

Carrier-Envelope-Phase Modulated Currents in Scanning Tunneling Microscopy

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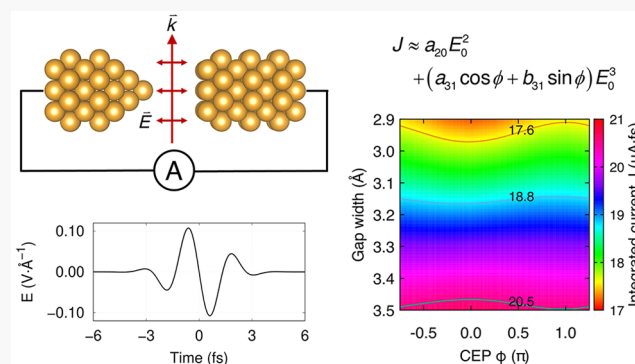
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ABSTRACT: Carrier-envelope-phase (CEP) stable optical pulses combined with state-of-the-art scanning tunneling microscopy (STM) can track and control ultrafast electronic tunneling currents. On the basis of nonequilibrium Green's function formalism, we present a time and frequency domain theoretical study of CEP-stable pulse-induced tunneling currents between an STM tip and a metal substrate. It is revealed that the experimentally observed phase shift between the maximum tunneling current and maximum electric field is caused by the third-order response to the electric field. The shift is also found to be sensitive to the duration of pulses. The tunneling process can thus be precisely manipulated by varying the phase and duration of these pulses.

KEYWORDS: scanning tunneling microscopy, carrier-envelope phase, phase shift, perturbative nonequilibrium Green's functions, tunneling control



Controlling the ultrafast electron dynamics in nanodevices is one of the central goals of nanoscience. Terahertz (THz) pulses coupled to STM have been recently used to monitor tunneling dynamics with subpicosecond temporal and nanometer spatial resolutions.^{1–6} However, THz pulses are too long to track the electron dynamics in its natural femtosecond time scale.^{7,8} Ultrashort optical pulses provide a promising route to modulate and track the tunneling process.⁹ The electron tunneling was found sensitive to the CEP of the optical pulse.^{10–13} Experimentally, the CEP can be precisely modulated and has been used to coherently control the light-driven tunneling current between STM tip and substrate^{4,9} and the photoemission of electrons from nanotips.^{14,15} Understanding the mechanism of CEP-dependent tunneling current in a junction is thus crucial for advancing ultrafast STM techniques.

The optical pulses created by mode-locked lasers consist of a train of Gaussian enveloped sinusoidal pulses.⁹ The vector potential of a laser pulse can be expressed as

$$A(t) = \frac{E_0}{\omega_0} e^{-t^2/2\sigma^2} \cos(\omega_0 t + \phi) \quad (1)$$

Here E_0 is the electric field amplitude, ω_0 is the central frequency, σ is the variance, and ϕ is the CEP, respectively. The full width at half-maximum of the pulse is $\text{fwhm} = 2\sqrt{2\ln 2}\sigma$. The electric field $E(t) = -\partial_t A(t)$ is an AC field. The CEP is related to the time difference between the peaks of the envelope and the sinusoidal wave as $\Delta t = \phi/\omega_0$.

This is controlled by attosecond technology.^{16–24} The vector potential and CEP are illustrated in Figure 1a. The electric field variation with the CEP is shown in Figure 1b.

Detecting the femtosecond transient tunneling current in real time is not possible with current technology. The quantity being measured is the average current during a pulse period, i.e., the integrated current.^{5,9} The electric field changes with the CEP, and the measured tunneling current will vary as well. However, the measured tunneling current is not synchronous with the electric field but has a phase shift.^{1,4,9} To study the role of the phase shift, we first derive perturbative expressions in the optical field for the integrated current based on nonequilibrium Green's function (NEGF) formalism^{25,26} and then verify them by time-domain nonperturbative^{27–29} and perturbative³⁰ simulations.

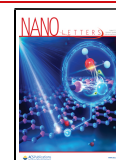
Consider an STM junction coupled to two leads, i.e., tip and substrate with noninteracting electrons. The time-dependent current passing through each lead can be expressed based on Keldysh NEGF formalism as³¹

$$I_\alpha(t) = e \int d\tau \text{tr}[\mathbf{G}^<(t, \tau) \Sigma_\alpha^a(\tau, t) + \mathbf{G}^r(t, \tau) \Sigma_\alpha^<(\tau, t) + \text{hc}] \quad (2)$$

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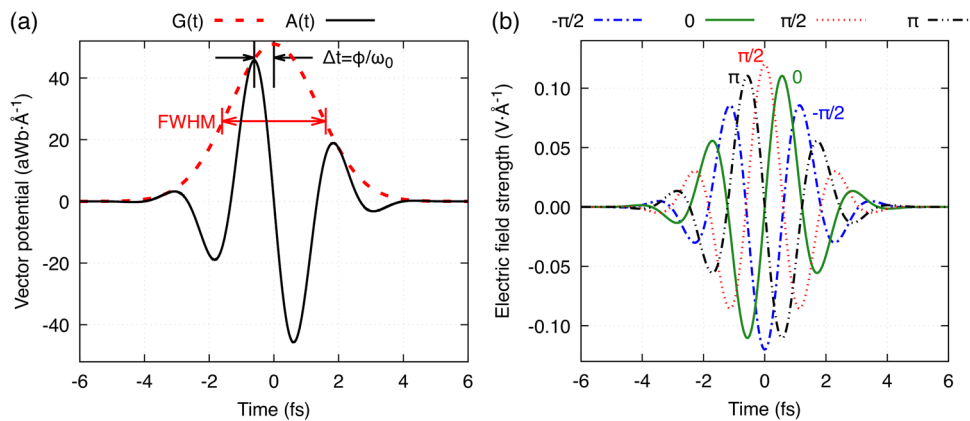


Figure 1. (a) One of the vector potentials used in our simulations. Red dotted line: the Gaussian envelope $G(t)$ with an fwhm of 3.2 fs and an amplitude of 51.0 aWb/Å; black solid line: the vector potential $A(t)$ with a central frequency equivalent to 1.55 eV (800 nm) and a CEP of 0.5 π . The time difference between the peaks of the envelope and the sinusoidal wave is $\Delta t = \phi/\omega_0$. (b) The electric field variation with the CEP.

Here $G^{</r}$ is the lesser/retarded Green's function, and $\Sigma_{\alpha}^{a/<}$ is the advanced/lesser self-energy of lead α , respectively. Because the tunneling gap investigated is relatively narrow, the optical electric field can be approximated as quasi-static.⁶ Upon the application of an electric field, the induced bias at each lead can thus be taken as $V_{\alpha}(t) = \pm E(t)L/2$. Here $E(t)$ is the near-field strength determined by eq 1, and L is the device region length along the transport direction. Since the applied electric field wavelength is much longer than the device size, the amplitude E_0 can be considered to be constant within the system of interest. The potential will be almost flat inside the tip and substrate, owing to screening, and the potential drops between the gap, inducing the local field enhancement. The coupling Hamiltonian to the electric field is $\Delta(t) = -\mu E(t)$ within dipole approximation, where μ is the electric dipole moment matrix of the device region. The variations of Green's function and self-energy with the electric field are induced by both the bias and Hamiltonian changes. In the wide-band limit approximation,³² the advanced self-energy is invariant to the external voltage, and the n -th order response of lesser self-energy can be expressed as³³

$$\Sigma_n^{<}(t, \tau) = \frac{1}{n!} \left[\frac{i}{\hbar} \int_t^{\tau} dt' eV(t') \right]^n \Sigma_0^{<}(t, \tau) \quad (3)$$

The n -th order variations of the retarded and lesser Green's functions are then³⁴

$$\mathbf{G}_n^r(t, t_{n+1}) = \int dt_1 \cdots dt_n \mathbf{G}_0^r(t, t_1) \prod_{j=1}^n \Delta(t_j) \mathbf{G}_0^r(t_j, t_{j+1}) \quad (4)$$

$$\mathbf{G}_n^{<}(t, \tau) = \sum_{i+j+k=n} \int dt_1 dt_2 \mathbf{G}_i^r(t, t_1) \Sigma_j^{<}(t_1, t_2) \mathbf{G}_k^a(t_2, \tau) \quad (5)$$

$\Sigma^{<}$ is the sum of self-energies of all leads. Thus, the n -th order response current is

$$I_{\alpha}^{(n)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}_n^{<}(t, \tau) \Sigma_{\alpha}^a(\tau, t) + \sum_{i+j=n} \mathbf{G}_i^r(t, \tau) \Sigma_{\alpha}^{<}(\tau, t) + \text{hc} \right] \quad (6)$$

The integrated current is

$$J_{\alpha}^{(n)} = \int dt I_{\alpha}^{(n)}(t) \quad (7)$$

It is more convenient to work on frequency domain whereby

$$J_{\alpha}^{(n)} = I_{\alpha}^{(n)}(\omega = 0) \quad (8)$$

It is straightforward to get the integrated first-order response current as³¹

$$J_{\alpha}^{(1)} = \frac{eE(\omega = 0)}{2\pi\hbar} \sum_{\beta} \int d\epsilon \text{tr} [\mathcal{G}_{\alpha\beta}^{(1)}(\epsilon)] \quad (9)$$

$\mathcal{G}_{\alpha\beta}^{(1)}$ is given in Supporting Information. The integrated electric field is given by

$$E(\omega = 0) = \frac{i\omega}{\omega_0} (\sqrt{2\pi} e^{-\sigma^2 \omega_0^2/2} \sigma \cos \phi) E_0 \Big|_{\omega=0} = 0 \quad (10)$$

We, therefore, obtain the zero first-order integrated current

$$J_{\alpha}^{(1)} = 0 \quad (11)$$

Thus, phase control is impossible in the linear response.^{35,36} Similarly, for the second-order integrated current, we have

$$J_{\alpha}^{(2)} = \frac{e}{4\pi^2 \hbar^2} \sum_{\beta} \int d\epsilon_1 E(-\epsilon_1/\hbar) E(\epsilon_1/\hbar) \int d\epsilon \text{tr} [\mathcal{G}_{\alpha\beta}^{(2)}(\epsilon, \epsilon_1)] \quad (12)$$

$\mathcal{G}_{\alpha\beta}^{(2)}(\epsilon, \epsilon_1)$ is given in Supporting Information. For the electric field part, we obtain

$$E(-\epsilon_1/\hbar) E(\epsilon_1/\hbar) = \left(\frac{\epsilon_1 \sigma E_0}{\hbar \omega_0} \right)^2 \pi e^{-\sigma^2 [\omega_0^2 + (\epsilon_1/\hbar)^2]} [\cos(2\phi) + \cosh(2\sigma^2 \omega_0 \epsilon_1/\hbar)] \quad (13)$$

Denoting the coefficients before E_0^2 as a_{20} and $\cos(2\phi) E_0^2$ as a_{22} , respectively, we have

$$J_{\alpha}^{(2)} = [a_{20} + a_{22} \cos(2\phi)] E_0^2 \quad (14)$$

Experimental $\hbar\omega_0$ is around several eV, and $\sigma/\hbar$ is around several eV⁻¹, thus $\cosh(2\sigma^2 \omega_0 \epsilon_1/\hbar) \gg 1$ for not too small ϵ_1 . Consequently, we have $a_{20} \gg a_{22}$; i.e., $J_{\alpha}^{(2)} \approx a_{20} E_0^2$. The second-order response thus acts as a rectifier for all ϕ .

The third-order integrated current is

$$J_{\alpha}^{(3)} = \frac{e}{8\pi^3 \hbar^3} \sum_{\beta} \iint d\epsilon_2 d\epsilon_1 E\left(\frac{\epsilon_1}{\hbar}\right) E\left(\frac{\epsilon_2 - \epsilon_1}{\hbar}\right) E\left(-\frac{\epsilon_2}{\hbar}\right) \int d\epsilon \text{tr} [\mathcal{G}_{\alpha\beta}^{(3)}(\epsilon, \epsilon_1, \epsilon_2)] \quad (15)$$

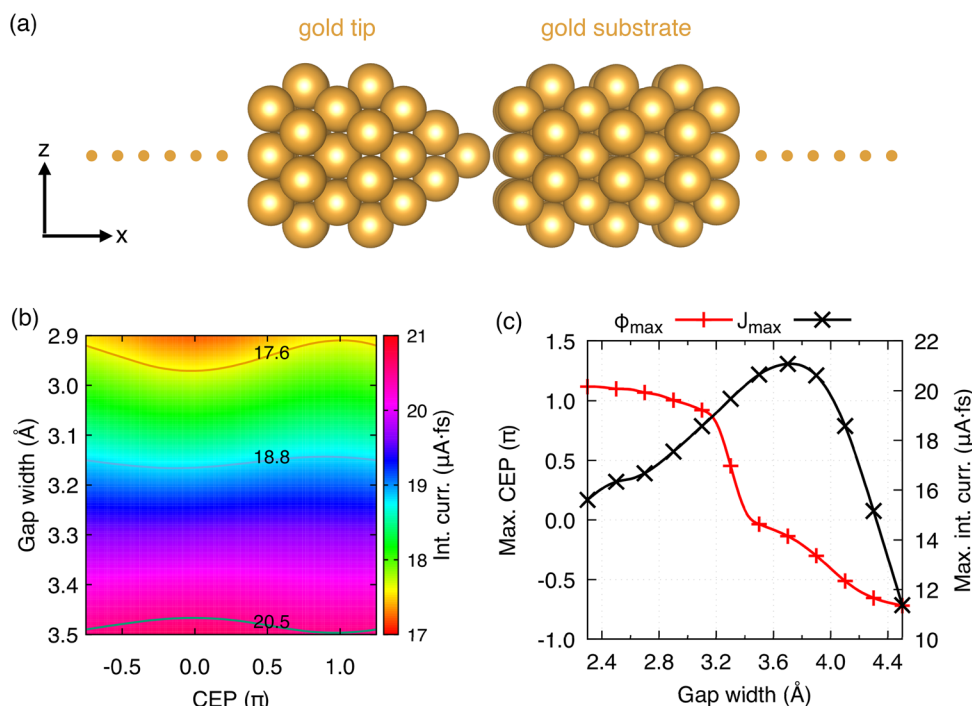


Figure 2. CEP modulated tunneling in a gold-tip-gold-substrate system. (a) The structure of the device region used in TDDFTB-OS simulations. Both the tip and substrate are semi-infinite in the $-x/+x$ -directions, respectively. (b) Modulated integrated current as a function of CEP and gap width. The color bar displays the integrated current in $\mu\text{A}\cdot\text{fs}$. Three contours are plotted to guide the eyes of readers to the maxima of integrated current. The applied electric field is transmitted along the z -direction and polarized along the x -direction. $E_0 = 0.05 \text{ V/\AA}$, $\hbar\omega_0 = 1.55 \text{ eV}$ (800 nm), and $\text{fwhm} = 3.4 \text{ fs}$. (c) The maximum integrated current (J_{max}) and the corresponding CEP (ϕ_{max}) as a function of gap width. ϕ_{max} is plotted as the red “+” line with scales at the left vertical axis. J_{max} is plotted as the black “x” line with scales at the right vertical axis.

The electric field part is

$$E\left(\frac{\epsilon_1}{\hbar}\right)E\left(\frac{\epsilon_2 - \epsilon_1}{\hbar}\right)E\left(-\frac{\epsilon_2}{\hbar}\right) \\ = \left(\frac{\sigma E_0}{\hbar\omega_0}\right)^3 \pi \sqrt{\frac{\pi}{2}} \epsilon_1(\epsilon_2 - \epsilon_1)\epsilon_2 e^{-\sigma^2[(\epsilon_1^2 - \epsilon_1\epsilon_2 + \epsilon_2^2)/\hbar^2 + 3\omega_0^2/2]} \\ [iA\cos\phi - B\sin\phi + \cos(3\phi)] \quad (16)$$

where

$$A = \cosh(2\sigma^2\omega_0\epsilon_1/\hbar) + \cosh[2\sigma^2\omega_0(\epsilon_1 - \epsilon_2)/\hbar] + \cosh(2\sigma^2\omega_0\epsilon_2/\hbar) \\ B = \sinh(2\sigma^2\omega_0\epsilon_1/\hbar) - \sinh[2\sigma^2\omega_0(\epsilon_1 - \epsilon_2)/\hbar] - \sinh(2\sigma^2\omega_0\epsilon_2/\hbar) \quad (17)$$

$\mathcal{G}_{a\beta}^{(3)}(\epsilon, \epsilon_1, \epsilon_2)$ is given in Supporting Information. We thus have

$$J_{\alpha}^{(3)} = [a_{31}\cos\phi + b_{31}\sin\phi + a_{33}\cos(3\phi)]E_0^3 \quad (18)$$

With the same argument, we have $a_{31}, b_{31} \gg a_{33}$.

The integrated current to third order in the optical field is thus given by

$$J_{\alpha} \approx a_{20}E_0^2 + a_{31}E_0^3\cos\phi + b_{31}E_0^3\sin\phi \\ = a_{20}E_0^2 + R\cos(\phi - \phi_{\text{max}})E_0^3 \quad (19)$$

where

$$R = \sqrt{a_{31}^2 + b_{31}^2} \\ \tan\phi_{\text{max}} = b_{31}/a_{31} \quad (20)$$

Here ϕ_{max} is the CEP that maximizes the integrated current. It is apparent that the ϕ_{max} is mainly caused by the third-order response. Note that a_{20} , a_{31} , and b_{31} are not solely determined by the system but are also related to the central frequency and fwhm of the external electric field.

To examine further, we carry out a quantum mechanical simulation of a gold-tip-gold-substrate system using the open system time-dependent density-functional tight-binding (TDDFTB-OS) method.^{27–29} Both the tip and substrate are semi-infinite along the transmission direction x . The device region containing 128 gold atoms is depicted in Figure 2a, with the gap width varying from 2.3 to 4.5 Å. The Gaussian pulse with an amplitude of 0.05 V/Å is transmitted along z and polarized along x . The central frequency is 1.55 eV (800 nm) with an fwhm of 3.4 fs. The wide-band limit approximation is applied, and the induced Hamiltonian $\Delta(t)$ is obtained by solving the Poisson equation. DFTB parameters provided by A. Fihey et al. are used.³⁷ The system is propagated up to 160 fs with the applied electric field. The integrated current (J) containing the response to all orders with different gap widths and CEP is plotted in Figure 2b. ϕ_{max} and the corresponding maximum integrated current (J_{max}) at each gap width are depicted in Figure 2c. J_{max} has a peak near a gap width of 3.7 Å. When the gap width is small, the tip and substrate fuse into a single metal and thus screen out the electric field.³⁸ From Figure 2c, it is clear that ϕ_{max} changes with the gap width monotonically, which is consistent with the experimental results.⁹

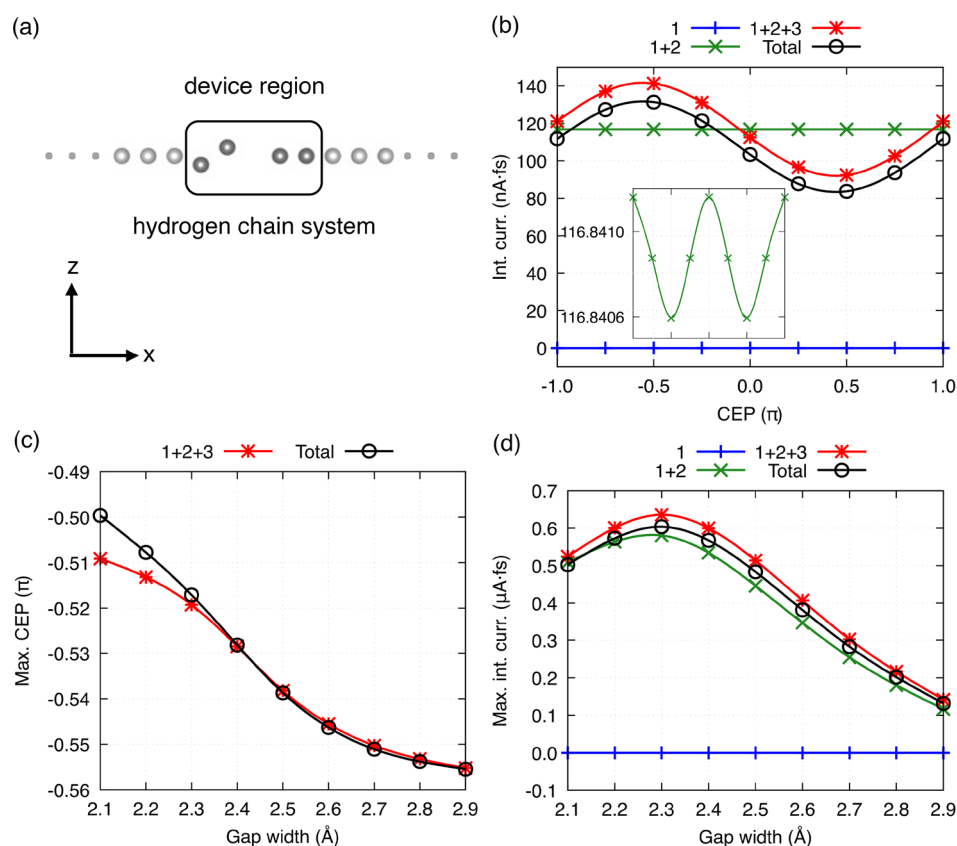


Figure 3. Perturbative time-domain simulation of a hydrogen chain system. (a) The chain structure. The device region included in the TDDFTB-OS simulation is framed by the black rounded rectangle. Both leads are semi-ininitely extended to the $-x/+x$ -directions. (b) The sum of different orders integrated current as a function of CEP at a gap width of 2.9 Å. (c) The maximum CEP calculated from the sum of different orders integrated current as a function of gap width. (d) The maximum integrated current calculated from the sum of different orders responses as a function of gap width. The first-order integrated current $J^{(1)}$, the sum of first- and second-order integrated current $J^{(1)}+J^{(2)}$, the sum of first-, second-, third-order integrated current $J^{(1)}+J^{(2)}+J^{(3)}$, and full-order integrated current J are plotted as blue “+”, green “x”, red “*”, and black “O” lines, respectively.

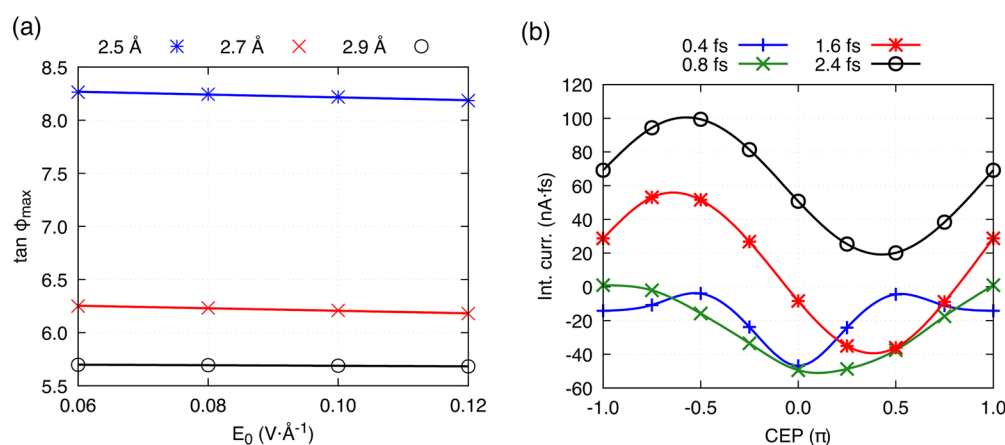


Figure 4. (a) The relation between the tangent of maximum CEP ($\tan \phi_{\max}$) and the amplitude of electric field (E_0). $\phi_{\max}$ calculated from the hydrogen chain system depicted in Figure 3a at gap widths of 2.5, 2.7, and 2.9 Å are plotted as the blue “*”, red “x”, and black “O” marks, respectively. The solid lines are the linearly fitted curves. (b) Total integrated current with pulses of different fwhm for the hydrogen chain system at gap width 2.9 Å. The minimum of integrated current shifts to 0π as fwhm decreases.

To verify the frequency-domain analysis, which implies that the shift of $\phi_{\max}$ is at least a third-order effect, we have performed a time-domain perturbative TDDFTB-OS simulation³⁰ for the asymmetric hydrogen chain system shown in Figure 3a. Simulating the order-by-order dynamics of the gold–gold system is too expensive. The hydrogen chain system

is chosen to demonstrate our point. The device region contains 4 hydrogen atoms with the gap width varying between 2.1 and 2.9 Å. The distance between hydrogen atoms on each side is 1.5 Å along x . The Gaussian pulse with an amplitude of 0.12 V/Å is transmitted along z and polarized along the transmission direction (x). The central frequency is 1.55 eV

(800 nm) with an fwhm of 3.2 fs. DFTB parameters were provided by M. Elstner et al.³⁹ A 29 fs simulation was required to get the saturated integrated current. In Figure 3b, we analyze the maximum integrated current calculated by the sum of different orders responses as a function of the CEP at a gap width of 2.9 Å. $J^{(1)}$ is zero within numerical errors, and $J^{(2)}$ increases the integrated current almost homogeneously for all CEP. The inset shows the negligible variation of $J^{(2)}$ with the CEP. $J^{(3)}$ shifts $\phi_{\max}$ to -0.555π , and the higher-order responses do not shift $\phi_{\max}$ further. $\phi_{\max}$ and $J_{\max}$ calculated from the sum of different orders responses with various gap widths are plotted separately in Figure 3c,d. Again, only for the sum of first-, second-, and third-order integrated current $\phi_{\max}$ depends on the gap width and is virtually identical to $\phi_{\max}$ obtained from the total nonperturbative integrated current. From Figure 3d, the difference between $J_{\max}$ obtained from the sum of first-, second-, and third-order responses and $J_{\max}$ from the total responses is negligible. We thus verified that the shift of $\phi_{\max}$ from 0 is at least a third-order effect in the optical field. This is consistent with M. Garg and K. Kern's experimental work,⁹ where the shift of $\phi_{\max}$ was observed at a field-driven regime, in which third-order photon processes contribute.

To validate eqs 19 and 20, $\phi_{\max}$ is calculated for different E_0 values, where third-order response dominates using the same time-domain method for the same hydrogen chain system. The relation between $\tan \phi_{\max}$ and E_0 is shown in Figure 4a for gap widths of 2.5 Å (blue), 2.7 Å (red), and 2.9 Å (black). The marked points in Figure 4a are the simulated results, while the solid lines are the linearly fitted functions. The lines are almost horizontal; thus, eq 20 is confirmed. At small gap widths, the second-order response $\cos(2\phi)$ is stronger, causing the slight inclination of fitted lines. This relation may further be checked with existing experimental techniques in the future.

For a fixed gap width, $\phi_{\max}$ is also determined by the fwhm of the pulse. We consider two extreme cases where the fwhm tends to zero and infinity. When expanding eqs 13 and 16 at $\sigma = 0$, we have

$$\begin{aligned} E\left(-\frac{\varepsilon_1}{\hbar}\right)E\left(\frac{\varepsilon_1}{\hbar}\right) &= O(\sigma^2)[1 + \cos(2\phi)] \\ E\left(\frac{\varepsilon_1}{\hbar}\right)E\left(\frac{\varepsilon_2 - \varepsilon_1}{\hbar}\right)E\left(-\frac{\varepsilon_2}{\hbar}\right) &= O(\sigma^3)[3\cos\phi + \cos(3\phi)] \\ &+ [O(\sigma^5)\cos\phi + O(\sigma^9)\sin\phi + O(\sigma^5)\cos(3\phi)] \end{aligned} \quad (21)$$

It is clear that $\cos\phi$ and $\cos(2\phi)$ dominate in the integrated current, so $\phi_{\max}$ will be less shifted from 0 for a small fwhm. We simulated the hydrogen chain system with a decreasing fwhm of pulses at a gap width of 2.9 Å using the time-domain method and plotted the results in Figure 4b. It is clear the $\cos\phi$ component becomes more dominant as fwhm decreases, and for a small enough fwhm, the $\cos(2\phi)$ term emerges. For a large fwhm scenario, $E(\varepsilon_1)E(\varepsilon_2 - \varepsilon_1)E(-\varepsilon_2)$ vanishes, so the integrated current will be independent of the CEP.

We note that the shifted $\phi_{\max}$ is not due to the system's asymmetry. We can still observe the shift $\phi_{\max} = 0.6\pi$ for a symmetric hydrogen chain with a gap width of 2.9 Å at an electric field amplitude of 0.12 V/Å and an fwhm of 3.2 fs. Instead, asymmetry leads to an even order integrated current and results in the rectifier effect.⁴⁰ Especially for a large fwhm, the odd order signal vanishes, and the total integrated current will be virtually independent of the CEP.

If the tip–substrate system is extremely close to the light source, the integrated field might survive.⁴¹ Then, the first-order integrated current will interfere with the third-order integrated current and further shift the $\phi_{\max}$.

In this work, the wide-band limit approximation is adopted. The inclusion of the electronic structures of the leads will make $\mathcal{G}_{\alpha\beta}^{(i)}$ considerably complicated.³¹ For the current in the negative differential resistance regime, the coefficients before cosine and sine terms will change, but our conclusion will still hold.

In summary, time-domain TDDFTB-OS simulations demonstrate that the tunneling current can be modulated by tuning the CEP of few-cycle optical pulses. The phase shift between the current and the field in a gold-tip-gold-substrate system is found to vary monotonically with the gap width. Analytical expressions of the different order integrated current are obtained by perturbative expansion of the effective tunneling current under the wide-band limit approximation. We find that the first-order integrated current is zero, and the third-order integrated current is phase shifted. The even order responses can generate a rectification of STM for asymmetric systems. We have also performed time-domain perturbative simulations and confirmed that the phase shift is mainly caused by the third and possibly higher-order responses. The phase shift further depends on the central frequency and fwhm of the pulses. These findings demonstrate how to modulate the tunneling process by tuning the pulse profile. This perturbative technique can also be applied to the pump–probe setup and will be reported in the future.

■ ASSOCIATED CONTENT

Supporting Information

The Supporting Information is available free of charge at <https://pubs.acs.org/doi/10.1021/acs.nanolett.1c01900>.

Detailed derivation of first-, second-, and third-order integrated currents in the nonequilibrium Green's function formalism under wide-band limit approximation (PDF)

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Notes

The authors declare no competing financial interest.

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Supporting information

Carrier-Envelope-Phase Modulated Currents in Scanning Tunneling Microscopy

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First Order Integrated Current

For simplicity, we use the following abbreviated integration notation,

$$\int dt_1 dt_2 \cdots dt_n \mathbf{A}_0(t, t_1) \mathbf{A}_1(t_1, t_2) \cdots \mathbf{A}_n(t_n, t) = \int d\tau \mathbf{A}_0 \mathbf{A}_1 \cdots \mathbf{A}_n. \quad (\text{S1})$$

$\mathbf{A}_i$ are general matrices, and they can also be single-variable, $\mathbf{A}_i(t_i)$. The fold of integration can be deduced from the number of $\mathbf{A}_i$.

We split the response to electric field into two parts, *i.e.*, (1) the sole response to induced bias; (2) the remaining part. The first part is

$$I_\alpha^{(1,1)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Sigma_\alpha^a + \mathbf{G}^r \Sigma_{\alpha 1}^< + \text{h.c.} \right]. \quad (\text{S2})$$

Perform Fourier transform from t to ω ,

$$\begin{aligned} \mathcal{F}_t \left[\int d\tau \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Sigma_\alpha^a \right] &= \sum_\beta \frac{eV_\beta(\omega)}{2\pi\hbar^2\omega} \int d\varepsilon \mathbf{G}_+^r \Sigma_\beta^< \mathbf{G}^a \Sigma_\alpha^a, \\ \mathcal{F}_t \left[\int d\tau \mathbf{G}^r \Sigma_{\alpha 1}^< \right] &= \frac{eV_\alpha(\omega)}{2\pi\hbar^2\omega} \int d\varepsilon \mathbf{G}_+^r \Sigma_\alpha^<. \end{aligned} \quad (\text{S3})$$

Notations $\mathbf{A} = \mathbf{A}(\varepsilon)$, $\mathbf{A}_+ = \mathbf{A}(\varepsilon + \hbar\omega)$, $\mathbf{A}_- = \mathbf{A} - \mathbf{A}_+$ are used in eq S3. Choose $V_\alpha(t) = E(t)L/2$ and $V_\beta(t) = -E(t)L/2$, we have

$$\begin{aligned} I_\alpha^{(1,1)}(\omega) &= \frac{eE(\omega)}{2\pi\hbar} \sum_\beta \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(1,1)}(\varepsilon, \omega) \right], \\ \mathcal{G}_{\alpha\beta}^{(1,1)}(\varepsilon, \omega) &= \frac{eL}{2\hbar\omega} \left[(2\delta_{\alpha\beta} - 1) \left(\mathbf{G}_+^r \Sigma_\beta^< \mathbf{G}^a \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}_+^r \Sigma_\beta^< \mathbf{G}^a \right) + \delta_{\alpha\beta} \left(\mathbf{G}_+^r \Sigma_\beta^< - \Sigma_\beta^< \mathbf{G}^a \right) \right]. \end{aligned} \quad (\text{S4})$$

The second part is

$$I_{\alpha}^{(1,2)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Delta \mathbf{G}^< \Sigma_{\alpha}^a + \mathbf{G}^< \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha}^< + \text{h.c.} \right]. \quad (\text{S5})$$

Perform Fourier transform from t to ω ,

$$I_{\alpha}^{(1,2)}(\omega) = \frac{e}{2\pi\hbar} \int d\varepsilon \text{tr} \left[\mathbf{G}_+^r \Delta(\omega) \mathbf{G}^< \Sigma_{\alpha}^a + \mathbf{G}_+^< \Delta(\omega) \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}_+^r \Delta(\omega) \mathbf{G}^r \Sigma_{\alpha}^< + \text{h.c.} \right]. \quad (\text{S6})$$

Set $\omega = 0$ and denote $\Lambda_{\alpha} = \text{Im}[\Sigma_{\alpha}^a]$, we have

$$\begin{aligned} J_{\alpha}^{(1,2)} &= \frac{ie}{\pi\hbar} \int d\varepsilon \text{tr} \left[\Delta(\omega=0) \left(\mathbf{G}^< \Lambda_{\alpha} \mathbf{G}^r + \mathbf{G}^a \Lambda_{\alpha} \mathbf{G}^< + F \mathbf{G}^r \Lambda_{\alpha} \mathbf{G}^r - F \mathbf{G}^a \Lambda_{\alpha} \mathbf{G}^a \right) \right] \\ &= 0. \end{aligned} \quad (\text{S7})$$

$F(\varepsilon)$ is the Fermi-Dirac distribution, and $\mathbf{G}^< = -F[\mathbf{G}^r - \mathbf{G}^a]$ is used in the last line of eq S7.

The first order integrated current is

$$\begin{aligned} J_{\alpha}^{(1)} &= \frac{eE(\omega=0)}{2\pi\hbar} \sum_{\beta} \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(1)}(\varepsilon) \right], \\ \mathcal{G}_{\alpha\beta}^{(1)}(\varepsilon) &= \lim_{\omega \rightarrow 0} \mathcal{G}_{\alpha\beta}^{(1,1)}(\varepsilon, \omega). \end{aligned} \quad (\text{S8})$$

Second Order Integrated Current

Following the split scheme used in the first order integrated current, the first part of second order current is

$$I_{\alpha}^{(2,1)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Sigma_2^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Sigma_{\alpha 2}^< + \text{h.c.} \right]. \quad (\text{S9})$$

Perform Fourier transform from t to ω and set $\omega = 0$,

$$J_{\alpha}^{(2,1)} = \frac{e}{4\pi^2\hbar^2} \sum_{\beta} \int d\varepsilon_1 E(-\varepsilon_1 / \hbar) E(\varepsilon_1 / \hbar) \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(2,1)}(\varepsilon, \varepsilon_1) \right], \quad (\text{S10})$$

$$\mathcal{G}_{\alpha\beta}^{(2,1)}(\varepsilon, \varepsilon_1) = -\frac{e^2 L^2}{4\varepsilon_1^2} \left[\mathbf{G}^r \underline{\Sigma}_{\beta}^< \mathbf{G}^a \Sigma_{\alpha}^a - \Sigma_{\alpha}^r \mathbf{G}^r \underline{\Sigma}_{\beta}^< \mathbf{G}^a + \delta_{\alpha\beta} \left(\mathbf{G}^r \underline{\Sigma}_{\beta}^< - \underline{\Sigma}_{\beta}^< \mathbf{G}^a \right) \right].$$

Notation $\underline{\mathbf{A}} = \mathbf{A} - \mathbf{A}(\varepsilon + \varepsilon_1)$ is used. Plug the relation $\Sigma_{\alpha}^<(\varepsilon) = 2i\Lambda_{\alpha}F(\varepsilon)$ into eq S10, it can be shown that

$$J_{\alpha}^{(2,1)} = 0. \quad (\text{S11})$$

The second part is

$$I_{\alpha}^{(2,2)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^< \Sigma_{\alpha}^a + \mathbf{G}^< \Delta \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a \right. \\ \left. + \mathbf{G}^r \Delta \mathbf{G}^< \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha}^< + \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha 1}^< \right. \\ \left. + \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \text{h.c.} \right]. \quad (\text{S12})$$

Perform Fourier transform from t to ω and set $\omega = 0$,

$$J_{\alpha}^{(2,2)} = \frac{e}{4\pi^2\hbar^2} \sum_{\beta} \int d\varepsilon_1 E(-\varepsilon_1 / \hbar) E(\varepsilon_1 / \hbar) \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(2,2)}(\varepsilon, \varepsilon_1) \right],$$

$$\mathcal{G}_{\alpha\beta}^{(2,2)}(\varepsilon, \varepsilon_1) = \delta_{\alpha\beta} \left(\mathbf{G}^r \mu \mathbf{G}_{+1}^r \mu \mathbf{G}^< \Sigma_{\alpha}^a + \mathbf{G}^< \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_{\alpha}^a \right. \\ \left. + \mathbf{G}^r \mu \mathbf{G}_{+1}^< \mu \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \mu \mathbf{G}_{+1}^r \mu \mathbf{G}^r \Sigma_{\alpha}^< - \frac{eL}{2\varepsilon_1} \mathbf{G}^r \mu \mathbf{G}_{+1}^r \underline{\Sigma}_{\alpha}^< \right) \\ - \frac{eL}{2\varepsilon_1} (2\delta_{\alpha\beta} - 1) \left(\mathbf{G}^r \mu \mathbf{G}_{+1}^r \underline{\Sigma}_{\beta}^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \underline{\Sigma}_{\beta}^< \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_{\alpha}^a \right) \\ + \text{h.c.} \quad (\text{S13})$$

Notations $\mathbf{A}_{+1} = \mathbf{A}(\varepsilon + \varepsilon_1)$, $\underline{\mathbf{A}} = \mathbf{A} - \mathbf{A}_{+1}$ are used.

The second order integrated current is

$$J_{\alpha}^{(2)} = \frac{e}{4\pi^2\hbar^2} \sum_{\beta} \int d\varepsilon_1 E(-\varepsilon_1/\hbar) E(\varepsilon_1/\hbar) \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(2)}(\varepsilon, \varepsilon_1) \right], \quad (\text{S14})$$

$$\mathcal{G}_{\alpha\beta}^{(2)}(\varepsilon, \varepsilon_1) = \mathcal{G}_{\alpha\beta}^{(2,2)}(\varepsilon, \varepsilon_1).$$

Third Order Integrated Current

Following the split scheme used in the first order integrated current, the first part of third order current is

$$I_{\alpha}^{(3,1)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Sigma_3^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Sigma_{\alpha 3}^< + \text{h.c.} \right]. \quad (\text{S15})$$

Perform Fourier transform from t to ω and set $\omega = 0$,

$$J_{\alpha}^{(3,1)} = \frac{e}{8\pi^3\hbar^3} \sum_{\beta} \iint d\varepsilon_2 d\varepsilon_1 E\left(\frac{\varepsilon_1}{\hbar}\right) E\left(\frac{\varepsilon_2 - \varepsilon_1}{\hbar}\right) E\left(-\frac{\varepsilon_2}{\hbar}\right) \int d\varepsilon \text{tr} \left[\mathcal{G}_{\alpha\beta}^{(3,1)}(\varepsilon, \varepsilon_1, \varepsilon_2) \right],$$

$$\mathcal{G}_{\alpha\beta}^{(3,1)}(\varepsilon, \varepsilon_1, \varepsilon_2) = \frac{e^3 L^3}{16\varepsilon_1(\varepsilon_1 - \varepsilon_2)\varepsilon_2} \left[(2\delta_{\alpha\beta} - 1) \left(\mathbf{G}^r \Sigma_{\beta}^< \mathbf{G}^a \Sigma_{\alpha}^a - \Sigma_{\alpha}^r \mathbf{G}^r \Sigma_{\beta}^< \mathbf{G}^a \right) + \delta_{\alpha\beta} \left(\mathbf{G}^r \Sigma_{\beta}^< - \Sigma_{\beta}^< \mathbf{G}^a \right) \right]. \quad (\text{S16})$$

Notation $\tilde{\mathbf{A}} = \mathbf{A}(\varepsilon + \varepsilon_1) - \mathbf{A}(\varepsilon + \varepsilon_2)$ is used. The second part is

$$I_{\alpha}^{(3,2)}(t) = e \int d\tau \text{tr} \left[\mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^< \Sigma_{\alpha}^a + \mathbf{G}^< \Delta \mathbf{G}^a \Delta \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a \right. \\
+ \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^< \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Delta \mathbf{G}^< \Delta \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a \\
+ \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha}^< + \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha 1}^< \\
+ \mathbf{G}^r \Delta \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Delta \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a \\
+ \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_1^< \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_{\alpha 2}^< \\
\left. + \mathbf{G}^r \Delta \mathbf{G}^r \Sigma_2^< \mathbf{G}^a \Sigma_{\alpha}^a + \mathbf{G}^r \Sigma_2^< \mathbf{G}^a \Delta \mathbf{G}^a \Sigma_{\alpha}^a + \text{h.c.} \right]. \quad (\text{S17})$$

Perform Fourier transform from t to ω and set $\omega = 0$,

$$\begin{aligned}
J_\alpha^{(3,2)} &= \frac{e}{8\pi^3\hbar^3} \sum_\beta \iint d\varepsilon_2 d\varepsilon_1 E\left(\frac{\varepsilon_1}{\hbar}\right) E\left(\frac{\varepsilon_2 - \varepsilon_1}{\hbar}\right) E\left(-\frac{\varepsilon_2}{\hbar}\right) \int d\varepsilon \operatorname{tr} \left[\mathcal{G}_{\alpha\beta}^{(3,2)}(\varepsilon, \varepsilon_1, \varepsilon_2) \right], \\
\mathcal{G}_{\alpha\beta}^{(3,2)}(\varepsilon, \varepsilon_1, \varepsilon_2) &= -\delta_{\alpha\beta} \left(\mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \mu \mathbf{G}^< \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}^< \mu \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right) \\
&\quad -\delta_{\alpha\beta} \left(\mathbf{G}^< \mu \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \mu \mathbf{G}^< \right) \\
&\quad -\delta_{\alpha\beta} \left(\mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^< \mu \mathbf{G}^a \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^< \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right) \\
&\quad -\delta_{\alpha\beta} \left(\mathbf{G}^r \mu \mathbf{G}_{+2}^< \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^< \mu \mathbf{G}^a \right) \\
&\quad -\delta_{\alpha\beta} \left(\mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \mu \mathbf{G}^r \Sigma_\alpha^< - \Sigma_\alpha^< \mathbf{G}^a \mu \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right) \\
&\quad + \frac{eL}{2} \delta_{\alpha\beta} \left[\frac{1}{\varepsilon_1} \mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \Sigma_\alpha^< - \frac{1}{\varepsilon_2} \Sigma_\alpha^< \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right] \\
&\quad + \frac{eL}{2} (2\delta_{\alpha\beta} - 1) \left[\frac{1}{\varepsilon_1} \mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \Sigma_\beta^< \mathbf{G}^a \Sigma_\alpha^a - \frac{1}{\varepsilon_2} \Sigma_\alpha^r \mathbf{G}^r \Sigma_\beta^< \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right] \\
&\quad + \frac{eL}{2} (2\delta_{\alpha\beta} - 1) \left[\frac{1}{\varepsilon_2} \mathbf{G}^r \Sigma_\beta^< \mathbf{G}_{+2}^a \mu \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_\alpha^a - \frac{1}{\varepsilon_1} \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^r \mu \mathbf{G}_{+1}^r \Sigma_\beta^< \mathbf{G}^a \right] \\
&\quad + \frac{eL(2\delta_{\alpha\beta} - 1)}{2(\varepsilon_2 - \varepsilon_1)} \left[\mathbf{G}^r \mu \mathbf{G}_{+2}^r \Sigma_\beta^< \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_\alpha^a - \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^r \Sigma_\beta^< \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right] \\
&\quad + \frac{e^2 L^2 \delta_{\alpha\beta}}{8(\varepsilon_2 - \varepsilon_1)} \left[\frac{1}{\varepsilon_1} \mathbf{G}^r \mu \mathbf{G}_{+2}^r (\Sigma_\alpha^< - \Sigma_\alpha^<) - \frac{1}{\varepsilon_2} (\Sigma_\alpha^< + \Sigma_\alpha^<) \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right] \\
&\quad + \frac{e^2 L^2}{8(\varepsilon_2 - \varepsilon_1)} \left[\frac{1}{\varepsilon_1} \mathbf{G}^r \mu \mathbf{G}_{+2}^r (\Sigma_\beta^< - \Sigma_\beta^<) \mathbf{G}^a \Sigma_\alpha^a - \frac{1}{\varepsilon_2} \Sigma_\alpha^r \mathbf{G}^r (\Sigma_\beta^< + \Sigma_\beta^<) \mathbf{G}_{+1}^a \mu \mathbf{G}^a \right] \\
&\quad + \frac{e^2 L^2}{8(\varepsilon_2 - \varepsilon_1)} \left[\frac{1}{\varepsilon_2} \mathbf{G}^r (\Sigma_\beta^< + \Sigma_\beta^<) \mathbf{G}_{+1}^a \mu \mathbf{G}^a \Sigma_\alpha^a - \frac{1}{\varepsilon_1} \Sigma_\alpha^r \mathbf{G}^r \mu \mathbf{G}_{+2}^r (\Sigma_\beta^< - \Sigma_\beta^<) \mathbf{G}^a \right].
\end{aligned}$$

(S18)

Notations $\mathbf{A}_{+1} = \mathbf{A}(\varepsilon + \varepsilon_1)$, $\mathbf{A}_{+2} = \mathbf{A}(\varepsilon + \varepsilon_2)$, $\underline{\mathbf{A}} = \mathbf{A} - \mathbf{A}_{+1}$, $\underline{\underline{\mathbf{A}}} = \mathbf{A} - \mathbf{A}_{+2}$, $\tilde{\mathbf{A}} = \mathbf{A}_{+1} - \mathbf{A}_{+2}$ are used.

The third order integrated current is

$$J_{\alpha}^{(3)} = \frac{e}{8\pi^3 \hbar^3} \sum_{\beta} \iint d\varepsilon_2 d\varepsilon_1 E\left(\frac{\varepsilon_1}{\hbar}\right) E\left(\frac{\varepsilon_2 - \varepsilon_1}{\hbar}\right) E\left(-\frac{\varepsilon_2}{\hbar}\right) \int d\varepsilon \operatorname{tr} \left[\mathcal{G}_{\alpha\beta}^{(3)}(\varepsilon, \varepsilon_1, \varepsilon_2) \right], \quad (\text{S19})$$

$$\mathcal{G}_{\alpha\beta}^{(3)}(\varepsilon, \varepsilon_1, \varepsilon_2) = \mathcal{G}_{\alpha\beta}^{(3,1)}(\varepsilon, \varepsilon_1, \varepsilon_2) + \mathcal{G}_{\alpha\beta}^{(3,2)}(\varepsilon, \varepsilon_1, \varepsilon_2).$$