsilon + \|x\|^\alpha} dx \right)^2 \\ &\stackrel{(a)}{=} \underbrace{\mathbb{V}[I_o^{\text{Ind}}] + \lambda^2 \sum_{k \in \mathbb{N}} \mathbb{V}[T_k]^2 \left(\int_{\mathbb{S}_k} \frac{1}{\epsilon + \|x\|^\alpha} dx \right)^2}_{>0}, \end{aligned} \quad (77)$$

where (a) follows from (76). Comparing (76) and (77) yields the following observation.

Remark 9: The variance of the interference at the typical receiver with correlated shadowing is greater than that with independent shadowing, i.e., $\mathbb{V}[I_o^{\text{Cor}}] > \mathbb{V}[I_o^{\text{Ind}}]$.

3) *Moments of the CSP Φ :* Given the point process Φ , the CSP is

$$\begin{aligned} \mathbb{P}[\eta > \theta \mid \Phi] &= \mathbb{P} \left[\frac{h_t S_t r_t^{-\alpha}}{I_o} > \theta \mid \Phi \right] \\ &\stackrel{(a)}{=} \mathbb{E}_{S_t, (S_j)} \left[\prod_{j \in \mathbb{N}} \left(1 + \frac{\theta r_t^\alpha r_j^{-\alpha} S_j}{S_t} \right)^{-1} \right], \end{aligned}$$

$$\begin{aligned} \mathbb{E}_{T_k} \left[\exp \left(-\bar{c} \int_{\mathbb{S}_k} \left(1 - \frac{1}{1 + s\ell(\|x\|)T_k} \right) dx \right) \right] \\ > \exp \left(-\bar{c} \int_{\mathbb{S}_k} \left(1 - \mathbb{E}_{T_k} \left[\frac{1}{1 + s\ell(\|x\|)T_k} \right] \right) dx \right). \end{aligned} \quad (75)$$

where (a) follows the derivations of (109) in **Appendix A**.

Then, the moments of the CSP_Φ can be represented as

$$\mathcal{M}_{P_s}(b) = \mathbb{E} \left[\prod_{j \in \mathbb{N}} \left(1 + \frac{\theta r_t^\alpha r_j^{-\alpha} S_j}{S_t} \right)^{-b} \right].$$

By following the same methodology to the proof of Theorem 5, we have $\mathcal{M}_{P_s}(b)$ derived in the following theorem.

Theorem 6. In a Rayleigh fading environment, the conditional moments of the CSP_Φ under correlated shadowing and independent shadowing are given, respectively, by (78) and (79), shown on the top of the next page.

Remark 10: It follows from **Remark 8** that with Rayleigh fading, the moments of the CSP_Φ under correlated shadowing are greater than that with independent shadowing.

Next, we numerically evaluate the network performance under correlated and independent shadowing assumptions. We assume $T_k = \kappa^{N_k(x)}$ [108], where κ is the attenuation factor accounting for the signal loss while penetrating a blockage (e.g., a wall) and $N_k(x)$ represents the number of blockages between x and the receiver of interest. $N_k(x)$ is assumed to be a Poisson random variable for each link [109], i.e., $N_k(x) \sim \mathcal{P}(\lambda_b \|x\|)$, where λ_b denotes the blockage density. With correlated shadowing, the transmitters in the same cell \mathbb{S}_k are associated with the same shadowing coefficient $T_k = \kappa^{N_k(d_{k,o})}$, $N_k(d_{k,o}) \sim \mathcal{P}(\lambda_b d_{k,o})$, where $d_{k,o}$ represents the distance between the center of \mathbb{S}_k and the origin. Differently, with independent shadowing, the transmitters at $x_j \in \mathbb{S}_k$ are associated with i.i.d. shadowing coefficients $S_j = \kappa^{N_k(d_{k,o})}$, $N_k(d_{k,o}) \sim \mathcal{P}(\lambda_b d_{k,o})$. Besides, we assume that the link between the target receiver and the serving transmitter is subject to no shadowing, i.e., $S_t = 1$. For the spatial configuration of the shadowing cells, we consider the example in Fig. 23, where $\mathbf{S} = \{\mathbb{S}_k\}_{k \in \mathbb{N}}$ is a set of squares with length L . It is worth noting that the analytical framework can be extended to other types of point processes by changing the distribution of points as long as $|\mathbb{S}_k| < \infty$, i.e., the point process has a finite number of points in each shadowing cell almost surely.

Figure 24 illustrates the success probability, i.e., $\mathcal{M}_{P_s}^{\text{Cor}}(1)$ and $\mathcal{M}_{P_s}^{\text{Ind}}(1)$, as a function of the SIR threshold. It can be observed that the success probability with correlated shadowing is greater than that with independent shadowing, which agrees with **Remark 10**. Moreover, the gap between the success probabilities with correlated and independent shadowing decreases as the cell size (i.e., L^2) becomes smaller, as in this case the signals from the interferers experience nearly independent shadowing even with the correlated shadowing model.

Furthermore, to evaluate the temporal effect of shadowing, we evaluate the temporal CSP with both correlated and independent fading, i.e., $\frac{\mathcal{M}_{P_s}^{\text{Cor}}(2)}{\mathcal{M}_{P_s}^{\text{Cor}}(1)}$ and $\frac{\mathcal{M}_{P_s}^{\text{Ind}}(2)}{\mathcal{M}_{P_s}^{\text{Ind}}(1)}$, as shown in Figure 25. We can observe that correlated shadowing results in more temporal dependence between two successful transmission events. The gap of CSPs between correlated and independent shadowing decreases as the cell size L^2 becomes larger. Moreover, the dependence between two successful temporal transmission events increases as the cell size becomes smaller, especially when the SIR threshold is small. The reasons for the above observations can be ascribed to the adopted shadowing model, which essentially leads to exponential path loss.

C. Summary and Discussion

In this section, we have established an analytical framework to model the interference distribution and success probabilities in networks with spatially-correlated shadowing and spatially-independent shadowing. In particular, the analytical framework characterizes deterministic shadowing cells where the transmitters inside are associated with the same shadowing effects. The key lessons learned are as follows.

- Spatially-correlated shadowing generates the same interference as but higher variance than spatially-independent shadowing.
- The moments of the CSP_Φ under spatially-correlated shadowing is greater than that with spatially-independent shadowing. The performance increase reduces when the cell size shrinks.
- The successful transmission events are more temporally correlated with spatially-correlated shadowing than with spatially-independent shadowing.

Open Technical Issues: Most of the existing literature, e.g., [108], [110]–[113], only investigates network performance under the assumptions that the shadowing coefficients of the interferers in the same shadowing cells are either fully spatially correlated or independent. However, in practice, the interferers in the same shadowing cell may only experience partially correlated shadowing effect due to their location difference. More accurate shadowing models need to be developed by taking into account the properties (e.g., shape, density and mobility) of the obstacles that may exhibit location-dependency (e.g., urban and rural areas). Moreover, temporal shadowing variation due to mobility of obstacles (e.g., vehicles), access points (e.g., drone hot spots) or users is an important factor to be investigated in practical systems.

Different from the deterministic shadowing cells introduced in this section, reference [114] introduces a correlated shadow-

$$\mathcal{M}_{P_s}^{\text{Cor}}(b) = \prod_{k \in \mathbb{N}} \mathbb{E}_{T_k, T_{c(t)}} \left[\exp \left(-\lambda \int_{\mathbb{S}_k} \left(1 - \left(\frac{1}{1 + \theta r_t^\alpha \|x\|^{-\alpha} T_k / T_{c(t)}} \right)^b dx \right) \right), \quad (78)$$

$$\mathcal{M}_{P_s}^{\text{Ind}}(b) = \prod_{k \in \mathbb{N}} \mathbb{E}_{T_{c(t)}} \left[\exp \left(-\lambda \int_{\mathbb{S}_k} \left(1 - \mathbb{E}_{T_k} \left[\left(\frac{1}{1 + \theta r_t^\alpha \|x\|^{-\alpha} T_k / T_{c(t)}} \right)^b \right] dx \right) \right). \quad (79)$$

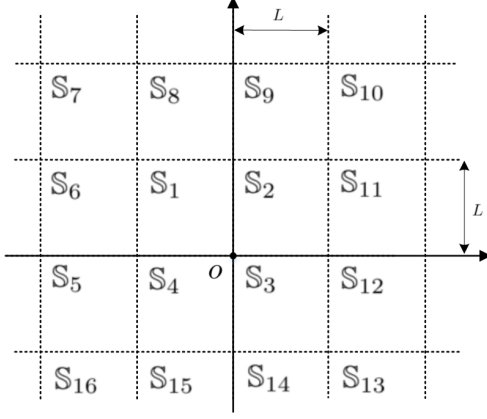


Fig. 23: Configuration of shadowing cells for simulations.

ing model dependent on the Poisson Voronoi cell. This model is suitable for studying coverage-oriented cellular networks where BSs deployment is configured to guarantee the cell-boundary users to acquire sufficient signal strength. An intriguing future direction is to extend the study of cell-dependent correlated shadowing to capacity-oriented cellular networks.

VI. SPATIAL-TEMPORAL INTERACTIONS BETWEEN QUEUES

The majority of the existing literature heavily relies on the assumption that each transmitter always has packets in the buffer to send out, which does not characterize random traffic flows. While the temporal randomness of the traffic flows complicates the analysis, it is nevertheless essential to understanding system-level performance. This is due to the fact that the traffic patterns in the evolving wireless networks are getting increasingly more dynamic and heterogeneous. The main difficulty of random traffic characterization originates from the correlation among the buffer statuses of different transmitters, often referred to as *interacting queues* [115], [116]. Since the queues interact spatially and temporally, an exact analysis of the mutual interference is quite challenging.

A. System Model

We consider both Poisson downlink networks (as introduced in Section III-A) and *Poisson bipolar networks* [3, Def. 5.8]. In a Poisson downlink network, the transmitters (i.e., BSs) and receivers (i.e., users) are distributed following independent homogeneous PPPs, denoted as $\Phi_B = \{x_j\}_{j \in \mathbb{N}}$ and Φ_u with intensity λ_B and λ_u , respectively. The points in Φ_B are assumed

to be ordered from nearest to farthest to the origin, i.e., $\|x_j\| < \|x_{j+1}\|$. It is assumed that each user is associated with its nearest BS for downlink transmission. In a Poisson bipolar network, the transmitters are distributed as a homogeneous PPP Φ with intensity λ . Each transmitter is paired with one receiver in a uniformly random direction with a link distance r_t . Without loss of generality, we study the performance of the typical receiver, conditioned to be at the origin, in both models.

We consider a discrete-time transmission and queueing model. Specifically, the data transmissions are divided into equal-duration time slots. We consider fixed-length data packets and assume it takes exactly one time slot to send out one packet. If a transmitter is scheduled for transmission, it can only send out the accumulated packet(s) that arrived prior to the transmission. In light of queueing, we assume that the incoming packets at each queue are stored in a buffer with infinite size and sent out on a first-in-first-out basis. A transmitted packet is removed from the head of the queue only if it is successfully decoded at the target receiver.

For Poisson downlink networks, each BS maintains an individual queue for the arrived packets of each associated user [30]. Hence, each BS has a number of queues equal to the number of users in its Voronoi cell. The temporal arrival of traffic at each queue follows an i.i.d. Bernoulli process with arrival rate ξ_u representing the probability of a new arrival per time slot. The users associated with the same BS are served based on *random scheduling* [30], i.e., each BS randomly selects one user within its Voronoi cell with equal probability to serve in each time slot. If the selected user has a non-empty queue at the BS, the BS is scheduled for transmission. Otherwise, the BS is muted.

For Poisson bipolar networks, the packet arrival at each transmitter $x_j \in \Phi$ follows an i.i.d. Bernoulli process with arrival rate ξ . At each time slot, the transmitters with non-empty buffer are all scheduled for transmission [117].

Given that the typical receiver is receiving data, its SIR in Poisson downlink and Poisson bipolar networks are given, respectively, by

$$\eta = \frac{h_1 \|x_1\|^{-\alpha}}{\sum_{j=2}^{\infty} \iota_j h_j \|x_j\|^{-\alpha}}, \quad x_j \in \Phi_B, \quad (80)$$

and

$$\eta = \frac{h_t \|x_t\|^{-\alpha}}{\sum_{j \in \mathbb{N}} \iota_j h_j \|x_j\|^{-\alpha}}, \quad x_j \in \Phi, \quad (81)$$

where ι_j denote the state indicator of the transmitter located at j which equals 1 and 0 when the transmitter is on and off, respectively.

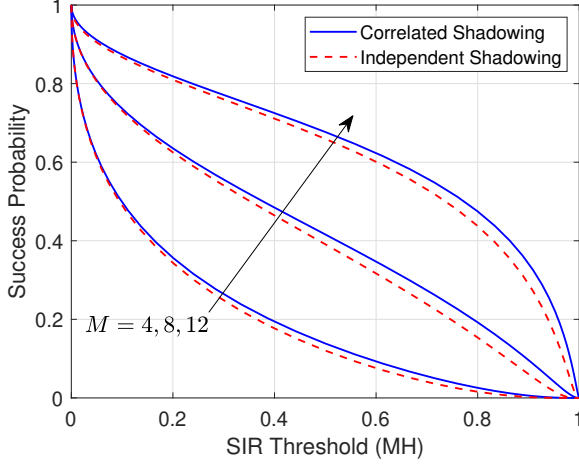


Fig. 24: Success probability versus SIR threshold under correlated and independent shadowing ($\lambda = 1$, $\kappa = 0.5$, $\alpha = 4$, $r_t = 1$).

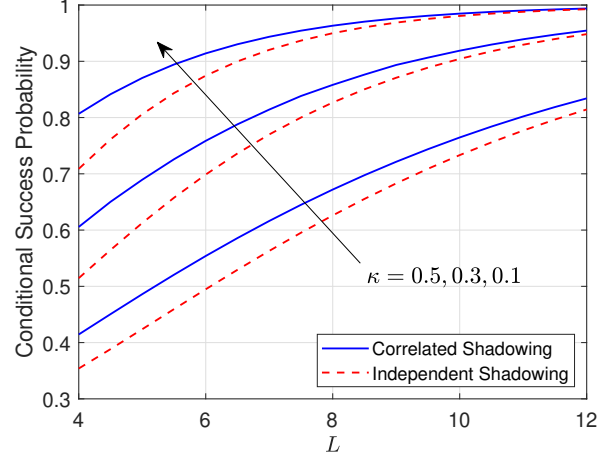


Fig. 25: CSP versus L under correlated and independent shadowing ($\lambda = 1$, $\alpha = 4$, $r_t = 1$, $\theta = 1$).

B. Performance Analysis

In large random networks, characterizing the exact queue interaction among the transmitters is quite challenging. Fortunately, in a large-scale network, the correlation among the interacting queues tends to be “weak” and “global” [118]. Therefore, the impact of interacting queues tends to be negligible. In the following, we show how to approximate the success probability in Poisson downlink networks with “interacting queues” by exploiting the mean-field property (e.g., as in [117], [118]).

- We first compute the success probability of the typical user’s cell, i.e., the cell containing the origin, based on the assumption that each BS at $x_j \in \Phi \setminus \{x_1\}$ is active independently with probability $p_A = \mathbb{E}[\iota_j]$.
- We then derive the active probability of the serving BS of the typical user as a function of packet arrival rate and success probability (i.e., service rate) based on queueing theory. By inserting the success probability obtained in the previous step, we can establish a fixed-point equation of p_A .
- We finally obtain the success probability by plugging in p_A obtained by solving the fixed-point equation.

Based on the above methodology, for a Poisson downlink network, we have the following result.

Theorem 7. With infinite buffer size and random scheduling, the success probability of a Poisson downlink network under Rayleigh fading can be approximated by P_s , obtained by solving the fixed-point equation

$$P_s = \left(1 + \frac{\xi_u (2F_1(1, -\delta; 1 - \delta; -\theta) - 1)}{\sum_{n=1}^{\infty} p_{N_u}(n) P_s / n} \right)^{-1}, \quad (82)$$

where

$$p_{N_u}(n) = \frac{\nu^\nu \Gamma(n + \nu) (\lambda_u / \lambda_B)^n}{n! \Gamma(\nu) (\lambda_u / \lambda_B + \nu)^{n+\nu}}, \quad (83)$$

with $\delta = 2/\alpha$ and $\nu = 3.5$.

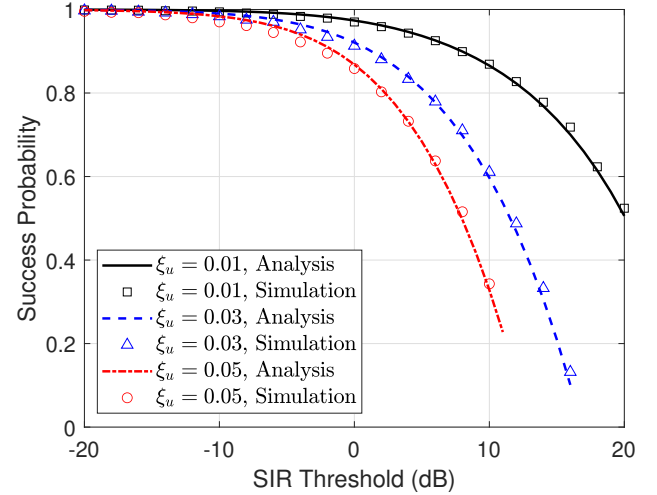


Fig. 26: Success probability versus SIR threshold in Poisson downlink networks ($\alpha = 4$, $\lambda_u / \lambda_B = 5$).

Proof. See Appendix E. □

Fig. 26 shows the success probability under different packet arrival rates ξ_u in a Poisson downlink network. We observe that ignoring the temporal and spatial correlation among queue iterations closely approximates the success probability achieved in the presence of the interactions. This is due to the “mean field” effect in a large-scale network. We also note that the traffic arrival rate ξ plays a pivotal part in the success probability. For example, to achieve a target success probability of 80%, the disparity of the supported SIR thresholds with $\xi_u = 0.01$ and $\xi_u = 0.05$ can be over 10 dB. A larger ξ_u decreases the success probability due to the increased density of interferers.

Next, we discuss the success probability in Poisson bipolar networks. By following the same methodology as used for

Poisson downlink networks, we obtain the result presented in the following theorem.

Theorem 8. If all links in a Poisson bipolar network experience Rayleigh fading and have infinite buffer size, the success probability of a typical link is given by (84), where $\delta = \frac{2}{\alpha}$.

Proof. See Appendix F. \square

Next, we explore the asymptotics of the success probability by utilizing the expansion of the Lambert- \mathcal{W} function as $\mathcal{W}(z) \sim z$ when $z \rightarrow 0$, which follows from $xe^x \sim x$ when $x \rightarrow 0$.

Fig. 27 depicts the success probability versus the SIR threshold with different settings of the packet arrival rate. The analytical results closely match the simulation results, which validates the effectiveness of the adopted approximation. Additionally, the success probability with $\xi = 1.0$ overlaps that with $\xi = 0.85$ when the SIR threshold is large (e.g., when $\theta > 20$ dB). The reason is that in both cases the service rate of each queue is below the packet arrival rate, and thus the buffer is always non-empty. As a result, all the transmitters remain active for transmission, which renders the same success probability under different packet arrival rates. This observation also shows that the success probability is lower-bounded by the case with a full load.

Fig. 28 further illustrates the success probability in a Poisson bipolar network under different densities. It can be observed that the approximation tends to lose its accuracy with the increase of the network density. The reason is that, in a Poisson bipolar network, the interferers can be arbitrarily close to the target receiver, and thus the queues of the serving transmitters and the interferers are strongly coupled. The spatial-temporal correlation of the buffer status cannot be ignored in this regime.

C. Summary and Discussion

This section has developed models to analyze wireless systems with unsaturated buffers. Specifically, by integrating results from queueing theory, we have presented the derivations of success probabilities for Poisson bipolar and Poisson downlink systems given the packet arrival rate. The key lessons learned are as follows.

- The spatial correlation among the queue statuses of different transmitters and the temporal correlation among the queue status of the same transmitter can be ignored when different queues are weakly coupled, e.g., in Poisson downlink networks, and cannot be ignored when the queues are strongly coupled, e.g., in Poisson bipolar networks.
- The temporal correlation among the successful transmission events decreases with increasing packet arrival rate.

Open Technical Issues: In Section VI, we have shown that in large Poisson downlink networks, the interaction among the queues becomes weak, and the impact of the temporal and spatial correlation tends to be negligible. However, this effect only appears due to the cellular infrastructure, where

the interfering BSs are further away from the typical user than the serving BS. In infrastructure-less networks that do not impose any restrictions on the interferers' locations, the interaction between the queues tends to be strong, thus, their spatial-temporal correlation cannot be ignored. Thus, further research efforts are needed to characterize the dynamics of strongly coupled queues in large-scale networks.

VII. SPATIALLY-CORRELATED INTERFERENCE AND RELAYING

Multihop relaying is an effective technique to extend the communication range with limited transmit power at each hop, enhancing the reliability and throughput of point-to-point communication [119]. In multihop relaying, the transmission performance at different hops can be impacted by some common interferers. Therefore, the spatial characteristics of the interferers can have a substantial impact on the end-to-end transmission performance [78]. The goal of this section is to quantify the spatially and temporally-correlated interference and demonstrate its effect on the performance of multihop relaying.

A. System Model

We consider a multihop relay network consisting of a source node and a receiver node, which is M hops away from the transmitter node, in a random field of interferers. The locations of the source node S and the m -th hop receiver (i.e., $(m+1)$ -th hop transmitter) are deterministic and denoted by x_S and z_m , respectively. The transceiver node in each hop works in a half-duplex fashion. Let d_m represent the Euclidean distance of the m -th hop link, i.e., $d_1 = \|z_1 - x_S\|$ and $d_m = \|z_m - z_{m-1}\|$, for $2 \leq m \leq M$. The route from the source node to the M -th hop receiver is fixed. The decode-and-forward (DF) relaying protocol is adopted such that each hop first decodes the received signal and forwards the re-encoded version to the next hop. We consider a Poisson field of interferers where the locations of the interferers $\Phi \subset \mathbb{R}^2$ are a PPP with intensity λ , as illustrated in Fig. 29(a). Note that the relaying system of interest is considered to be independent of the interferers.

We assume that each transmitter in the considered system uses unit transmit power. Let $\Phi_m = \{x_{j,m}\}_{j \in \mathbb{N}}$ denote the point process of interferers during the m -th hop transmission. The SIR at the m -th hop can be expressed as

$$\eta_m = \frac{h_m d_m^{-\alpha}}{\sum_{j \in \mathbb{N}} h_{j,m} \|x_{j,m} - z_m\|^{-\alpha}}, \quad (85)$$

where h_m represents the power gain of small-scale fading for the m -th transmission hop and $h_{j,m}$ represents the small-scale fading gain between the interferer j and the m -th hop receiver, which are both i.i.d. exponential random variables with unit mean.

For the analysis of the success probability of multihop relaying, we consider both QSI and FVI with which the transmission of different hops are subject to the interference from the same point process, i.e., $\Phi_1 = \Phi_2 = \dots = \Phi_M = \Phi$,

$$P_s \approx \max \left\{ \exp \left(\mathcal{W} \left(-\xi \lambda \pi r_t^2 \theta^\delta \Gamma(1+\delta) \Gamma(1-\delta) \right) \right), \exp \left(-\lambda \pi r_t^2 \theta^\delta \Gamma(1+\delta) \Gamma(1-\delta) \right) \right\}, \quad (84)$$

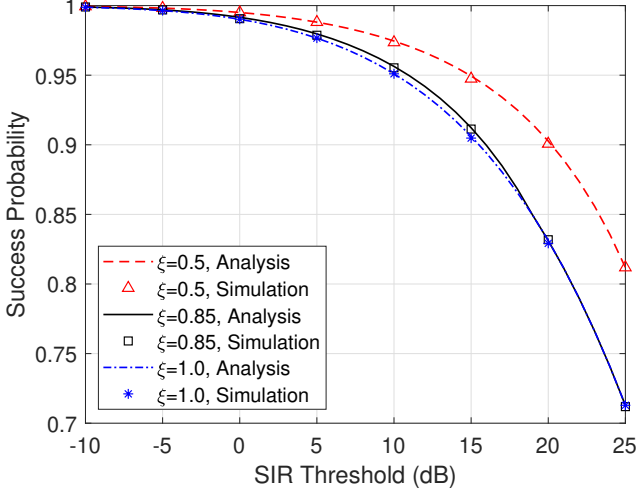


Fig. 27: Success probability as a function of the SIR threshold in Poisson bipolar networks ($\alpha = 4$, $r_t = 2$, $\lambda = 0.001$).

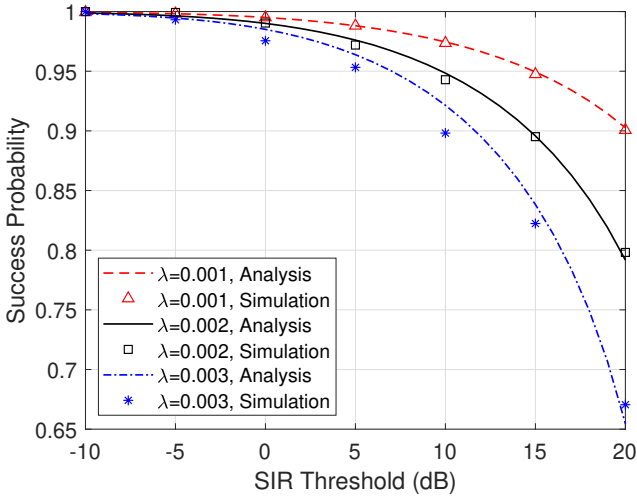


Fig. 28: Success probability in Poisson bipolar networks under different intensities.

and independent point processes, respectively. With DF relaying protocol [121], the end-to-end success probability of an M -hop relaying system with QSI and FVI are given, respectively, by

$$\begin{aligned} \mathcal{P}_M^{\text{QSI}} &= \mathbb{E} \left[\mathbb{P} \left[\bigcap_{m=1}^M \{ \eta_m > \theta \} \mid \Phi \right] \right] \\ &= \mathbb{E} \left[\prod_{m=1}^M \mathbb{P} [\eta_m > \theta \mid \Phi] \right] \end{aligned} \quad (86)$$

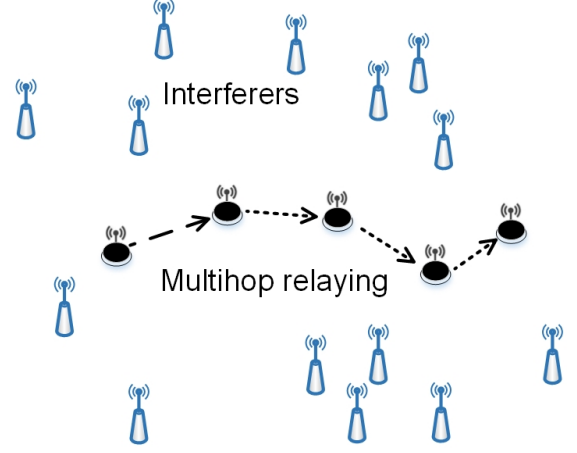


Fig. 29: Multihop relaying in the presence of random interferers [120].

and

$$\begin{aligned} \mathcal{P}_M^{\text{FVI}} &= \mathbb{P} \left[\bigcap_{m=1}^M \{ \eta_m > \theta \} \right] \\ &= \prod_{m=1}^M \mathbb{P} [\eta_m > \theta]. \end{aligned} \quad (87)$$

B. Moments of the End-to-End CSP_Φ

This subsection characterizes the moments of the end-to-end CSP_Φ of the multihop relaying.

- We compute the end-to-end JSP that the transmissions of the M hops all succeed given the point process, i.e., $\mathbb{P} \left[\bigcap_{m=1}^M \{ \eta_m > \theta \} \mid \Phi_m \right]$ (e.g., as in [122]).
- We derive the moments of the CSP_Φ based on the PGFL of the PPP.

Following the above methodology, we obtain the moments of the end-to-end CSP_Φ under both QSI and FVI in the following theorem.

Theorem 9. The moments of the end-to-end CSP for an M -hop relaying system in a Poisson field of interferers are given by (88).

Proof. See **Appendix G**. \square

Fig. 30 illustrates the success probability for an M -hop linear-route multihop relaying system where all relays are placed on the source-destination line and all M links have distance l . With both fields of interferers, we can observe that correlated QSI provides a higher success probability than independent FVI. This can be understood from the perspective of the CSP of m -th hop given that the transmissions of the

$$\mathcal{M}_{P_s}^{\text{Poi}}(b, M) = \begin{cases} \exp\left(-\lambda \int_{\mathbb{R}^2} \left(1 - \prod_{m=1}^M \left(\frac{1}{1 + \theta d_m^\alpha \|x - z_m\|^{-\alpha}}\right)^b\right) dx\right), & \text{QSI} \\ \exp\left(-\lambda \prod_{m=1}^M \int_{\mathbb{R}^2} \left(1 - \left(\frac{1}{1 + \theta d_m^\alpha \|x - z_m\|^{-\alpha}}\right)^b\right) dx\right), & \text{FVI} \end{cases} \quad (88)$$

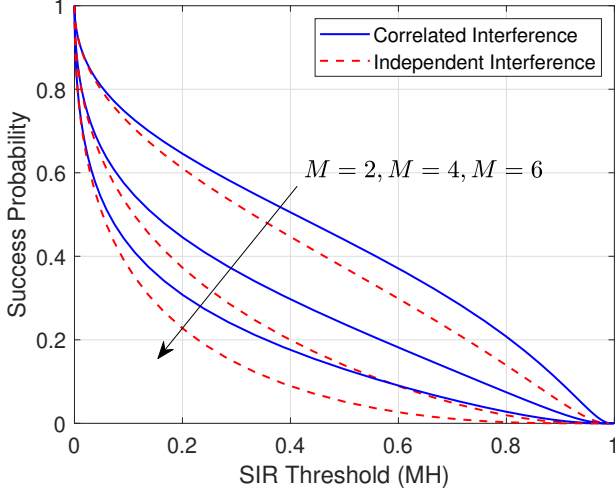


Fig. 30: Success probability of a multihop relaying system in a Poisson field of interferers ($\alpha = 4$, $\lambda = 0.1$, $d_1 = d_2 = \dots = d_M = 1$).

previous hops all succeed. With FVI, we can see from (88) that $\mathcal{M}_{P_s}^{\text{Poi}}(b, M)/\mathcal{M}_{P_s}^{\text{Poi}}(b, M-1) = \mathcal{M}_{P_s}^{\text{Poi}}(b, 1)$. This is consistent with the fact that with FVI, the successful transmission event of the m -th hop is independent of those of the previous hops.

Fig. 31 demonstrates the spatial CSP of the m -th hop given that the transmissions in the previous $m-1$ hops all succeed in the scenario with QSI, i.e., $(\mathcal{M}_{P_s}^{\text{Poi}}(1, M)/\mathcal{M}_{P_s}^{\text{Poi}}(1, M-1))$ for $M \geq 2$. It can be seen that, with QSI, the CSP increases considerably given the successful transmission of the first hop, especially when the SIR threshold θ is high. Thus, the end-to-end success probability with QSI exceeds that with FVI. It is worth noting that with temporally correlated QSI, the conditional outage probability, given the outage events of the previous hops, also increases. However, this does not worsen the end-to-end success probability, since the outage event of any single hop leads to an end-to-end outage.

Moreover, Fig. 32 depicts the temporal CSP of an M -hop relaying system given a previous end-to-end successful transmission, i.e., $\mathcal{M}_{P_s}^{\text{Poi}}(2, M)/\mathcal{M}_{P_s}^{\text{Poi}}(1, M)$. We can observe that the CSP increases with the number of hops. The reason is that a successful transmission with a larger number of hops indicates a good channel condition, and thus the next end-to-end transmission is more likely to succeed.

C. Summary and Discussion

In this section, we have discussed the impact of spatial-temporal interference on the performance of multihop relaying.

In particular, we have presented the derivations of end-to-end success probability of multihop relaying in Poisson fields of interferers under both FVI and QSI. The main lessons learned are as follows.

- Correlated spatial-temporal interference (i.e., QSI) imposed by the same interferers at different hops results in higher success probability than independent spatial-temporal interference (i.e., FVI).
- With correlated spatial-temporal interference, as m increases, the successful transmission of the m -th hop is more dependent on the successful transmissions of the previous hops.
- With independent spatial-temporal interference, the CSP of the m -th hop given the successful transmission of the previous hops is the same as the success probability of the m -th hop without the condition.
- With correlated spatial-temporal interference, the end-to-end successful transmission events are more correlated when the number of hops is larger.

Open Technical Issues: This section has presented the characterization of end-to-end relaying performance under QSI and FVI, which considers a purely static network and an independent network, respectively. The two types of interference represent two extreme cases with fully correlated and independent interferer locations. In general, the ambient interferers (e.g., mobile users) may have a certain degree of movement during different transmission attempts and result in neither a static nor independent network environment. In such an environment, the resulting temporal interference is only partially correlated. An accurate characterization of the spatial-temporal correlation at different locations would be required for performance evaluation of multihop relaying systems.

VIII. TEMPORALLY-CORRELATED INTERFERENCE AND RETRANSMISSION

The goal of this section is to derive the success probabilities of a target transmission link with retransmissions under temporally correlated and independent interference.

A. System Model

This section considers the Poisson ad hoc network model (as introduced in Section III-A). Let $\Phi^{(k)} = \{x_j^{(k)}\}_{j \in \mathbb{N}}$ denote the node locations in time slot k . The aggregated interference and

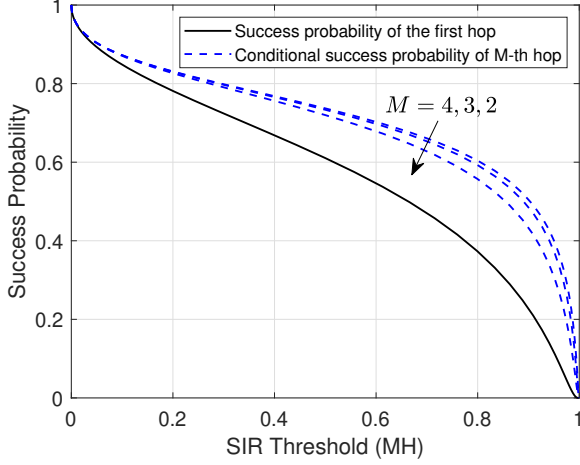


Fig. 31: The spatial CSP of the M -th hop relaying in a Poisson field of interferers with QSI.

receive SIR at the target receiver located at o in time slot k can be expressed, respectively, as

$$I_o^{(k)} = \sum_{j \in \mathbb{N}} h_j^{(k)} \|x_j^{(k)}\|^{-\alpha}$$

$$\eta^{(k)} = \frac{h_t^{(k)} \|x_t^{(k)}\|^{-\alpha}}{I_o^{(k)}}, \quad (89)$$

where $h_t^{(k)}$ and $h_j^{(k)}$ denotes the small-scale fading gains between the transmitters at $x_t^{(k)}$, $x_j^{(k)}$ and the target receiver in time slot k which are i.i.d. exponential random variables with unit mean.

We consider both QSI and FVI with which different transmission attempts are influenced by the same set and different sets of interferers, respectively. As a result, the interferences, and thus the successful transmission events, in the considered system are temporally correlated with and independent of QSI and FVI, respectively.

B. Performance Analysis

Let $A_k \triangleq \{\eta^{(k)} > \theta\}$ denote the successful transmission event in time slot k . The JSP of K transmissions in the cases with QSI and FVI are defined, respectively, as

$$\mathcal{J}_K^{\text{QSI}} \triangleq \mathbb{E} \left[\mathbb{P} \left[\bigcap_{k=1}^K A_k \mid \Phi^{(k)} \right] \right]$$

$$= \mathbb{E} \left[\prod_{k=1}^K \mathbb{P}[A_k \mid \Phi^{(k)}] \right] \quad (90)$$

and

$$\mathcal{J}_K^{\text{FVI}} \triangleq \mathbb{P} \left[\bigcap_{k=1}^K A_k \right] = \prod_{k=1}^K \mathbb{P}[A_k]. \quad (91)$$

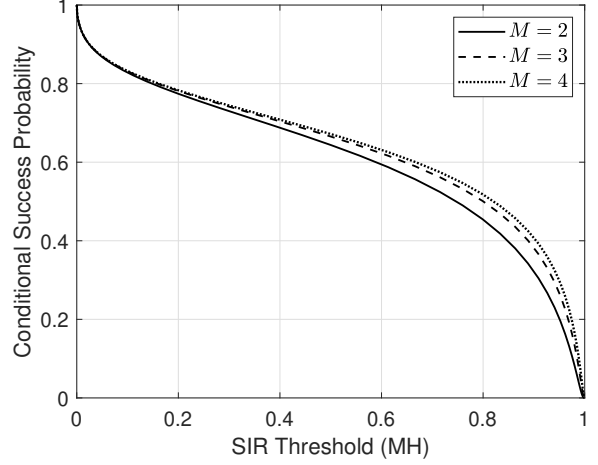


Fig. 32: The temporal CSP of an M -hop relaying system in a Poisson field of interferers with QSI.

1) *Joint Success Probability:* In this subsection, we show how to derive the JSP of K transmissions based on the PGFL of the PPP. The final results are presented in the following theorem [81].

Theorem 10. (Temporal JSP) With a Poisson field of interferers, the probability that a link over distance r_t has K successful transmissions in a row with QSI and FVI are given, respectively, as

$$\mathcal{J}_K^{\text{QSI}} = \exp(-c\lambda\theta^\delta r_t^2 D_K(\delta)), \quad (92)$$

$$\mathcal{J}_K^{\text{FVI}} = \exp(-c\lambda\theta^\delta r_t^2 K), \quad (93)$$

where $c = \pi\Gamma(1+\delta)\Gamma(1-\delta)$ and $D_K(\delta) = \frac{\Gamma(K+\delta)}{\Gamma(K)\Gamma(1+\delta)}$.

Proof. See **Appendix H**. \square

Remark 11: $\mathcal{J}_1^{\text{QSI}} = \mathcal{J}_1^{\text{FVI}}$ as $D_1(\delta) = 1$, which indicates that the success probability of any single time slot is not influenced by the type of interference experienced.

Remark 12: Since $\delta < 1$ (i.e., $\alpha > 2$), it is readily checked that $D_K(\delta) < K$ for $K \in \mathbb{N}$, and thus $\mathcal{J}_K^{\text{QSI}} > \mathcal{J}_K^{\text{FVI}}$. This reveals that temporally-correlated QSI results in higher JSP than the temporally-independent FVI. Fig. 33 illustrates the JSP for $K = 2, 3, 4$ with both QSI and FVI.

Remark 13: When $\delta \rightarrow 0$ (i.e., $\alpha \rightarrow \infty$), it can be found that $\mathcal{J}_1^{\text{QSI}} = \mathcal{J}_2^{\text{QSI}} = \dots = \mathcal{J}_K^{\text{QSI}} = e^{-\lambda\pi r_t^2}$, equivalently $\mathbb{P}[A_{k+1}|A_k] = 1$ for $k = 1, \dots, K$. This indicates that despite the effect of small-scale fading, the successful transmission events are fully correlated.

Remark 14: When $\delta \rightarrow 1$ (i.e., $\alpha \rightarrow 2$), it can be found that a) $\mathcal{J}_K^{\text{QSI}} \rightarrow e^{-c\lambda\theta^\delta r_t^2 K}$, equivalently $\mathbb{P}[A_{k+1}|A_k] = e^{-c\lambda\theta^\delta r_t^2}$ for $k = 1, \dots, K$. This means the successful transmission events are independent; b) $c \rightarrow \infty$, which indicates that $\mathcal{J}_K^{\text{QSI}} \rightarrow 0$.

2) *Conditional Success Probability:* As a consequence of Theorem 10, the CSPs of succeeding at $K+1$ -th transmission

given the previous K successful transmissions can be directly obtained following Bayes rule as

$$\begin{aligned} \mathcal{C}_{K+1,K}^{\text{QSI}} &= \mathbb{P}[A_{K+1} | A_1, \dots, A_K] \\ &= \frac{\mathcal{J}_{K+1}^{\text{QSI}}}{\mathcal{J}_K^{\text{QSI}}} \\ &= \exp\left(-c\lambda\theta^\delta r_t^2 (D_{K+1}(\delta) - D_K(\delta))\right), \end{aligned}$$

and

$$\mathcal{C}_{K+1,K}^{\text{FVI}} = \frac{\mathcal{J}_{K+1}^{\text{FVI}}}{\mathcal{J}_K^{\text{FVI}}} = \exp(-c\lambda\theta^\delta r_t^2).$$

Remark 15: It can be readily checked that $\partial \mathcal{C}_{K+1,K}^{\text{QSI}} / \partial K > 0$ which indicates that $\mathcal{C}_{K+1,K}^{\text{QSI}}$ monotonically increases with the number of transmission attempts K .

Fig. 34 shows the temporal CSPs with QSI when $K = 1, 2, 3, 4$. The numerical results illustrate the above-discussed properties. Given that the previous transmissions succeed, the CSP considerably increases, especially with a high SIR threshold θ .

3) *Correlation Coefficient:* Let $Q_k = \mathbb{1}_{A_k}$ be the indicator that event A_k happens. The correlation coefficient between Q_k and Q_j with $k \neq j$ is

$$\begin{aligned} \zeta_{Q_k, Q_j}^{\text{QSI}} &= \frac{\mathbb{P}[A_k \cap A_j] - \mathbb{P}^2[A_k]}{\mathbb{P}[A_k \cap A_k] - \mathbb{P}^2[A_k]} \\ &\stackrel{(a)}{=} \frac{\mathcal{J}_2^{\text{QSI}} - (\mathcal{J}_1^{\text{QSI}})^2}{\mathcal{J}_1^{\text{QSI}}(1 - \mathcal{J}_1^{\text{QSI}})} \\ &= \frac{\exp(-c\lambda\theta^\delta r_t^2(\delta + 1)) - \exp(-2c\lambda r_t^2\theta^\delta)}{\exp(-c\lambda r_t^2\theta^\delta)(1 - \exp(-c\lambda r_t^2\theta^\delta))} \\ &= \frac{\exp(c\lambda\theta^\delta r_t^2(1 - \delta)) - 1}{\exp(c\lambda\theta^\delta r_t^2) - 1}, \end{aligned} \quad (94)$$

where (a) holds as $\{Q_k\}_{k \in \{1, \dots, K\}}$ are identically distributed. Moreover, $\zeta_{Q_k, Q_j}^{\text{FVI}} = 0$ as $\mathcal{J}_2^{\text{FVI}} = (\mathcal{J}_1^{\text{FVI}})^2$. Since $\zeta_{Q_k, Q_j}^{\text{QSI}} > 0$ and $\zeta_{Q_k, Q_j}^{\text{FVI}} = 0$, it is evident that QSI and FVI result in temporally-correlated and independent successful transmission events, respectively.

Remark 16: When $\delta \rightarrow 0$ and $\delta \rightarrow 1$, $\zeta_{Q_k, Q_j}^{\text{QSI}} \rightarrow 1$ and $\zeta_{Q_k, Q_j}^{\text{QSI}} \rightarrow 0$, indicating that the Q_k and Q_j become fully correlated and fully uncorrelated, respectively.

Remark 17: For given δ and r , $\zeta_{Q_k, Q_j}^{\text{QSI}}$ is a decreasing function of λ and θ . This can be checked that since $0 < \delta < 1$ (i.e., $\alpha > 2$), the denominator scales up at a higher rate than the numerator of (94) with the increase of λ or θ . Fig. 35 shows the correlation coefficient of successful transmission events with QSI.

4) *Success Probability with Retransmissions:* In the case of transmission failure, multiple transmissions can be carried out to deliver a message. Let us consider a retransmission protocol where the receiver requests the associated transmitter to send the message again upon a transmission failure until reaching a maximum number of transmission attempts denoted as K . At the receiver side, the received signal at each time slot is decoded

independently. The success probability with retransmissions can be expressed as

$$\mathcal{P}_K \triangleq \mathbb{P}\left[\bigcup_{k=1}^K A_k\right]. \quad (95)$$

Note that from the above definition, we have $\mathcal{P}_1 = \mathcal{J}_1$, where \mathcal{J}_k , $k \in \mathbb{N}$, is defined in (7).

Let $\mathcal{A} \triangleq \{A_1, \dots, A_K\}$ denote the set of the successful transmission events and $\mathcal{P}(\mathcal{A})$ represent the power set of \mathcal{A} . By applying the inclusion-exclusion principle, we obtain the success probability with retransmissions as

$$\begin{aligned} \mathcal{P}_K &= \sum_{A \in \mathcal{P}(\mathcal{A})} (-1)^{|A|+1} \mathbb{P}[A] \\ &\stackrel{(a)}{=} \sum_{k=1}^K (-1)^{k+1} \binom{K}{k} \mathcal{J}_k, \end{aligned} \quad (96)$$

where (a) follows as $\{\eta_k\}_{k \in \{1, 2, \dots, K\}}$ are identically distributed, and \mathcal{J}_k has been obtained in Theorem 10.

Fig. 36 shows the success probability with retransmissions with both QSI and FVI. It can be seen that $\mathcal{P}_K^{\text{FVI}}$ exceeds $\mathcal{P}_K^{\text{QSI}}$. The reason is that the transmission failures are temporally correlated and thus are likely to occur in succession with QSI. By contrast, with FVI, transmissions have a better chance to succeed as previous transmission failure does not infer a lower success probability in the current time slot. This can be understood by checking the dependency of transmission failure. Let $\bar{S} \triangleq \{\eta^{(k)} < \theta\}$. By following the inclusion-exclusion principle, the joint outage probability of two transmission events is given by

$$\begin{aligned} \bar{\mathcal{J}}_2 &= \mathbb{P}[\bar{A}_1 \cap \bar{A}_2] \\ &= 1 - \mathbb{P}[A_1 \cup A_2] \\ &= 1 - (2\mathcal{J}_1 - \mathcal{J}_2) \\ &= 1 - 2\exp(-c\lambda\theta^\delta r_t^2) + \exp(-c\lambda\theta^\delta r_t^2(1 + \delta)). \end{aligned}$$

Subsequently,

$$\frac{\bar{\mathcal{J}}_2^{\text{QSI}}}{(\bar{\mathcal{J}}_1^{\text{QSI}})^2} = 1 + \frac{\exp(c\lambda\theta^\delta r_t^2(1 - \delta)) - 1}{(\exp(c\lambda\theta^\delta r_t^2) - 1)^2} > 1,$$

and $\frac{\bar{\mathcal{J}}_2^{\text{FVI}}}{(\bar{\mathcal{J}}_1^{\text{FVI}})^2} = 1$.

Moreover, it is evident that the conditional outage probability increases given the occurrence of previous transmission failure by checking

$$\begin{aligned} \mathbb{P}[\bar{A}_1 | \bar{A}_2] &= \frac{\mathbb{P}[\bar{A}_1 \cap \bar{A}_2]}{\mathbb{P}[\bar{A}_1]} \\ &= 1 - \exp(-c\lambda\theta^\delta r_t^2) \frac{1 - \exp(-\delta c\lambda\theta^\delta r_t^2)}{1 - \exp(-c\lambda\theta^\delta r_t^2)} \\ &\geq 1 - \exp(-c\lambda\theta^\delta r_t^2) = \mathbb{P}[\bar{A}_1]. \end{aligned}$$

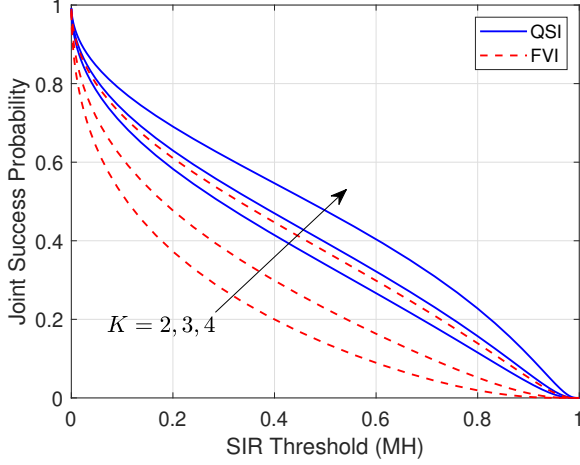


Fig. 33: JSP with a different number of transmission attempts in Poisson ad hoc networks ($\alpha = 4$, $\lambda = 0.1$).

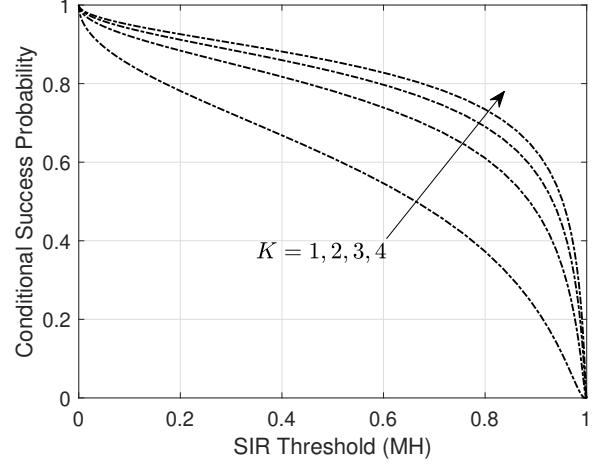


Fig. 34: CSP with a different number of transmission attempts in Poisson ad hoc networks ($\alpha = 4$, $\lambda = 0.1$).

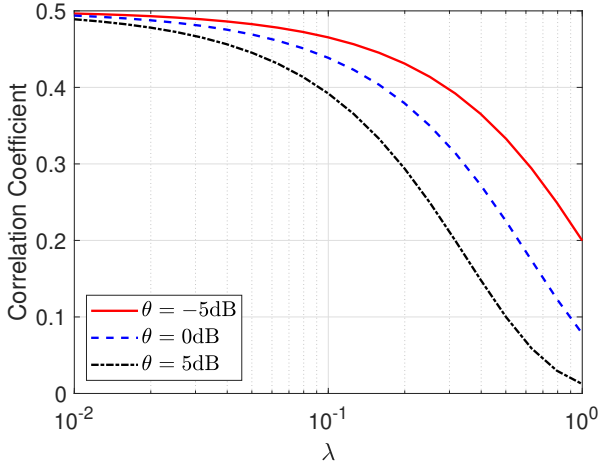


Fig. 35: Correlation coefficient in Poisson ad hoc networks ($\alpha = 4$).

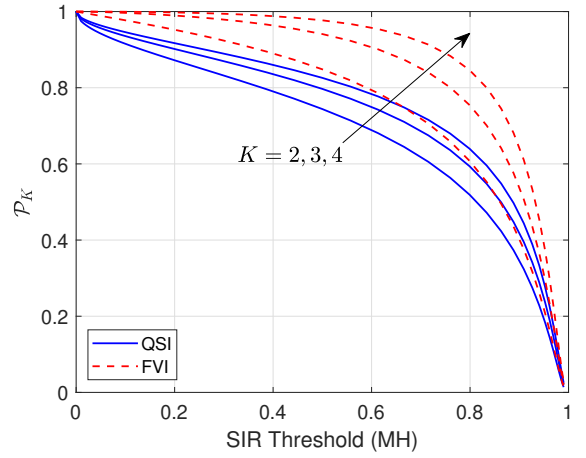


Fig. 36: Success probability with retransmission in Poisson ad hoc networks ($\alpha = 4$, $\lambda = 0.1$).

C. HARQ Retransmission Schemes

For transmission and decoding, similar to [85], we consider two categories of HARQ, i.e., Type-I HARQ and Type-II HARQ schemes.

- If the SIR at a target receiver for the initial transmission does not exceed the threshold θ , a one-time retransmission request is sent to the serving transmitter. After receiving the retransmitted signals, the target receiver with Type-I HARQ abandons the received signal from the initial transmission and decodes only from the received signal from the retransmission. Given a maximum number of transmissions K and a SIR threshold θ , the success probability is given by

$$\mathcal{P}_I = \mathbb{P} \left[\bigcup_{k=1}^K \left\{ \eta^{(k)} > \theta \right\} \right], \quad (97)$$

where $\eta^{(k)}$ is the SIR of k -th transmission of the same content with Type-I HARQ retransmission scheme.

- With Type-II HARQ with chase combining (CC) codes and maximal ratio combining (MRC) of signals from both the initial transmission and retransmission, the success probability is given by

$$\mathcal{P}_{II} = \mathbb{P} \left[\bigcup_{k=1}^K \left\{ \Upsilon_k > \theta \right\} \right], \quad (98)$$

where $\Upsilon_k = \sum_{i=1}^k \eta^{(i)}$ is the effective SIR after the k -th transmission of the same content with the Type-II HARQ retransmission scheme.

In what follows, the success probabilities of Type-I HARQ and Type-II HARQ-CC schemes are derived for the case with $K = 2$, i.e., each packet is retransmitted once if the initial transmission is not successful. The cases with $K \geq 3$ can be obtained by following the same methodology straightforwardly.

1) *Success Probability of Type-I HARQ*: When $K = 2$, the success probability with Type-I HARQ given in (97) can be expressed as [123]

$$\begin{aligned} \mathcal{P}_I &= \mathbb{P}[\eta^{(1)} > \theta] + \mathbb{P}[\eta^{(1)} > \theta, \eta^{(2)} \leq \theta] \\ &\stackrel{(a)}{=} \sum_{k=1}^2 \mathbb{P}[\eta^{(k)} > \theta] - \mathbb{P}[\eta^{(1)} > \theta, \eta^{(2)} > \theta] \\ &\stackrel{(b)}{=} \begin{cases} \sum_{k=1}^2 \mathbb{P}[\eta^{(k)} > \theta] - \mathbb{P}[\eta^{(1)} > \theta, \eta^{(2)} > \theta \mid \Phi], & \text{QSI} \\ \sum_{k=1}^2 \mathbb{P}[\eta^{(k)} > \theta] - \prod_{k=1}^2 \mathbb{P}[\eta^{(k)} > \theta], & \text{FVI} \end{cases} \end{aligned} \quad (99)$$

where (a) applies the inclusion-exclusion principle and (b) follows as the point processes across different time slots are the same and i.i.d. for QSI and FVI, respectively. Subsequently, based on the definition of $\mathcal{J}_K^{\text{QSI}}$ and $\mathcal{J}_K^{\text{FVI}}$ in (90) and (91), respectively, we can obtain the following corollary.

Corollary 3. Given that each unsuccessfully transmitted packet is retransmitted once, the success probability under Type-I HARQ retransmission scheme with QSI and FVI are given, respectively, as

$$\mathcal{P}_I^{\text{QSI}} = 2\mathcal{J}_1^{\text{QSI}} - \mathcal{J}_2^{\text{QSI}}, \quad (100)$$

$$\mathcal{P}_I^{\text{FVI}} = 2\mathcal{J}_1^{\text{FVI}} - \left(\mathcal{J}_1^{\text{FVI}}\right)^2, \quad (101)$$

where $\mathcal{J}_K^{\text{QSI}}$ and $\mathcal{J}_K^{\text{FVI}}$ for $K \in \mathbb{N}$ are given in (92) and (93), respectively.

2) *Success Probability of Type-II HARQ-CC*: Next, we discuss how to obtain the success probability of Type-II HARQ-CC. When $K = 2$, the success probability with Type-II HARQ given in (97) can be expressed as $\mathcal{P}_{II} = \mathbb{P}[\eta^{(1)} > \theta] + \mathbb{P}[\eta^{(1)} + \eta^{(2)} > \theta, \eta^{(1)} \leq \theta]$ where the first term on the right-hand side of the equality has been obtained in Section VIII.C.1. To obtain the second term, we need to compute the distribution of the SIR of any single time slot. In particular, we first calculate the CDF of the SIR and then obtain the corresponding PDF by taking the derivative of the CDF. With the PDF of the SIR, we then derive the joint probability that events $\eta^{(1)} + \eta^{(2)} > \theta$ and $\eta^{(1)} < \theta$ both occur. The final results are presented in the following corollary.

Corollary 4. Given that each unsuccessfully transmitted packet is retransmitted once, the success probability under Type-II HARQ-CC retransmission scheme with QSI and FVI are given, respectively, as

$$\begin{aligned} \mathcal{P}_{II}^{\text{QSI}} &= \exp(-c\lambda\theta^\delta r_t^2) + 2\pi\lambda \int_0^\theta \int_0^\infty \frac{r_t^\alpha r^{-\alpha} J(r, u)}{(1 + ur_t^\alpha r^{-\alpha})^2} r dr \\ &\quad \times \exp\left(-2\pi\lambda \int_0^\infty \left(1 - \frac{J(r, u)}{1 + ur_t^\alpha r^{-\alpha}}\right) r dr\right) du, \end{aligned} \quad (102)$$

and

$$\mathcal{P}_{II}^{\text{FVI}} = \exp(-c\lambda\theta^\delta r_t^2) + 2\pi\lambda \int_0^\theta \int_0^\infty \frac{r_t^\alpha r^{-\alpha}}{(1 + ur_t^\alpha r^{-\alpha})^2} r dr$$

$$\begin{aligned} &\times \exp\left(-2\pi\lambda \int_0^\infty \left(1 - \frac{1}{1 + ur_t^\alpha r^{-\alpha}}\right) r dr\right) \\ &\times \exp\left(-2\pi\lambda \int_0^\infty \left(1 - J(r, u)\right) r dr\right) du, \end{aligned} \quad (103)$$

where $c = \pi\Gamma(1 + \delta)\Gamma(1 - \delta)$ and $J(r, u) = \frac{1}{1 + (\theta - u)r_t^\alpha r^{-\alpha}}$.

Proof. See Appendix I. \square

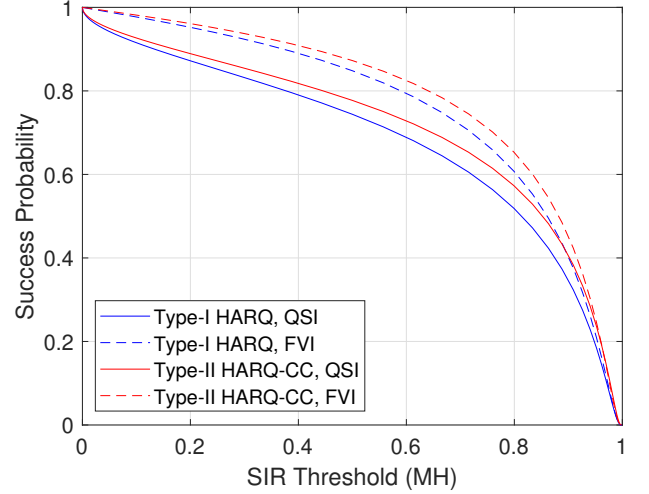


Fig. 37: Success probabilities of HARQ retransmission schemes in Poisson ad hoc networks ($K = 2$).

Fig. 37 depicts the success probabilities of Type-I HARQ and Type-II HARQ-CC when $K = 2$. As shown, Type-II HARQ-CC provides a higher success probability at all SIR thresholds. Nevertheless, the success probabilities of the two schemes are comparable when the SIR threshold is small (e.g., $\theta < 0.05$ MH). This indicates that, in the high-coverage regime, the SIR gain due to MRC has an imperceptible effect on the success probability. The reason is that low SIR thresholds render a high success probability of the first transmission, which makes the impact of retransmission negligible.

D. Summary and Discussion

This section has discussed the influence of temporal interference correlation on the retransmission performance of a transmission link in a Poisson field of interferers. In particular, we have derived the JSP of multiple transmissions, the correlation coefficient for two transmissions, and the success probability for a given number of transmission attempts. Furthermore, based on the above results, we have shown how to analyze the success probabilities under Type-I HARQ and Type-II HARQ retransmission schemes. The lessons learned are as follows.

- Temporally-correlated interference results in a higher JSP than temporally-independent interference. The performance gap between the two cases increases with the number of transmission attempts.
- The CSP of K -th transmission given that the previous $K - 1$ transmissions all succeed increases with K and

- Temporally-correlated interference results in higher success probability with retransmission than temporally-independent interference. The gap between the success probabilities with retransmission under the two types of interference increases with the number of transmission attempts.
- The performance gain of Type-II HARQ over Type-I HARQ is larger with temporally-correlated interference than with temporally-independent interference.

- The performance gain of Type-II HARQ over Type-I HARQ is larger with temporally-correlated interference than with temporally-independent interference.

Open Technical Issues: This section has presented models to characterize the impact of temporal interference correlation in a retransmission-based (or multi-packet transmission) system. We have considered both Type-I HARQ and Type-II HARQ systems and evaluated both CSP and JSP. From a practical point of view, the impact of “interacting queues” may need to be considered in a retransmission scenario. Also, the performance of HARQ systems in the presence of interference correlation in different wireless systems (e.g., non-orthogonal multiple access (NOMA) systems and MIMO systems) will be worth investigating.

In some emerging scenarios, such as device-to-device communication [124] and mobile social networks [125], an information source may rely on mobile users to spread messages to multiple destinations. To model the process of information spreading in such systems, a possible direction is to incorporate user mobility models into the stochastic geometry analysis.

IX. SPATIALLY AND TEMPORALLY CORRELATED INTERFERENCE IN MOBILE SYSTEMS

This section investigates the effect of mobility on the spatial-temporal SIR correlation. We analyze the JSP in mobile networks with spatially and temporally correlated interference.

A. Introduction

We consider two mobile network scenarios. In the first scenario, similar to [77], we consider a downlink communication in a Poisson cellular network, and provide the steps to calculate the spatial-temporal JSP of a mobile user at two different time instants. In the second scenario, similar to [24], we consider a Poisson bipolar model where the desired transmitter and receiver are static while other transmitters in the network move according to a random mobility model. For this scenario, we also explain the steps to calculate the JSP at the static receiver [126].

Let us denote the success event indicator at time t_1 at location u_1 by $A_1 = \mathbb{1}_{\{\text{SIR}_1 > \theta\}}$ and at time t_2 at location u_2 by $A_2 = \mathbb{1}_{\{\text{SIR}_2 > \theta\}}$, where SIR_1 is the SIR at the receiver at location u_1 at time t_1 , and SIR_2 is the SIR at the receiver at location u_2 at time t_2 ; θ is the target SIR. In the following, we present the steps to calculate $\mathbb{E}[A_1 A_2]$ (i.e., the spatial-temporal JSP) for the two mentioned network scenarios. Note that when the network is stationary over time, $\mathbb{E}[A_1 A_2]$ does not depend on t_1

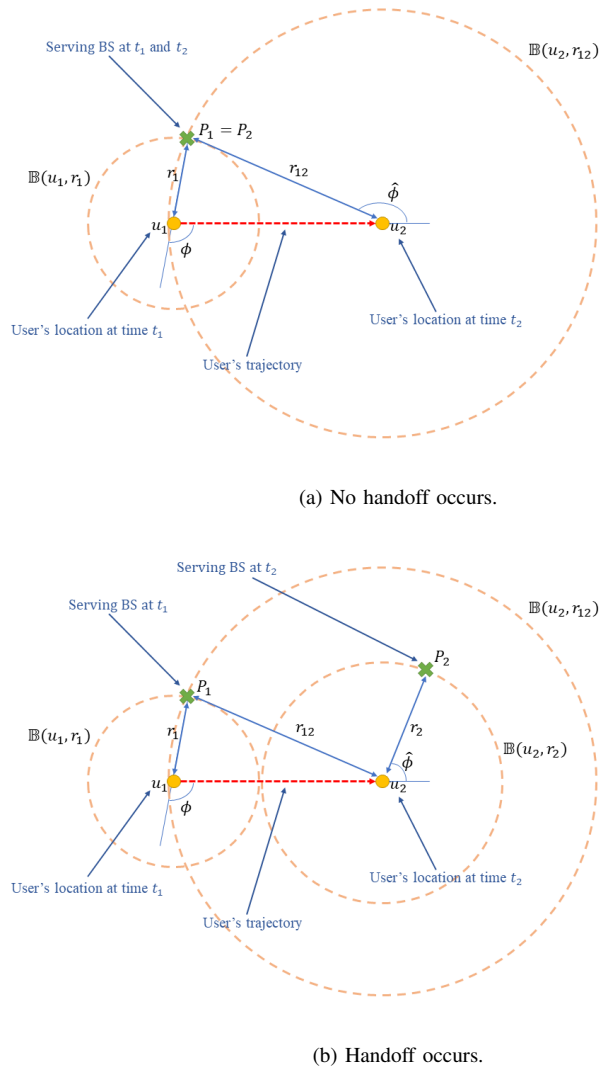


Fig. 38: Scenarios with and without handoff [77].

and t_2 ; it only depends on $\Delta = |t_2 - t_1|$. Generally, the random variables A_1 and A_2 are spatially and temporally correlated. When $u_1 = u_2$, $\mathbb{E}[A_1 A_2]$ captures the temporal correlation, i.e., the correlation at one location in two different time instants. When $t_1 = t_2$, $\mathbb{E}[A_1 A_2]$ captures the spatial correlation, i.e., correlation at two different locations at the same time.

B. Analysis of Poisson Downlink Networks

1) *Model I:* Consider a Poisson downlink network Φ with intensity λ where all the users are associated with their nearest BSs. The small-scale fading is i.i.d. across time and space. Consider a mobile user that is located at distance r_1 from its associated BS at time t_1 . The user moves away from the associated BS with a constant speed v at an angle ϕ at time t_1 , where ϕ is uniformly distributed in $[0, 2\pi]$ (Fig. 38). Before further discussion, let us introduce some notations that help

us in deriving the spatial-temporal JSP at time t_1 and t_2 with $t_2 > t_1$ [77].

- P_1, P_2 : Location of the serving BS at time t_1 and t_2 , respectively.
- r_1, r_2 : Distance between the mobile user and its associated BS at time t_1 and time t_2 , respectively.
- r_{12} : Distance between the mobile user at time t_2 and its associated BS at time t_1 .
- $\mathbb{B}(x, r)$: Closed disk with center at x and radius being r . For brevity, $\mathbb{B}(u_1, r_1)$, $\mathbb{B}(u_2, r_2)$, and $\mathbb{B}(u_2, r_{12})$ are denoted by \mathcal{B}_1 , \mathcal{B}_2 , and \mathcal{B}_{12} , respectively.
- H : Event that at least one handoff occurs in the interval $[t_1, t_2]$.

As the mobile user moves from u_1 to u_2 , the mobile user may be served by the same BS (no handoff occurs) or it may be handed off to other BSs (handoff occurs). At time t_2 , when there is no BS in \mathcal{B}_{12} , the mobile user is still connected to the same BS as time t_1 , i.e., $P_1 = P_2$. In this case, according to the definitions and from triangle equations, we have $r_2 = r_{12} = \sqrt{r_1^2 + v^2(t_2 - t_1)^2 + 2r_1v(t_2 - t_1)\cos\phi}$ and $\mathcal{B}_{12} = \mathcal{B}_2$. On the other hand, when there is a BS in \mathcal{B}_{12} , at time t_2 , the mobile user is served by a new BS, i.e., $P_1 \neq P_2$. In this case $r_2 < r_{12}$ and $\mathcal{B}_2 \subset \mathcal{B}_{12}$. In the following, we briefly provide the steps for deriving the JSP at time t_1 and t_2 . Note that since $u_1 \neq u_2$ and $t_1 \neq t_2$, the JSP captures the spatial-temporal SIR correlation.

2) *Steps to Derive the JSP [77]: Step 0*: From the contact distribution function of the homogeneous PPP, the PDF of the contact distance at time t_1 follows as (18). Moreover, due to the symmetry, we can assume ϕ is uniformly distributed in $[0, \pi]$.

Step (i): Calculating $\mathbb{P}(H | r_1, \phi)$ (handoff probability given r_1 and ϕ):

$$\begin{aligned} \mathbb{P}(H | r_1, \phi) &\stackrel{(a)}{=} 1 - \mathbb{P}(\Phi(\mathcal{B}_{12} \setminus \mathcal{B}_1) = 0 | r_1, \phi) \\ &\stackrel{(b)}{=} 1 - \exp(-\lambda|\mathcal{B}_{12} \setminus \mathcal{B}_1|), \end{aligned}$$

where (a) follows the fact that handoff occurs when there is a BS in \mathcal{B}_{12} . In (a), \mathcal{B}_1 is excluded from \mathcal{B}_{12} since there is no BS inside \mathcal{B}_1 . (b) follows from the void probability of the PPP.

Step (ii): Deriving the distribution of r_2 : Given r_1 and ϕ , when there is no handoff, $r_2 = \sqrt{r_1^2 + v^2(t_2 - t_1)^2 + 2r_1v(t_2 - t_1)\cos\phi}$. However, when handoff occurs, as discussed earlier, $r_2 < r_{12}$. In this case, we can derive the conditional PDF of r_2 as $f_{r_2}(z | H, r_1, \phi) = \frac{d}{dz}F_{r_2}(z | H, r_1, \phi)$, where $F_{r_2}(z | H, r_1, \phi)$ is the conditional CDF of r_2 and can be obtained by

$$\begin{aligned} F_{r_2}(z | H, r_1, \phi) &= \frac{\mathbb{P}(r_2 \leq z, H | r_1, \phi)}{\mathbb{P}(H | r_1, \phi)} \\ &= \frac{\mathbb{P}(\Phi(\mathbb{B}(u_2, z) \setminus \mathcal{B}_1) > 0, \Phi(\mathcal{B}_{12} \setminus \mathcal{B}_1) > 0 | r_1, \phi)}{\mathbb{P}(H | r_1, \phi)} \\ &\stackrel{(a)}{=} \frac{\mathbb{P}(\Phi(\mathbb{B}(u_2, z) \setminus \mathcal{B}_1) > 0 | r_1, \phi)}{\mathbb{P}(H | r_1, \phi)} \end{aligned}$$

$$= \frac{1 - \exp(-\lambda|\mathbb{B}(u_2, z) \setminus \mathcal{B}_1|)}{\mathbb{P}(H | r_1, \phi)}, \quad (104)$$

for $z \in [\max(0, r_1 - v), r_{12}]$. (a) follows since $\mathbb{B}(u_2, z) \subset \mathcal{B}_{12}$ in the case of handoff. When $r_1 > v$, the nearest BS is at least at distance $r_1 - v$, therefore, $F_{r_2}(z | H, r_1, \phi) = 0$ for $z < \max(0, r_1 - v)$. Moreover, as explained earlier, $F_{r_2}(z | H, r_1, \phi) = 1$ for $r_{12} < z$.

When handoff occurs, we also need the distribution of $\hat{\phi}$, angle between the vectors $\overrightarrow{u_1u_2}$ and $\overrightarrow{u_2P_2}$. Given r_1, r_2 , and v , one of the two following cases holds⁶: 1) $|r_1 - r_2| < v \leq r_1 + r_2$ (\mathcal{B}_1 and \mathcal{B}_2 overlap): $\hat{\phi}$ is uniformly distributed in $[-\pi + \arccos(\frac{r_2^2 + v^2 - r_1^2}{2r_2v}), \pi - \arccos(\frac{r_2^2 + v^2 - r_1^2}{2r_2v})]$, 2) $r_1 + r_2 < v$ (\mathcal{B}_1 and \mathcal{B}_2 are disjoint): $\hat{\phi}$ is uniformly distributed in $[0, 2\pi]$.

Step (iii): Calculating the spatial-temporal JSP $\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta)$ as in (105), where \bar{H} is the complement of H and denotes the event that handoff does not occur.

Term I in (105) can be obtained by (106), where Φ_I denotes the locations of the interferers. In the case of no handoff (\bar{H}), Φ_I is a PPP with intensity λ in $\mathbb{R}^2 \setminus (\mathcal{B}_1 \cup \mathcal{B}_{12})$. (a) is obtained using the fact that, given Φ_I, \bar{H}, r_1 , and ϕ , SIR_1 and SIR_2 are independent since small-scale fading is i.i.d. at different time instants and spatial locations. The inner expectation w.r.t. Φ_I can be calculated by using the PGFL of PPP, and the outer expectation by using PDF of r_1 and ϕ provided in **Step 0**. $\mathbb{P}(\bar{H} | r_1, \phi)$, probability that handoff does not occur, is also provided in **Step (i)**. **Term II** in (105) can be derived as (107). In the case of handoff (H), Φ_I is a PPP with intensity λ in $\mathbb{R}^2 \setminus (\mathcal{B}_1 \cup \mathcal{B}_2)$. Note that $\Phi_I \cup \{P_2\}$ is the set of interferers at time t_1 and $\Phi_I \cup \{P_1\}$ is the set of interferers at time t_2 . (a) is obtained using the fact that, given $\Phi_I, \hat{\phi}, r_2, H, r_1$, and ϕ , SIR_1 and SIR_2 are independent. The expectation w.r.t. Φ_I can be obtained by using the PGFL of PPP. The expectation w.r.t. $\hat{\phi}$ and r_2 can be obtained from **Step (ii)**. $\mathbb{P}(H | r_1, \phi)$ is provided in **Step (i)**. Outer expectation w.r.t. r_1 and ϕ can also be obtained by using PDFs given in **Step 0**.

C. Analysis of Poisson Bipolar Networks

1) *Model II*: In the second scenario, similar to [24], we consider a Poisson bipolar network, where, at time t_1 , transmitters form a PPP Φ with intensity λ . Each transmitter is assigned a unique receiver at distance r_0 at a random direction. Due to the stationarity of PPP, we investigate the performance of the typical receiver and consider its location as the origin. Since the homogeneous PPP is isotropic, without loss of generality, we can assume that the transmitter of this receiver is located at $[r_0, 0]^T$. From Slivnyak's theorem [3], at time t_1 , the other transmitters (interferers), denoted by Φ , form a homogeneous PPP with intensity λ . Assume that the selected receiver and its corresponding transmitter are static while interferers are mobile according to a uniform mobility model. Hence, at any time t , the interferers form a homogeneous PPP with intensity λ .

⁶If $v \leq |r_1 - r_2|$, no handoff occurs when the user moves from u_1 to u_2 .

$$\begin{aligned}
\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta) &= \mathbb{E} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid r_1, \phi)] \\
&= \mathbb{E} [\underbrace{\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid \bar{H}, r_1, \phi) \mathbb{P}(\bar{H} \mid r_1, \phi)}_{\text{Term I}} \\
&\quad + \underbrace{\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid H, r_1, \phi) \mathbb{P}(H \mid r_1, \phi)}_{\text{Term II}}] . \tag{105}
\end{aligned}$$

$$\begin{aligned}
&\mathbb{E} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid \bar{H}, r_1, \phi) \mathbb{P}(\bar{H} \mid r_1, \phi)] \\
&= \mathbb{E} [\mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid \Phi_I, \bar{H}, r_1, \phi)] \mathbb{P}(\bar{H} \mid r_1, \phi)] \\
&\stackrel{(a)}{=} \mathbb{E} [\mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta \mid \Phi_I, \bar{H}, r_1, \phi) \mathbb{P}(\text{SIR}_2 > \theta \mid \Phi_I, \bar{H}, r_1, \phi)] \mathbb{P}(\bar{H} \mid r_1, \phi)] , \tag{106}
\end{aligned}$$

$$\begin{aligned}
&\mathbb{E} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid H, r_1, \phi) \mathbb{P}(H \mid r_1, \phi)] \\
&= \mathbb{E} [\mathbb{E}_{r_2} [\mathbb{E}_{\hat{\phi}} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid \hat{\phi}, r_2, H, r_1, \phi)]] \mathbb{P}(H \mid r_1, \phi)] \\
&\stackrel{(a)}{=} \mathbb{E} [\mathbb{E}_{r_2} [\mathbb{E}_{\hat{\phi}} [\mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta \mid \Phi_I, \hat{\phi}, r_2, H, r_1, \phi) \mathbb{P}(\text{SIR}_2 > \theta \mid \Phi_I, \hat{\phi}, r_2, H, r_1, \phi)]]] \mathbb{P}(H \mid r_1, \phi)] . \tag{107}
\end{aligned}$$

We study the JSP at two different time slots t_1 and t_2 . Note that since the desired receiver is fixed at the origin ($u_1 = u_2$), the JSP captures the temporal correlation. Let $w_{i,\Delta}$ denote the displacement vector for transmitter i^7 . We consider a random individual mobility model (such as random walk), where each interferer moves independently of other interferers following the same random mobility model. Thus, $w_{i,\Delta}$ is independently and identically distributed for each interferer.

2) *Derivation of Spatial-temporal JSP:* We can calculate the JSP in the case of random individual mobility model where $u_1 = u_2$ for spatially and temporally i.i.d. fading as follows:

$$\begin{aligned}
&\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta) \\
&= \mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta \mid \Phi_I)] \\
&\stackrel{(a)}{=} \mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta \mid \Phi_I) \mathbb{P}(\text{SIR}_2 > \theta \mid \Phi_I)] \\
&\stackrel{(b)}{=} \mathbb{E}_{\Phi_I} [\mathbb{P}(\text{SIR}_1 > \theta \mid \Phi_I) \\
&\quad \times \mathbb{E}_{(w_{\Delta,i})} [\mathbb{P}(\text{SIR}_2 > \theta \mid \{w_{\Delta,i}\}, \Phi_I)]] , \tag{108}
\end{aligned}$$

where (a) is obtained, since given the initial location of the interferers Φ_I , SIR_1 and SIR_2 are independent. In (b), the inner expectation is w.r.t. the displacement vectors of all interferers, where $\Delta = |t_2 - t_1|$.

D. Numerical Results

To understand the effect of mobility, in this section, we consider the CSP

$$\mathbb{P}(\text{SIR}_2 > \theta \mid \text{SIR}_1 > \theta) = \frac{\mathbb{P}(\text{SIR}_1 > \theta, \text{SIR}_2 > \theta)}{\mathbb{P}(\text{SIR}_1 > \theta)} .$$

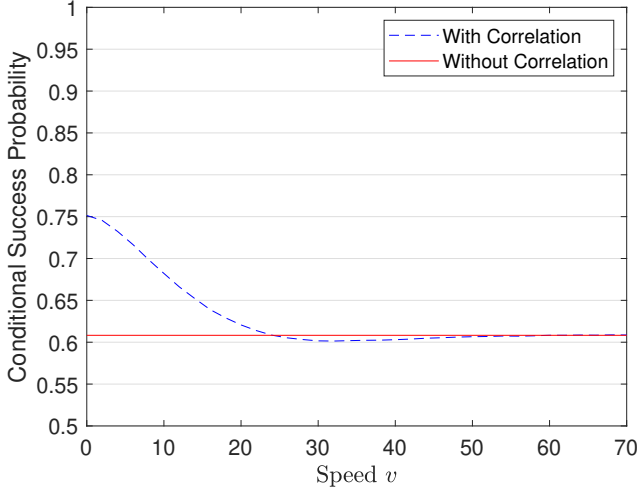
⁷ Δ reflects the fact that statistics of the displacement vector depends on the time difference $\Delta = |t_2 - t_1|$ rather than t_1 and t_2 separately.

Since the network performance depends on the nodes' displacements⁸, we set $|t_2 - t_1| = 1$ and only consider the effect of the speed on the spatial-temporal JSP.

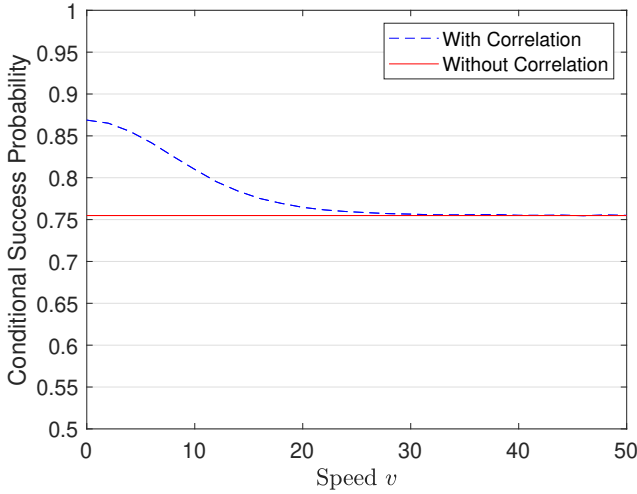
In Fig. 39, we illustrate the effect of correlation on the CSP for both scenarios. For the random individual mobility model, we have assumed all the mobile users move with the same speed v , and each mobile user moves in a random independent direction φ_i , uniformly distributed in the interval $[0, 2\pi]$. To better understand the impact of correlation, in Fig. 39, we also compare the results with the independent case, where the CSP is equal to $\mathbb{P}(\text{SIR}_2 > \theta)$. According to the displacement theorem [3], at any time t , interferers form a PPP with intensity λ . Thus, in Model II, $\mathbb{P}(\text{SIR}_{u_2}(t) > \theta)$ does not depend on t (and consequently v). Using the stationarity of the PPP, we can obtain the same result for Model I.

As shown in Fig. 39, at high speeds, we can ignore the effect of correlation, and assume that the two sets of interferers at time instants t_1 and t_2 are independent. As a result, most of the existing works in mobile networks only focus on the highly mobile scenario where they can ignore the effect of correlation. It is worthwhile to note that the gap between the results with correlation (dashed blue curves) and without correlation (solid red lines) is proportional to the correlation coefficient between the success event at t_1 and success event at t_2 . As illustrated in Fig. 39, the correlation between success events at t_1 and t_2 decreases as the speed increases.

⁸The displacement of a node that moves at a speed v for a time duration $|t_2 - t_1| = \Delta$ is the same as when it moves with speed $v\Delta$ for duration $|t_2 - t_1| = 1$.



(a) Model I.



(b) Model II.

Fig. 39: Simulation of the effect of speed on the spatial-temporal JSP for Model I in (a) and Model II in (b) for $\lambda = 0.001$ and $\theta = -1$ dB. In Model II, the desired link distance is set to $d = 8$.

E. Summary and Discussion

This section has characterized the spatial-temporal JSP of two network scenarios 1) with mobile users and static interferers and 2) with mobile interferers and mobile users. The main lessons learned are as follows.

- The spatial-temporal correlation among different transmission events decreases with speed and can be ignored when the speed is high enough.
- The gap between the spatial-temporal CSP is proportional to the correlation coefficient between successful transmission events.

In Poisson networks, mobility introduces spatial randomness which decreases the correlation across time [77]; thus, spatial diversity provides time diversity when nodes are mobile. This

is also discussed in the literature by studying the mean local delay [127]. In a static network, when there is a close interferer, retransmitting an unsuccessfully received signal is usually not helpful. As a result, in static networks, the mean local delay may be infinite, i.e., a significant fraction of users suffer from large delays. However, in mobile networks, interference correlation decreases across time and there is a higher chance to receive a retransmitted signal successfully. Therefore, mobility reduces the mean local delay by providing spatial diversity. Specifically, in the highly mobile scenario, the mean local delay is always finite.

Open Technical Issues: Analyzing the performance of mobile wireless networks with correlation effects is challenging for most scenarios and usually yields complicated results. Consequently, most of the related works consider simple system models such as downlink communication in a Poisson cellular network (as in Model I) or Poisson bipolar networks with mobile interferers (as in Model II), but even for the latter model, there are no analytical results.

In the downlink communication in Model I, we have studied the performance of a mobile node in a network of static transmitters (i.e., BSs). In contrast, in the uplink, we have mobile transmitters (the desired transmitter and interferers). Due to the random mobility of the nodes and irregular shapes of the cells, it may not be possible to calculate the load distribution while considering the correlation effect, which in turn, makes the analysis of the uplink scenario very challenging. In fact, an exact analysis is impossible, but approximations, bounds, or asymptotic results can almost always be derived.

In Model II, as shown in (108), we need the distribution of the displacement vector w_Δ . However, it may not be possible to derive this distribution for most of the mobility models and for all values of Δ . Therefore, most of the existing works resort to a simplified mobility model or consider only one movement step during which the mobile node moves along a straight line.

In addition to the spatial-temporal JSP, the SIR meta distribution in cellular networks with mobile BSs has been studied in [128]. The joint SIR meta distribution and product SIR meta distribution in mobile networks are also worth investigating.

X. FUTURE DIRECTIONS

In this section, we discuss future applications of stochastic geometry analysis and related technical challenges.

A. Novel Performance Metrics and Stochastic Geometry Analysis

Although the existing literature has extensively investigated different types of large-scale communication networks, the majority of them focus on the spatial and temporal average performance, such as mean success probability, delay and throughput. Recently, the SIR meta distribution [79] has been introduced to provide insights on the distribution of success probability over space and/or time. The meta distribution has

been extended to evaluate the distributions of rate and energy [129], mean delay [30], transmission rate [130], secrecy rate [131], and secrecy success probability [43].

The concept of meta distribution can also be applied to evaluate the distribution of other performance metrics such as the covert probability for low probability of detection/intercept communications [132], [133]. Moreover, the age-of-information (AoI) [134], [135] that quantifies the time elapsed from the generation to observation of the information explicitly measures the information freshness that the conventional metrics (e.g., link delay) do not. Some recent works have shown that network protocol designs based on AoI provide better consumer service quality in real-time status updating, such as Google Scholar [136]. To enhance information freshness, the existing network protocols need to be re-examined and stochastic geometry analysis can be exploited in understanding AoI in large-scale systems. Characterization of the above-mentioned performance metrics by considering the spatial and temporal correlations introduced in this tutorial opens a broad research area.

B. Stochastic Geometry Analysis for Cross-layer Study

The majority of the existing literature on stochastic geometry analysis assumes that the transmission demand is uniformly distributed and regulated by some medium access control (MAC) protocols such as Aloha or CSMA. However, in practice, the MAC-level transmission scheduling can be affected by application-level service scheduling [137]. For example, in a mobile edge computing system where the mobiles can offload computation tasks to the edge, different requests can be scheduled to different edge servers based on the quality-of-service requirements of the tasks as well as the functionality and the status of the servers. In such a scenario, to analyze the communications performances for the offloaded tasks in a large-scale system, stochastic geometry can be used. For such an analysis, characterizing the point process corresponding to the interferers would be challenging due to the dynamic resource scheduling introduced by the cross-layer design. In general, modeling and analysis of cross-layer effects is an important research direction.

C. Machine Learning Approaches for Large-Scale Networks and Stochastic Geometry Analysis

Machine learning (ML) techniques have been increasingly adopted for resource allocation and control in wireless networks due to their capabilities in learning network environment variations (e.g., due to traffic pattern and channel uncertainties), classification of the relevant quantities, predicting future outcomes [138]. For instance, deep reinforcement learning approaches are capable of estimating channel state information (CSI) without explicit feedback/detection and thus can be exploited to design resource management schemes with low communication and computation overhead compared to conventional CSI-based schemes [139], [140]. With ML-based designs, network resource allocations will be more intelligent and environment-aware. Stochastic geometry tools can

be used to analyze the effects of correlation introduced by environment-aware resource allocation in large-scale systems. Also, stochastic geometry can be used to study the impact of network interference as well as interference correlation on the performance of ML-based designs (e.g., in terms of stability, convergence, and accuracy).

D. Stochastic Geometry Analysis of Emerging Network Scenarios

1) *Programmability of Radio Environment with Reconfigurable Intelligent Surfaces: A Reconfigurable Intelligent Surface (RIS)* [141], [142], also known as *software-defined hypersurface* and *large intelligent surface/antennas*, is a digitally-tunable metasurface. A metasurface is a thin and planar structure composed of sub-wavelength passive scattering particles which alter the electromagnetic waves through a surface impedance boundary condition. The electromagnetic response of an RIS can be reprogrammed without re-fabrication. Due to this distinguishing feature, an RIS can be utilized to dynamically alter the electromagnetic behavior (e.g., amplitude, phase, and frequency [143]) of the incident signals, and thus reshaping the wireless propagation environments.

RIS has been envisioned as a candidate technology for future generation networks [144] and expected to be widely deployed on, e.g., surfaces of buildings, vehicles and billboards, to facilitate the ever-increasing traffic demands. In such an environment, the signal distributions are jointly manipulated by multiple RISs. The incident signals at nearby RISs come from common sources and exhibit spatial correlations. Characterization of the aggregate effect of large-scale RIS deployments by taking into account the spatial correlation is an open research problem.

2) *Joint Radar and Communication*: Joint radar and communication (JRC) [145] is a novel paradigm that facilitates radar detection and data communication over the same frequency bands to improve the spectrum efficiency. In JRC systems, a dual-functional transmitter (e.g., autonomous cars or unmanned drones) simultaneously transmits data and detects radar targets (e.g., barriers, mobile vehicles and pedestrians) based on a shared and integrated hardware platform. Specifically, a portion of the transmitted signal received at the target receiver is used for communication, while the other portion of the signal reflected from the radar targets at the dual-function transmitter is utilized for detection [146]. Due to the common signal source and interferers, the communication and radar performance in JRC systems are correlated. Furthermore, to accommodate both radar and communication demands, JRC brings a series of resource allocation problems, such as power allocation, spectrum allocation and sharing, and beamforming design. Analyzing radar and communication performance in large-scale systems based on stochastic geometry analysis under different resource allocation schemes is a promising future research direction.

3) *Internet of Space Things*: The Internet of Space Things (IoST) [147] is an expansion of the ground Internet of Things (IoT) to the aerial and space domains enabled by drones

and miniaturized satellites/CubeSats [148], respectively. With the holistic integration of the ground with aerial and space communication systems, IoST is expected to realize ubiquitous connectivity virtually to support a broad range of applications including monitoring, tracking, in-space backhauling [149], wireless power transfer [150], [151] and remote healthcare solutions. Performance characterization of IoST requires efforts of developing three-dimensional modeling incorporating disparate propagation characteristics, mobility patterns, resource allocation schemes of communication systems as well as their intricate correlations at different domains.

4) *Terahertz Communications*: The terahertz band (i.e., 0.1-10 THz) provides ultra-wide spectrum resources and enable new applications for beyond 5G [152]. For example, terabits per second (Tbps) data rate can be realized to support “fiber-like” communication performance which offers a seamless transition between optical fibers and THz wireless links with no latency. Moreover, the micro-scale wavelength makes THz band more suitable for *in-vivo* nano-network communications than any other frequency bands because terahertz waves have strong penetrating force and can be absorbed by in-vivo substances such as liquids and organs [153]. The availability of new spectrum bands will necessitate novel resource allocation designs. An intriguing research direction is to develop stochastic geometry models that take into account the correlation effects from: i) physical blockages, ii) direction of arrival/departure of high-gain beam steering, and iii) temporal broadening effect resulted from frequency selectivity in the ultra-wide THz band.

5) *Internet of Nanothings (IoNT)*: Internet of Nanothings (IoNT) is a connected molecular system empowered by nanomachines, which are microscopic devices capable of performing high-precision functions for many real-world applications such as biomedicine, food industry and military operations [154]. As IoNT is expected to work in the environments of fluids, gases or particulates, IoNT will foreseeably function in a manner that is drastically different from the IoT due to differences in propagation environments, the scale of deployment, as well as physical constraints (e.g., in energy and computations) of miniaturized nano devices. Therefore, it is imperative to develop novel mathematical models to characterize the features of IoNT. Moreover, the protocol stack design and analysis for large-scale IoNT still remains an open field for exploration.

XI. SUMMARY

Stochastic geometry tools can be used to develop analytical frameworks for large-scale wireless systems considering the effects of spatial-temporal interference correlation, which is generally ignored in traditional performance analyses. In this tutorial, we have presented a comprehensive spatial-temporal analysis of large-scale communications systems. In particular, we have formulated models to characterize correlations in interference (and hence SIR) due to different effects such as distributions of the interferers, distribution of contact distance, shadowing, transmission buffer status (or network queues), multihop transmissions, retransmissions, and user mobility.

The performance of a target link in a large-scale system has been demonstrated both analytically and numerically, considering the effect of spatial-temporal interference correlation in different scenarios. In particular, we have derived the joint distribution of spatially-correlated SIR with multihop relaying, temporally-correlated SIR with retransmissions and spatially and temporally-correlated SIR with mobility. For each of the above scenarios, we have shed light on the technical challenges in stochastic geometry analysis. Finally, we have presented future research directions in stochastic geometry analysis of emerging wireless communication systems the performance of which will be affected by spatial-temporal interference correlations.

APPENDIX

A. Proof of Theorem 2 (Moments of the CSP_Φ in Random Fields of Interferers)

Proof. Given Φ , the CSP of a link with transmission distance r_t can be derived as

$$\begin{aligned}\bar{F}_{\eta|\Phi}(\theta) &= \mathbb{P}\left[h_t > \theta r_t^\alpha \sum_{j \in \mathbb{N}} h_j \|x_j\|^{-\alpha} \mid \Phi\right] \\ &= \mathbb{E}_{(h_j)} \left[\exp\left(-\theta r_t^\alpha \sum_{j \in \mathbb{N}} h_j \|x_j\|^{-\alpha}\right) \mid \Phi \right] \\ &= \prod_{j \in \mathbb{N}} \frac{1}{1 + \theta r_t^\alpha r_j^{-\alpha}}.\end{aligned}\quad (109)$$

The moments of the CSP given Matérn and Poisson fields of interferers can be obtained by applying their PGFLs as in (110), shown on the top of the next page, where (a) applies the PGFLs of MCP and PPP given in (34) and (23), respectively. By inserting $f^M(y)$ given in (31) in (110), the final result with a Matérn cluster field of interferers in (49) directly yields.

We further extend the expression for the PPP in (110) as

$$\begin{aligned}\mathcal{M}_{P_s}(b) &\stackrel{(b)}{=} \exp\left(-\pi\lambda\delta \int_0^\infty \left(1 - \left(1 - \frac{\theta r_t^\alpha}{z + \theta r_t^\alpha}\right)^b\right) z^{\delta-1} dz\right) \\ &\stackrel{(c)}{=} \exp\left(-\pi\lambda\delta \sum_{k=1}^\infty \binom{b}{k} (-1)^{k+1} \theta^k r_t^{\alpha k} \right. \\ &\quad \left. \times \int_0^\infty \frac{z^{\delta-1}}{(z + \theta r_t^\alpha)^k} dz\right),\end{aligned}\quad (111)$$

where (b) employs the change of variable $z = r_t^2 \theta^{-\delta} x^2$ and the substitution $\delta = \frac{2}{\alpha}$, (c) takes the binomial expansion of $\left(1 - \frac{\theta r_t^\alpha}{z + \theta r_t^\alpha}\right)^b$ for $b \in \mathbb{C}$.

The integral in (111) can be transformed as follows [156, Eq. 3.196.2]

$$\begin{aligned}&\int_0^\infty \frac{z^{\delta-1}}{(z + \theta r_t^\alpha)^k} dz \\ &= \theta^{\delta-k} r_t^{2-\alpha k} \frac{\Gamma(k-\delta)\Gamma(\delta)}{\Gamma(k)}\end{aligned}$$

$$\mathcal{M}_{P_s}(b) = \mathbb{E} \left[\prod_{j \in \mathbb{N}} \left(\frac{1}{1 + \theta \|x_t\|^\alpha \|r_j\|^{-\alpha}} \right)^b \right] \\ \stackrel{(a)}{=} \begin{cases} \exp \left(-\lambda_p \int_{\mathbb{R}^2} \left[1 - \exp \left(-\bar{c} \left(1 - \int_{\mathbb{R}^2} \left(\frac{1}{1 + \theta r_t^\alpha \|x - y\|^{-\alpha}} \right)^b f^M(y) dy \right) \right) \right] dx \right), & \text{MCP} \\ \exp \left(-\lambda \int_{\mathbb{R}^2} \left(1 - \left(\frac{1}{1 + \theta r_t^\alpha \|x\|^{-\alpha}} \right)^b \right) dx \right), & \text{PPP,} \end{cases} \quad (110)$$

$$\stackrel{(d)}{=} \theta^{\delta-k} r_t^{2-\alpha k} \frac{(-1)^{k+1} \pi \csc(\pi \delta) \Gamma(\delta)}{\Gamma(\delta - k + 1) \Gamma(k)} \\ = \theta^{\delta-k} r_t^{2-\alpha k} (-1)^{k+1} \pi \csc(\pi \delta) \frac{(\delta - 1) \dots (\delta - k + 1)}{(k - 1)!} \\ = \theta^{\delta-k} r_t^{2-\alpha k} (-1)^{k+1} \pi \csc(\pi \delta) \binom{\delta - 1}{k - 1}, \quad (112)$$

$$\stackrel{(a)}{=} \mathbb{E} \left[\exp \left(-\theta \sum_{j=2}^{\infty} h_j \varrho_j^\alpha \right) \mid \Phi_c^R \right] \\ \stackrel{(b)}{=} \prod_{\substack{j=2 \\ \varrho_2 \leq \rho}}^{\infty} \frac{1}{1 + \theta \varrho_j^\alpha}, \quad (114)$$

where (d) follows as

$$\Gamma(k - \delta) = \frac{\pi \csc(\pi(k - \delta))}{\Gamma(\delta - k + 1)} = \frac{(-1)^{k+1} \pi \csc(\pi \delta)}{\Gamma(\delta - k + 1)}.$$

Inserting (112) into (111) yields,

$$\mathcal{M}_{P_s}(b) = \exp \left(-\lambda \theta^\delta r_t^2 \pi \delta \csc(\pi \delta) \sum_{k=1}^{\infty} \binom{b}{k} \binom{\delta - 1}{k - 1} \right) \\ \stackrel{(e)}{=} \exp \left(-\lambda \theta^\delta r_t^2 \frac{\Gamma(1 - \delta) \Gamma(b + \delta)}{\Gamma(b)} \right), \quad (113)$$

where (e) follows as $\pi \delta \csc(\pi \delta) = \Gamma(1 - \delta) \Gamma(\delta)$ and $\sum_{k=1}^{\infty} \binom{b}{k} \binom{\delta - 1}{k - 1} = \frac{\Gamma(b + \delta)}{\Gamma(b) \Gamma(1 + \delta)}$.

With a β -Ginibre field of interferers, the success probability can be obtained from (109) as

$$\mathcal{M}_{P_s}(b) \stackrel{(e)}{=} \mathbb{E}_{Q_j} \left[\prod_{j \in \mathbb{N}} \left(\frac{\beta}{1 + \theta r_t^\alpha Q_j^{-\alpha/2}} + 1 - \beta \right)^b \right] \\ \stackrel{(f)}{=} \prod_{j \in \mathbb{N}} \int_0^\infty \frac{(\pi \lambda / \beta)^j}{\Gamma(j)} q^{j-1} e^{-\pi \lambda q / \beta} \\ \times \left(\frac{\beta}{1 + \theta r_t^\alpha q^{-\alpha/2}} + 1 - \beta \right)^b dq,$$

where (e) adopts the substitution $Q_j = r_j^2$ and (f) follows from $Q_j \sim \mathcal{G}(j, \beta / \pi \lambda)$.

Summarizing the above results, we have the final results in Theorem 2. \square

B. Proof of Theorem 3 (Moments of the CSP_Φ of the Typical Cell-center User in a Poisson Downlink Network)

Proof. Let $\Phi_c \triangleq \{\Phi \mid o \in \mathcal{R}_c\}$. Given a Poisson point process Φ_c , the CSP of the typical cell-center user is

$$\bar{F}_{\eta|\Phi_c}(\theta) \\ = \mathbb{P}[h > \theta \|x_1\|^\alpha I_o \mid \Phi_c] \\ = \mathbb{E} \left[\exp \left(-\theta \|x_1\|^\alpha \sum_{j=2}^{\infty} h_j \|x_j\|^{-\alpha} \right) \mid \Phi_c \right]$$

where (a) changes the PPP to its RDP, i.e., $\rho_j = \|x_1\| / \|x_j\|$, and the condition $\varrho_2 \leq \rho$ in (b) implies $o \in \mathcal{R}_c$.

Then, the b -th moments of the CSP_Φ in a Poisson downlink network can be derived as

$$\mathcal{M}_{P_s}^c(b) \\ = \mathbb{E} \left[\prod_{j=2}^{\infty} \left(\frac{1}{1 + \theta \varrho_j^\alpha} \right)^b \mid \varrho_2 \leq \rho \right] \\ = \mathbb{E} \left[\prod_{j=2}^{\infty} \left(\frac{1}{1 + \theta \rho^\alpha \left(\frac{\varrho_j}{\rho} \right)^\alpha} \right)^b \mid \varrho_2 \leq \rho \right] \\ \stackrel{(b)}{=} \mathbb{E} \left[\prod_{j=2}^{\infty} \left(\frac{1}{1 + \theta \rho^\alpha \varrho_j^\alpha} \right)^b \right] \\ \stackrel{(c)}{=} \left(1 + 2 \int_0^1 \left(1 - \left(1 - \frac{\theta \rho^\alpha \varrho^\alpha}{1 + \theta \rho^\alpha \varrho^\alpha} \right)^b \varrho^{-3} d\varrho \right)^{-1} \right) \\ \stackrel{(d)}{=} \left(1 - 2 \sum_{k=1}^{\infty} \binom{b}{k} (-\theta \rho^\alpha)^k \right. \\ \times \left. \int_0^1 \left(\frac{\varrho^\alpha}{1 + \theta \rho^\alpha \varrho^\alpha} \right)^k \varrho^{-3} d\varrho \right)^{-1} \\ \stackrel{(e)}{=} \left(1 - 2 \sum_{k=1}^{\infty} \binom{b}{k} \frac{(-\theta \rho^\alpha)^k}{\alpha} \right. \\ \times \left. \int_0^1 \left(\frac{x}{1 + \theta \rho^\alpha x} \right)^k x^{-\frac{2}{\alpha} - 1} dx \right)^{-1} \\ \stackrel{(f)}{=} \left(1 - \sum_{k=1}^{\infty} \binom{b}{k} (-\theta \rho^\alpha)^k \frac{\delta}{k - \delta} \right. \\ \times \left. {}_2F_1(k, k - \delta; k + 1 - \delta; -\theta \rho^\alpha) \right)^{-1}, \quad (115)$$

where (b) holds due to the fact that the probability law of ϱ_2 conditioned on the probability law of $\varrho_2, \varrho_3, \dots$ conditioned on $\varrho_2 < \rho$ is the same as the law of $\varrho_2, \varrho_3, \dots$ without conditioning and (c) applies the PGFL of the RDP of a PPP, (d) takes the binomial expansion of $(1 - \frac{\theta \rho^\alpha \varrho^\alpha}{1 + \theta \rho^\alpha \varrho^\alpha})^b$ for $b \in \mathbb{C}$,

(e) applies the change of variable $x = \varrho^\alpha$, and (f) substitutes $\frac{2}{\alpha}$ with δ .

After some mathematical manipulations of (115), we have the final result in Theorem 3. \square

C. Proof of Theorem 4 (Moments of the CSP $_\Phi$ for the Typical Vertex User in a Poisson Downlink Network)

Proof. Based on the stationarity of a PPP, we have the success probability of the typical vertex user conditioned on Φ as

$$\begin{aligned} \bar{F}_{\eta|\Phi}^v(\theta) &= \mathbb{E}[e^{-\theta r_1^\alpha I_o} \mid r_1 = r_2 = r_3, \Phi] \\ &= \mathbb{E}\left[\exp\left(-\theta r_1^\alpha \left(h_1 r_2^{-\alpha} + h_2 r_3^{-\alpha} + \sum_{j=4}^{\infty} h_j r_j^{-\alpha}\right)\right) \mid \Phi\right] \\ &= (1+\theta)^{-2} \prod_{j=4}^{\infty} \frac{1}{1 + \theta r_1^\alpha r_j^{-\alpha}}. \end{aligned} \quad (116)$$

With $\bar{F}_{\eta|\Phi}^v(\theta)$ in (116), the moments of the CSP $_\Phi$ can be derived as

$$\begin{aligned} \mathcal{M}_{P_s}^v(b) &= (1+\theta)^{-2b} \mathbb{E}\left[\prod_{j=4}^{\infty} \left(\frac{1}{1 + \theta \|x_1\|^\alpha \|x_j\|^{-\alpha}}\right)^b\right] \\ &\stackrel{(a)}{=} (1+\theta)^{-2b} \\ &\quad \times \mathbb{E}_{r_1}\left[\exp\left(-2\pi\lambda \int_{r_1}^{\infty} \left(1 - \left(\frac{1}{1 + \theta r_1^\alpha x^{-\alpha}}\right)^b\right) x dx\right)\right] \\ &\stackrel{(b)}{=} (1+\theta)^{-2b} 2\pi^2 \lambda^2 \int_0^{\infty} \exp(-\lambda\pi r^2) r^3 \\ &\quad \times \exp\left(-2\pi\lambda \int_r^{\infty} \left(1 - \left(\frac{1}{1 + \theta r^\alpha x^{-\alpha}}\right)^b\right) x dx\right) dr, \end{aligned} \quad (117)$$

where (a) applies [155, Lemma 1] and the PGFL of a PPP, and (b) plugs in the PDF $f_{r_1}^v(r)$ given in (57).

After some mathematical manipulations of (117), we have the final results in Theorem 4. \square

D. Proof of Theorem 5 (Laplace Transform of Interference with Independent and Correlated Shadowing)

Proof. We first derive the Laplace transform of the interference with correlated shadowing as follows.

$$\begin{aligned} \mathcal{L}_{I_o^{\text{Cor}}}(s) &= \mathbb{E}[\exp(-s I_o^{\text{Cor}})] \\ &= \mathbb{E}\left[\exp\left(-s \sum_{j \in \mathbb{N}} h_j \ell(\|x_j\|) S_j\right)\right] \\ &\stackrel{(a)}{=} \mathbb{E}\left[\sum_{k \in \mathbb{N}} \sum_{x_j \in \mathbb{S}_k} \frac{1}{1 + s \ell(\|x_j\|) T_k}\right] \end{aligned}$$

$$\stackrel{(b)}{=} \mathbb{E}_{(T_k)} \left[\exp\left(-\lambda \sum_{k \in \mathbb{N}} \int_{\mathbb{S}_k} \left(1 - \frac{1}{1 + s \ell(\|x_j\|) T_k}\right) dx\right)\right],$$

where (a) follows as $h_j \sim \mathcal{E}(1)$ and the points within the same cell share the same shadowing coefficient, and (b) applies the PGFL of a PPP and holds as the accumulative interference of each shadowing cell \mathbb{S}_k is averaged over the same random variable T_k for $x_j \in \mathbb{S}_k$.

Similarly, the Laplace transform of I_o^{Ind} is given by

$$\begin{aligned} \mathcal{L}_{I_o^{\text{Ind}}}(s) &= \mathbb{E}[\exp(-s I_o^{\text{Ind}})] \\ &= \mathbb{E}\left[\exp\left(-s \sum_{j \in \mathbb{N}} h_j \ell(\|x_j\|) S_j\right)\right] \\ &\stackrel{(c)}{=} \exp\left(-\lambda \sum_{k \in \mathbb{N}} \int_{\mathbb{S}_k} \left(1 - \mathbb{E}_{T_k} \left[\frac{1}{1 + s \ell(\|x_j\|) T_k}\right]\right) dx\right), \end{aligned}$$

where (c) follows as S_j is an i.i.d mark for all $x_j \in \mathbb{S}_k$ with independent shadowing, with CDF F_k , as T_k . \square

E. Proof of Theorem 7 (Success Probability in a Poisson Downlink Network with Unsaturated Queues)

Proof. To bypass the difficulty of modeling coupled queue statuses, we assume the typical receiver observes temporally-independent interference across different time slots. If each interfering BS is active with independent probability $p_A = \mathbb{E}[\iota_j]$, the success probability can be derived as

$$\begin{aligned} P_s &= \mathbb{P}[\eta > \theta] \\ &\stackrel{(a)}{\approx} \mathbb{E}\left[\frac{h_1 \|x_1\|^{-\alpha}}{\sum_{j=2}^{\infty} \iota_j h_j \|x_j\|^{-\alpha}} > \theta\right] \\ &\stackrel{(b)}{=} \mathbb{E}\left[\prod_{j=2}^{\infty} \left(\frac{p_A}{1 + \theta \|x_1\|^\alpha \|x_j\|^{-\alpha}} + 1 - p_A\right)\right] \\ &\stackrel{(c)}{=} \mathbb{E}\left[\prod_{j=2}^{\infty} \left(\frac{p_A}{1 + \theta \varrho_j^\alpha} + 1 - p_A\right)\right] \\ &\stackrel{(d)}{=} \frac{1}{1 + 2 \int_0^1 \frac{p_A \theta \varrho^\alpha}{1 + \theta \varrho^\alpha} \varrho^{-3} d\varrho} \\ &\stackrel{(e)}{=} \frac{1}{1 + \delta p_A \theta \int_0^1 \frac{x}{1 + \theta x} x^{-\delta-1} dx} \\ &= \frac{1}{1 + \frac{\delta p_A \theta}{1 - \delta} {}_2F_1(1, 1 - \delta; 2 - \delta; -\theta)} \\ &\stackrel{(f)}{=} \frac{1}{1 - p_A + p_A {}_2F_1(1, -\delta; 1 - \delta; -\theta)}, \end{aligned} \quad (118)$$

where (a) applies the independent interference assumption, (b) follows as $h_1, h_j \sim \mathcal{E}(1)$, (c) changes the PPP to its RDP, (d) applies the PGFL of an RDP given in (24), (e) changes the variable $x = \varrho^\alpha$ and substitutes $\frac{2}{\alpha}$ with δ .

As can be seen from (118), p_A is the only unknown parameter. To obtain p_A , we need to analyze the probability that each BS is active from the perspective of queue dynamics. As a BS sends out a packet once its selected queue is not empty,

the active probability of the BS is equivalent to the utilization of its queue. Since the service rate of a BS is equivalent to the success probability, with random scheduling, we have the service rate of the typical queue at a BS conditioned on Φ_B as

$$\mu = \frac{1}{N_u} \mathbb{P}[\eta > \theta \mid \Phi_B], \quad (119)$$

where N_u denotes the number of users associated with the typical BS.

According to the law of total probability, the mean service rate of any user's queue at its serving BS is calculated as

$$\begin{aligned} \bar{\mu} &= \sum_{n=1}^{\infty} p_{N_u}(n) \mu \\ &= \sum_{n=1}^{\infty} p_{N_u}(n) \frac{\mathbb{P}[\eta > \theta]}{n}, \end{aligned} \quad (120)$$

where $\mathbb{P}[\eta > \theta]$ has been obtained in (118), $p_{N_u}(n)$ denotes the PDF of N_u . In a Poisson downlink network, $p_{N_u}(n)$ is approximated as [5]

$$p_{N_u}(n) \approx \frac{\nu^\nu \Gamma(n + \nu) (\lambda_u / \lambda_B)^n}{n! \Gamma(\nu) (\lambda_u / \lambda_B + \nu)^{n+\nu}},$$

where $\nu = 3.5$.

From (118) and (120), the mean service rate $\bar{\mu}$ is a function of active probability p_A . Based on the random scheduling, the active probability of a BS equals the utilization factor of the queue for the typical general user. Thus, we can establish a fixed-point equation as follows:

$$p_A = \frac{\xi_u}{\bar{\mu}} = \frac{\xi_u}{\sum_{n=1}^{\infty} p_{N_u}(n) P_s / n}.$$

Subsequently, we have the success probability of a Poisson downlink network with unsaturated queues given in Theorem 7. \square

F. Proof of Theorem 2 (Success Probability in a Poisson Bipolar Network)

Proof. Using the assumption that each transmitter is active with independent probability, the success probability in Poisson bipolar networks is given by

$$\begin{aligned} P_s &\stackrel{(a)}{\approx} \mathbb{E} \left[\frac{h_t \|x_t\|^{-\alpha}}{\sum_{j \in \mathbb{N} \setminus \{\zeta(x_t)\}} \iota_j h_j \|x_j\|^{-\alpha}} > \theta \right] \\ &\stackrel{(b)}{=} \mathbb{E} \left[\prod_{j \in \mathbb{N}} \left(\frac{p_A}{1 + \theta \|x_t\|^\alpha \|x_j\|^{-\alpha}} + 1 - p_A \right) \right] \\ &\stackrel{(c)}{=} \exp \left(-2\pi\lambda p_A \int_0^\infty \frac{\theta r_t^\alpha x^{-\alpha+1}}{1 + \theta r_t^\alpha x^{-\alpha}} dx \right) \\ &\stackrel{(d)}{=} \exp \left(-\pi\lambda p_A \delta \int_0^\infty \frac{\theta r_t^\alpha u^{\frac{2}{\alpha}-1}}{u + \theta r_t^\alpha} du \right) \\ &\stackrel{(e)}{=} \exp \left(-p_A \lambda \pi r_t^2 \theta^\delta \Gamma(1 + \delta) \Gamma(1 - \delta) \right), \end{aligned} \quad (121)$$

where (a) applies the independent interference assumption, (b) holds due to Slivnyak's theorem, (c) applies the PGFL of a PPP

and employs the substitution $\delta = 2/\alpha$, (d) applies the change of variable $u = x^\alpha$, and (e) follows from [156, Eq. 3.196.2] that $\int_0^\infty \frac{u^{\delta-1}}{u+A} du = A^\delta \Gamma(1 - \delta) \Gamma(\delta)$.

Since the service rate of the typical transmitter is equivalent to the success probability of a time slot, p_A can be written as $\min\{\frac{\xi}{P_s}, 1\}$ which is a well-known result from Geo/Geo/1 queue [157]. Subsequently, we can establish the following equation.

$$\begin{aligned} &\min \left\{ \xi / P_s, 1 \right\} \\ &= \min \left\{ \frac{\xi}{\exp(-p_A \lambda \pi r_t^2 \theta^\delta \Gamma(1 + \delta) \Gamma(1 - \delta))}, 1 \right\} \\ &= p_A. \end{aligned} \quad (122)$$

Since P_s is a function of P_A as shown in (121), solving the equation in (122) yields

$$p_A = \min \left\{ -\frac{\mathcal{W}(-\xi \lambda \pi r_t^2 \theta^\delta \Gamma(1 + \delta) \Gamma(1 - \delta))}{\lambda \pi r_t^2 \theta^\delta \Gamma(1 + \delta) \Gamma(1 - \delta)}, 1 \right\}, \quad (123)$$

where \mathcal{W} denotes the Lambert- \mathcal{W} function [158].

By inserting p_A in (123) into the expression of $P_s = \frac{\xi}{p_A}$, we have the final results presented in Theorem 8. \square

G. Proof of Theorem 9 (Moments of the End-to-End CSP $_\Phi$)

Proof. From (86), the end-to-end CSP $_\Phi$ of M -hop relaying given a Poisson field of interferers can be expressed as

$$\begin{aligned} &\bar{F}_{\eta|\Phi}(\theta) \\ &= \mathbb{P}[h_1 > \theta d_1^\alpha I_1, h_2 > \theta d_2^\alpha I_2, \dots, h_M > \theta d_M^\alpha I_M \mid \Phi_m] \\ &= \mathbb{E}_{(h_{j,m})} \left[\exp \left(-\theta \sum_{m=1}^M \sum_{j \in \mathbb{N}} \frac{h_{j,m} d_m^\alpha}{\|x_{j,m} - y_m\|^\alpha} \right) \mid \Phi_m \right], \\ &= \prod_{j \in \mathbb{N}} \prod_{m=1}^M \frac{1}{1 + \theta d_m^\alpha \|x_{j,m} - z_m\|^{-\alpha}}. \end{aligned} \quad (124)$$

With the end-to-end success probability with a Poisson field of interferers in (124), we further compute the b -th moment of it as in (125), where (a) follows as $h_{j,m} \sim \mathcal{E}(1)$ and (b) applies the PGFL of a PPP. \square

H. Proof of Theorem 10 (JSP of Multiple Transmissions)

Proof. The JSP for K transmissions is given as

$$\begin{aligned} \mathcal{J}_K &= \mathbb{P} \left[\bigcap_{k=1}^K \left\{ h^{(k)} > \theta r_t^\alpha I_o^{(k)} \right\} \right] \\ &\stackrel{(a)}{=} \mathbb{E} \left[\prod_{k=1}^K \exp \left(-\theta r_t^\alpha I_o^{(k)} \right) \right] \\ &= \mathbb{E} \left[\prod_{k=1}^K \exp \left(-\theta r_t^\alpha \sum_{j \in \mathbb{N}} h_j^{(k)} \|x_j^{(k)}\|^{-\alpha} \right) \right], \end{aligned} \quad (126)$$

where (a) follows as $h^{(k)} \sim \mathcal{E}(1)$ is independent across different time slots.

$$\begin{aligned}
\mathcal{M}_{P_s}^{\text{Poi}}(b, M) &\stackrel{(a)}{=} \mathbb{E}_{\Phi} \left[\prod_{j \in \mathbb{N}} \prod_{m=1}^M \left(\frac{1}{1 + \theta d_m^\alpha \|x_{j,m} - z_m\|^{-\alpha}} \right)^b \right], \\
&\stackrel{(b)}{=} \begin{cases} \exp \left(-\lambda \int_{\mathbb{R}^2} \left(1 - \prod_{m=1}^M \left(\frac{1}{1 + \theta d_m^\alpha \|x - z_m\|^{-\alpha}} \right)^b \right) dx \right), & \text{QSI} \\ \exp \left(-\lambda \prod_{m=1}^M \int_{\mathbb{R}^2} \left(1 - \left(\frac{1}{1 + \theta d_m^\alpha \|x - z_m\|^{-\alpha}} \right)^b \right) dx \right), & \text{FVI} \end{cases} \quad (125)
\end{aligned}$$

In the scenarios with QSI, the JSP can be expressed as

$$\begin{aligned}
\mathcal{J}_K^{\text{QSI}} &\stackrel{(b)}{=} \mathbb{E} \left[\prod_{j \in \mathbb{N}} \left(\frac{1}{1 + \theta r_t^\alpha \|x_j^{(k)}\|^{-\alpha}} \right)^K \right] \\
&\stackrel{(c)}{=} \exp \left(-2\pi\lambda \int_0^\infty \left(1 - \left(\frac{1}{1 + \theta r_t^\alpha x^{-\alpha}} \right)^K \right) x dx \right), \quad (127)
\end{aligned}$$

where (b) follows as K transmissions are subject to the same point process of interferers, and (c) applies the PGFL of a PPP. As (127) is equivalent to the moments of the CSP_Φ in Poisson ad hoc networks in (110) with b replaced by K , we have the final results of $\mathcal{J}_K^{\text{QSI}}$ in (92).

Moreover, in the scenarios with FVI, as different transmissions are affected by independent point processes, the JSP can be derived as

$$\begin{aligned}
\mathcal{J}_K^{\text{FVI}} &= \prod_{k=1}^K \mathbb{E}_{h_j^{(k)}} \left[\exp \left(-\theta r_t^\alpha \sum_{j \in \mathbb{N}} h_j^{(k)} \|x_j^{(k)}\|^{-\alpha} \right) \right] \\
&\stackrel{(d)}{=} \prod_{k=1}^K \mathbb{E} \left[\prod_{j \in \mathbb{N}} \frac{1}{1 + \theta r_t^\alpha \|x_j^{(k)}\|^{-\alpha}} \right] \\
&\stackrel{(e)}{=} \exp \left(-2\pi\lambda \int_0^\infty \left(1 - \frac{1}{1 + \theta r_t^\alpha x^{-\alpha}} \right) x dx \right)^K \\
&\stackrel{(f)}{=} \exp \left(-c\lambda r_t^2 \theta^\delta K \right),
\end{aligned}$$

where (d) holds as $h_j^{(k)} \sim \mathcal{E}(1)$, (e) follows as the point process of interferers across different transmissions are i.i.d. and applies the PGFL of a PPP, and (f) follows the same steps of the derivations of (121). \square

I. Proof of Corollary 4 (Success Probability of Type-II HARQ)

Proof. When $K = 2$, the success probability with Type-II HARQ-CC given in (98) can be rewritten as

$$\begin{aligned}
\mathcal{P}_{\text{II}} &= \mathbb{P}[\eta^{(1)} > \theta] + \mathbb{P}[\eta^{(1)} + \eta^{(2)} > \theta, \eta^{(1)} < \theta] \\
&= \mathbb{P}[\eta^{(1)} > \theta] \\
&\quad + \mathbb{E} \left[\mathbb{P}[\eta^{(2)} > \theta - \eta^{(1)} \mid \eta^{(1)}] \mathbb{1}_{\{\eta^{(1)} < \theta\}} \right] \\
&= \mathcal{J}_1 + \int_0^T \mathbb{P}[\eta^{(2)} > \theta - u] f_{\eta^{(1)}}(u) du, \quad (128)
\end{aligned}$$

where \mathcal{J}_1 is given in Theorem 10 and $f_{\eta^{(1)}}(t)$ represents the PDF of $\eta^{(1)}$.

Then, we derive the conditional CDF of $\eta^{(1)}$ given $\Phi^{(1)}$ as follows

$$\begin{aligned}
\mathbb{P}[\eta^{(1)} \leq t] &= \mathbb{E} \left[1 - \exp \left(-u r_t^\alpha \sum_{j \in \mathbb{N}} h_j^{(1)} \|x_j\|^{-\alpha} \right) \right] \\
&= 1 - \prod_{j \in \mathbb{N}} \frac{1}{1 + u r_t^\alpha r_j^{-\alpha}}. \quad (129)
\end{aligned}$$

Subsequently, the PDF of $\eta^{(1)}$ can be derived by taking the derivative of (129) with respect to u as follows.

$$\begin{aligned}
f_{\eta^{(1)}}(u) &= \frac{\partial \left(1 - \prod_{j \in \mathbb{N}} \frac{1}{1 + u r_t^\alpha r_j^{-\alpha}} \right)}{\partial u} \\
&= \sum_{j \in \mathbb{N}} \frac{r_t^\alpha r_j^{-\alpha}}{(1 + u r_t^\alpha r_j^{-\alpha})^2} \prod_{i \in \mathbb{N}, i \neq j} \frac{1}{1 + u r_t^\alpha r_i^{-\alpha}}. \quad (130)
\end{aligned}$$

By plugging (130) into the second term of (128), we have (131), where (a) applies the Campbell-Mecke formula given in (27).

Subsequently, by computing the expectations in (131) based on the PGFL of the PPP, we have the final results in **Corollary 4**. \square

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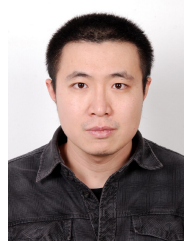
$$\begin{aligned}
& \int_0^T \mathbb{P}[\eta^{(2)} > \theta - u] f_{\eta^{(1)}}(u) du \\
&= \begin{cases} \int_0^T \mathbb{E} \left[\sum_{j \in \mathbb{N}} \frac{r_t^\alpha r_j^{-\alpha}}{(1 + ur_t^\alpha r_j^{-\alpha})^2} \frac{1}{1 + (\theta - u)r_t^\alpha r_j^{-\alpha}} \prod_{i \in \mathbb{N}}^{i \neq j} \frac{1}{1 + ur_t^\alpha r_i^{-\alpha}} \frac{1}{1 + (\theta - u)r_t^\alpha r_i^{-\alpha}} \right] du, & \text{QSI} \\ \int_0^T \mathbb{E} \left[\prod_{j' \in \mathbb{N}} \frac{1}{1 + (\theta - u)r_t^\alpha r_{j'}^{-\alpha}} \sum_{j \in \mathbb{N}} \frac{r_t^\alpha r_j^{-\alpha}}{(1 + ur_t^\alpha r_j^{-\alpha})^2} \prod_{i \in \mathbb{N}}^{i \neq j} \frac{1}{1 + ur_t^\alpha r_i^{-\alpha}} \right] du, & \text{FVI} \end{cases} \\
&\stackrel{(a)}{=} \begin{cases} 2\pi\lambda \int_0^T \int_0^\infty \frac{r_t^\alpha r^{-\alpha}}{(1 + ur_t^\alpha r^{-\alpha})^2} \frac{1}{1 + (\theta - u)r_t^\alpha r^{-\alpha}} r dr \mathbb{E} \left[\prod_{i \in \mathbb{N}}^{i \neq j} \frac{1}{1 + ur_t^\alpha r_i^{-\alpha}} \frac{1}{1 + (\theta - u)r_t^\alpha r_i^{-\alpha}} \right] du, & \text{QSI} \\ 2\pi\lambda \int_0^T \mathbb{E} \left[\prod_{j' \in \mathbb{N}} \frac{1}{1 + (\theta - u)r_t^\alpha r_{j'}^{-\alpha}} \right] \int_0^\infty \frac{r_t^\alpha r^{-\alpha}}{(1 + ur_t^\alpha r^{-\alpha})^2} r dr \mathbb{E} \left[\prod_{i \in \mathbb{N}} \frac{1}{1 + ur_t^\alpha r_i^{-\alpha}} \right] du, & \text{FVI}, \end{cases} \quad (131)
\end{aligned}$$

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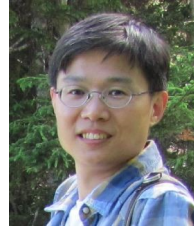


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