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Enhancing ECU identification security in CAN networks using distortion modeling and neural networks

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A novel technique for electronic control unit (ECU) identification is proposed in this study to address security vulnerabilities of the controller area network (CAN) protocol. The reliable ECU identification has the potential to prevent spoofing attacks launched over the CAN due to the lack of message authentication. In this regard, we model the ECU-specific random distortion caused by the imperfections in the digital-to-analog converter and semiconductor impurities in the transmitting ECU for fingerprinting. Afterward, a 4-layered artificial neural network (ANN) is trained on the feature set to identify the transmitting ECU and the corresponding ECU pin. The ECU-pin identification is also a novel contribution of this study and can be used to prevent voltage-based attacks. We have evaluated our method using ANNs over a dataset generated from 7 ECUs with 6 pins, each having 185 records, and 40 records for each pin. The performance evaluation against state-of-the-art methods revealed that the proposed method achieved 99.4% accuracy for ECU identification and 96.7% accuracy for pin identification, which signifies the reliability of the proposed approach.

KEYWORDS

intrusion detection system, electronic control unit (ECU), controller area network (CAN), machine learning, artificial neural network (ANN), digital-to-analog converter (DAC), performance matrix (PM)

1 Introduction

The modern vehicle functions as a cyber-physical system (CPS). It is equipped with numerous wireless and wired communication interfaces, along with a multitude of microcontrollers and electronic control units (ECUs). These components are interconnected via various in-vehicle networks (IVNs) (Lorenz, 2010; Grzemba, 2012; AUTOSAR, 2017; Hafeez, 2020; Hafeez et al., 2020b,c; Noori et al., 2023), such as the controller area network (CAN) (Tindell et al., 1994), local interconnected network (LIN) (AUTOSAR, 2017), media-oriented system transport (MOST) (Grzemba, 2012), and FlexRay (Lorenz, 2010). These networks link critical vehicle systems such as brakes, airbags, engine control, and safety features like electronic stability programs and adaptive cruise control. However, integrating wireless interfaces like Bluetooth and Wi-Fi with IVNs, alongside using the legacy CAN protocol for in-vehicle control, introduces significant security risks to connected autonomous vehicles (AVs) (Arif et al., 2012; Upstream Security Global Automotive Cybersecurity Report, 2019; Khalid et al., 2023).

Advances in vehicle technologies are unable to keep pace with the growing attack surfaces and vectors, leaving millions of vehicles vulnerable to a wide range of attacks,

e.g., man-in-the-middle and packet spoofing (Frösche and Stühling, 2017; Liu et al., 2017; Marchetti and Stabili, 2018). This is because the automotive industry still relies on the legacy CAN protocol for in-vehicle communication among ECUs, which lacks basic security features such as message authentication, confidentiality, and integrity, making it vulnerable to attacks (Studnia et al., 2013; Greenberg, 2015; Elkhail et al., 2021; Lee et al., 2024; Pal et al., 2024). Researchers have proposed various solutions to detect and prevent attacks on the CAN protocol for in-vehicle control networks. These methods can be classified into two categories: (i) message authentication code (MAC)-based approaches (Wolf et al., 2004; Hazem and Fahmy, 2012; Wang and Sawhney, 2014; Ueda et al., 2015; Gierlichs and Poschmann, 2016; Sugashima et al., 2016; Doan and Ganesan, 2017; Hafeez et al., 2017; Agrawal and Maiti, 2024); and (ii) intrusion detection-based approaches (Cho and Shin, 2016, 2017; Marchetti et al., 2016; Avatefipour et al., 2017; Lee et al., 2017; Marchetti and Stabili, 2017; Rieke et al., 2017; Stabili et al., 2017; Choi et al., 2018a,b; Groza and Murvay, 2018; Hafeez et al., 2018, 2019, 2020a; Kneib and Huth, 2018; Sagong et al., 2018a; Wu et al., 2018; Bellaire et al., 2022; Mohammadi and Malik, 2022; Refat et al., 2022; Baldini, 2023; Deng et al., 2023; Gul et al., 2023; Ibrahim and Safa, 2023; Liu et al., 2023; Ahmad et al., 2024; Aloraini et al., 2024; Cao et al., 2024; Elsayed and Zincir-Heywood, 2024; Jeong et al., 2024; Saini and Islam, 2024; Wang et al., 2024).

The MAC-based methods achieve security and privacy by encrypting the payload of the CAN packet before transmission (Singh et al., 2024). For instance, Wang and Sawhney (2014) demonstrated a MAC-based framework, VeCure, for CAN security. In VeCure, a 64-bit MAC was transmitted for every 64-bit message between the ECUs. Intuitively, this method exhibited high computational cost, adds 50% additional transmission overhead, and also requires a higher data rate. Ueda et al. (2015) designed an authentication mechanism for the CAN protocol to protect against spoofing attacks. The monitoring node provided the authentication code for all ECUs and verified the code for all CAN messages. Hazem and Fahmy (2012) proposed a lightweight CAN authentication protocol (LCAP). The LCAP required appending a “magic number” generated by a one-way hash function employed on the TESLA prototype (Perrig et al., 2000). This protocol still created a 25% overhead by requiring 16 bits of data field to append the authentication. The MAC-based approaches have intrinsic overhead that lowers the transmission performance, making them unreliable for CAN security (Wang and Sawhney, 2014; Avatefipour et al., 2017; Hafeez et al., 2017).

To address the limitations of MAC-based solutions, researchers have proposed intrusion detection-based approaches for CAN network traffic analysis (Hazem and Fahmy, 2012; Ueda et al., 2015; Cho and Shin, 2016; Nürnberger and Rossow, 2016; Groza and Murvay, 2018). These intrusion detection-based approaches have lower data rate requirements because they do not add additional bits during the message transmission, thus avoiding additional network overhead. Cho and Shin (2016) demonstrated a clock-based intrusion detection system (CIDS) that used ECU fingerprinting. Each ECU contained a crystal oscillator known as a clock, and the ECU fingerprinting measured the clock skewness against the received packets and detected the attack. However, Sagong et al. (2018a), and Tayyab et al. (2018) demonstrated that

CIDS can be bypassed by estimating the clock parameters. In Avatefipour et al. (2017), message authentication was performed through ECU detection by applying higher-order moments of the CAN signal in both time- and frequency domains. However, this approach needs to be more tolerant against the transmitter induction, and the performance of the system significantly decreases with an increased number of transmitters. Therefore, we need an IDS-based approach that can extract unique fingerprints from the signal, works for a higher number of transmitters, and exhibits low computational complexity.

This study presents a novel IDS-based message authentication approach to address the aforementioned limitations of existing in-vehicle security techniques. Our approach exploits the uniqueness in device-specific distortions, such as semiconductor impurities, DC offset, aliasing error, and mismatch between the nominal and measured values of electric components in a digital-to-analog converter (DAC), for message fingerprint generation. This study hypothesizes that distortions due to digital-to-analog conversion operation at the ECU output are device-dependent and can be used to link the received packet to the transmitting ECU. Therefore, we associate the received packet with a specific ECU and the ECU pin responsible for message transmission using an artificial neural network approach.

The main objective of this research is to investigate ECU-level uniqueness for a given network and pin-level uniqueness for a given ECU to authenticate the message. The proposed method relies on the distinctive physical artifacts of the DAC of the transmitting ECU for device-level fingerprinting. The imperfections in the material, design, and fabrication of the DAC contribute to distortion in the ECU signal. A statistical modeling of this distortion is developed, and used as a feature vector for transmitter identification (i.e., transmitting ECU and ECU-Pin) through a neural network architecture. Thus, the main contributions of the study are:

- A mathematical model of the distortion sources, including imperfections in the material, design, and fabrication of the DAC.
- A statistical model of the device ECU-level distortion for transmitter identification.
- A proposal that different transmitting pins in a single device have unique distortion and can be used for ECU-pin identification.

The rest of the study is organized as follows: Section 2 describes the system model and outlines sources of device-specific distortion in the CAN signal. Experimental setup, dataset, performance measures, results, and analysis are provided in Section 3. This section also provides a comparison of our method with the current state-of-the-art research. Concluding remarks and future directions are discussed in Section 4.

2 System model

Figure 1A shows a *subnet* of a CAN bus that contains $\gamma = \{1, 2, \dots, n\}$ ECUs represented as $E_{(i)}$, with a fingerprinting unit $E_{(s)}$ that sniffs the CAN bus and acquires an analog signal

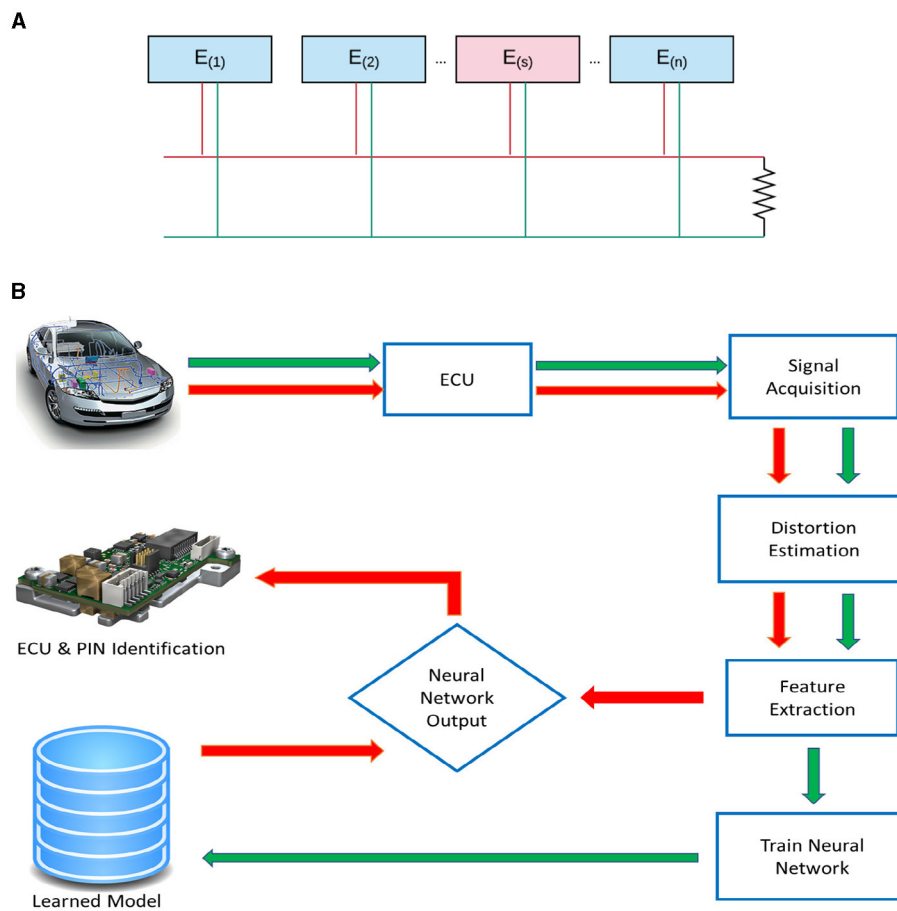


FIGURE 1

Generalized architecture of the CAN bus (A) and block diagram of the proposed device identification system (B). Green arrows $\rightarrow$ here represent the learning phase, and red arrows $\rightarrow$ here represent the testing phase of device identification.

$y_{(i)}^{(a)}(t)$ transmitted by the i^{th} ECU, $E_{(i)}$, where $i \in \gamma$. The $E_{(s)}$. As represented in Equation 1, the fingerprinting ECU, $E_{(s)}$, converts the analog measurement at its input, $y_{(i)}^{(a)}(t)$, to its digital equivalent, $y_{(i)}^{(a)}(n)$. The fingerprinting ECU, $E_{(s)}$, computes the distortion, $d_{(i)}(n)$, for $E_{(i)}$ in the received signal, $y_{(i)}^{(a)}(n)$, by subtracting it from the expected signal level, $y_{(i)}^{(e)}(n)$, i.e.,

$$d_{(i)}(n) = y_{(i)}^{(a)}(n) - y_{(i)}^{(e)}(n) \quad (1)$$

Here, $\{d_{(i)}(n) \in n_{(i)} \mid n_{(i)} : [L \rightarrow U]\}$, where $L = -0.10$ and $U = 0.10$ represent the lower and upper distortion values, respectively.

The $d_{(i)}(n)$ is then used for feature extraction to generate the feature vector $\mathbf{x}_{(r)} = \{x^{(1)}, x^{(2)}, \dots, x^{(m)}\}$ for $E_{(i)}$, where m represents the number of features, $r \in R$, and R is the total number of records. The extracted features are used in an artificial neural network architecture for training data and evaluation of test data. Shown in Figure 1B, is the generalized architecture of the CAN bus (Figure 1A) and block diagram of the proposed device identification system (Figure 1B). A brief description of each processing block is provided in the following subsections.

2.1 Signal acquisition

The $E_{(s)}$ acquires the analog signal $y_{(i)}^{(a)}(t)$ generated by $E_{(i)}$ and converts it into a digital signal $y_{(i)}^{(a)}(n)$ as represented in Equation 2.

$$y_{(i)}^{(a)}(n) = y_{(i)}^{(a)}(t)|_{t=nT_s}, T_s = 50 \times 10^{-9} \quad (2)$$

The $T_s = 50 \times 10^{-9}$ represents a sampling time of 50 ns and a sampling rate of 20 MSa/sec for the signal. The reason to generate $y_{(i)}^{(a)}(n)$ is that the $y_{(i)}^{(a)}(t)$ occurs at infinite instants of time, thus demanding large memory to get stored. However, the analog-to-digital conversion is performed because the $E_{(s)}$ have limited memory. The number of bits required to store each sample is n ; in our case, $n = 8$ bits. To extract distortion, we compute the expected signal $y_{(i)}^{(e)}(n)$ from $y_{(i)}^{(a)}(n)$ using Equation 3:

$$y_{(i)}^{(e)}(n) = \begin{cases} 3.5v & : 3.3 < y_{(i)}^{(e)}(n) < 3.7 \\ 2.5v & : \text{Otherwise} \end{cases} \quad (3)$$

Equation 3 here represents the mapping of the received signal, $y_{(i)}^{(e)}(n)$, onto the CAN high and CAN low voltage levels, e.g.,

3.5V and 2.5V. Specifically, the received signal, $y_{(i)}^{(e)}(n)$, is mapped to 3.5 v if $y_{(i)}^{(a)}(n)$ is between 3.3 v and 3.7 v; otherwise it will be 2.5 v. The signal $y_{(i)}^{(e)}(n)$ represents the expected signal from $E_{(i)}$; however, it has been observed through extensive analysis of CAN communication signals that the actual signal levels differ from the expected signal levels. As shown in [Figure 2A](#), the waveform of the CAN signal captured using a DS1012A oscilloscope for the $y_{(i)}^{(a)}(n)$ significantly differs from the expected signal $y_{(i)}^{(e)}(n)$ due to the distortion mainly attributed to semiconductor impurities, the mismatch between nominal values and measured values of electric components, the aliasing error of the finite impulse response (FIR) filter, and the DC offset of the DAC. Moreover, these imperfections are device-specific and can be used for ECU fingerprinting.

2.2 Distortion extraction

The imperfections observed in the signal acquisition stage are used for fingerprinting the ECUs. The fingerprinting is quantified in distortion modeling, which is acquired in the *Density Estimation* stage. Before distortion modeling, we acquire distortion as shown in [Figure 2B](#), which is represented in [Equation 4](#).

$$d_{(i)}(n) = y_{(i)}^{(a)}(n) - y_{(i)}^{(e)}(n). \quad (4)$$

There are four main reasons for distortions, which are discussed as follows:

2.2.1 Mismatch of nominal and measured values of electric components

Imperfections in the electric components are one of the sources of $d_{(i)}(n)$. These imperfections can be described as deviations of the measured values of electric components from their nominal values. Let $R_{o(i)}$ be the value of the feedback resistor of the $E_{(i)}$, and $\delta_{R(i)}$ represents the deviation from the nominal value, commonly known as the tolerance level. The actual resistance $R_{a(i)}$ can then be expressed as [Equation 5](#), which is as follows:

$$R_{a(i)} = R_{o(i)} + \delta_{R(i)}. \quad (5)$$

Let $d_{(i)}^{(R)}(n)$ represent the distortion due to $\delta_{R(i)}$, which is the first cause of distortion at the DAC output. The purpose of a DAC in an ECU is to convert bits into a physical signal in the form of voltage, as shown in [Figure 3](#). The reason for this conversion is that the signal propagates through a channel in the form of a physical signal.

2.2.2 Semiconductor impurities

During semiconductor device manufacturing, the impurities in the silicon used for manufacturing semiconductors cannot be removed entirely ([Pizzini et al., 1986](#); [Pizzini, 2010](#); [Roskill, 2011](#); [Yoshida and Langouche, 2015](#)). These impurities in the semiconductor can cause flicker distortion (also known as 1/f noise) at the DAC output ([Godfrey et al., 2015](#)). We represent the distortion due to semiconductor impurities as $d_{(i)}^{(f)}(n)$, which

is one of the sources of distortions at the DAC output. The imperfections in the device material and fabrication process are the other contributing factors to the observed distortion.

2.2.3 Non-ideal behavior of low-pass filter

Another contributing factor in distortion is an aliasing error due to the non-ideal behavior of the low-pass filter used in digital-to-analog conversion. The digital signal, acquired through the continuous signal sampling operation, introduces periodic repetition of its spectra. Let $Y_{(i)}^{(e)}(f)$ represent the Fourier transform of $y_{(i)}^{(e)}(n)$, which is input in DAC, and $Y_{(i)}^{(e)}(f)$ represents the Fourier transform in form of the continuous signal. Low-pass filtering is used to filter out $Y_{(i)}^{(e)}(f)$ from $Y_{(i)}^{(e)}(f)$ to avoid unwanted copies, which is commonly implemented using a FIR filter realization. [Equation 6](#) represents the relationship between input and output of the DAC,

$$Y_{(i)D}^{(e)}(f) = \sum_{n=-\infty}^{\infty} Y_{(i)}^{(e)}(f - nf_s) \quad (6)$$

Let $H(f)$ represent the transfer function of the FIR filter; the output of the FIR filter can be expressed through [Equation 7](#).

$$Y_{(i)FIR}^{(e)}(f) = Y_{(i)D}^{(e)}(f).H(f) \quad (7)$$

Ideally, $Y_{(i)FIR}^{(e)}(f)$ should be the same as $Y_{(i)}^{(e)}(f)$, but due to an aliasing error these values differ. The non-ideal behavior of FIR realization introduces aliasing at the DAC output. Let $d_{(i)}^{(a)}(n)$ denote distortion due to the non-ideal behavior of the low-pass filter realization for the $E_{(i)}$. This is the third cause of distortion in the output of the DAC.

2.2.4 DC offset error

The DC offset error in the DAC is another source of distortion ([Pavan et al., 2017](#)). Ideally, the dominant bit level is 3.5 V and the ideal recessive bit level is 2.5 V, as shown in [Figure 3](#), but DC offset $d_{(i)}^{(o)}(n)$ is added to the ideal voltage value due to grounding issues in the DAC. The total distortion $d_{(i)}(n)$ due to DAC for the $E_{(i)}$ can be expressed as [Equation 8](#),

$$d_{(i)}(n) = d_{(i)}^{(R)}(n) + d_{(i)}^{(f)}(n) + d_{(i)}^{(a)}(n) + d_{(i)}^{(o)}(n). \quad (8)$$

Similarly, we can derive from [Equation 4](#) that:

$$y_{(i)}^{(a)}(n) = y_{(i)}^{(e)}(n) + d_{(i)}(n). \quad (9)$$

Therefore, [Equation 9](#) validates our hypothesis that the distortion added to the received signal is dynamic and can be an effective measure for fingerprinting the ECUs.

Furthermore, the device-specific distortion is also unique for each pin within the $E_{(i)}$, represented as $E_{(i,l)}$, where $\{l \in \zeta \mid \zeta = 1, 2, \dots, \lambda\}$ are the total number of pins within $E_{(i)}$. This also elaborates that we need to determine the affected ECU and the relevant pin for attack modeling. Additionally, pin-level artifacts

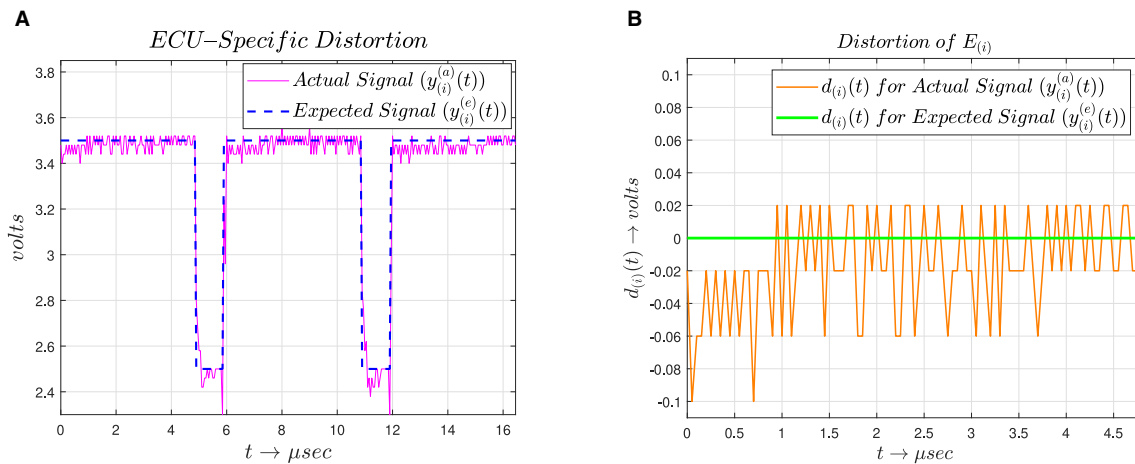


FIGURE 2 Screenshots of the actual and expected waveform vs. ECU signal. (A) Actual and expected waveform. (B) Distortion in the ECU signal.

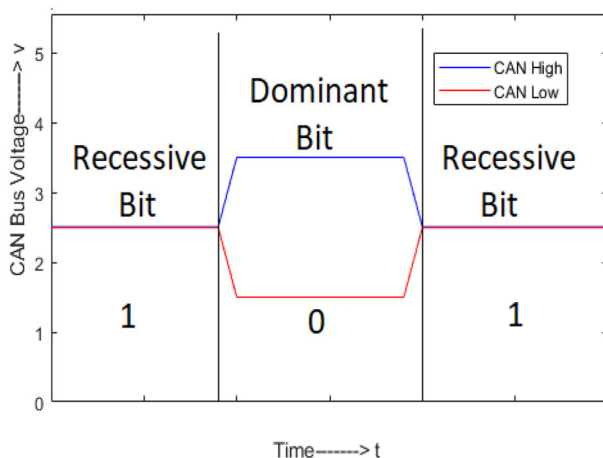


FIGURE 3 Ideal voltage levels for CAN.

effectively detect spoofing attacks launched from a different pin of the same ECU. Recently, [Sagong et al. \(2018b\)](#) demonstrated a voltage-based attack that permanently damages a target ECU pin. A voltage-based attack is launched by an adversary, which sends a high voltage through a pin (which has a maximum capacity of 5 V) to damage the pin permanently. However, these attacks can be avoided by fingerprinting the $E_{(i,l)}$.

2.3 Density estimation

After the distortion modeling, we use $d_{(i)}(n)$ for histogram generation. These histograms will then be used as fingerprints for $E_{(i)}$ and $E_{(i,l)}$. In order to make a histogram, we need to group the distortion values in m histogram bins with step size β , which can be computed as in [Equation 10](#).

$$\beta = \left(\frac{U - L}{m} \right) \quad (10)$$

Where U and L represent the lower and upper values for distortion, and m represents the number of bins. In our case, $m = 200$; thus, step size β becomes 10^{-3} . The histogram $h_{(i)}(k)$ for $N = 1500$ samples of $d_{(i)}(n)$ is computed using [Equation 11](#). Here, $k = \{0, 1, 2, \dots, m\}$.

$$h_{(i)}(k) = \sum_{n=1}^N \left[\delta \left(\frac{d_{(i)}(n)}{\beta} + 100 \right) + h_{(i)}(k) \right] \quad (11)$$

Where $\delta(\cdot)$ denotes the Kronecker delta function ([Adnan et al., 2018](#)), that can be computed through [Equation 12](#).

$$\delta(n-k) = \begin{cases} 1 & : n = k \\ 0 & : \text{Otherwise} \end{cases} \quad (12)$$

[Figure 4](#) shows the histogram $h_{(i)}(k)$ of $E_{(1)}-E_{(7)}$. Afterwards $h_{(i)}(k)$ is used as feature set $X_{(i)} = \{x^{(1)}, x^{(2)}, \dots, x^{(m)}\}$.

2.4 ANN-based model learning

An artificial neural network (ANN)-based model is used to identify the source ECU and the corresponding pin. For this the ANN gets $X_{(r)} = \{\mathbf{x}_{(1)}, \mathbf{x}_{(2)}, \dots, \mathbf{x}_{(R)}\}$ as input-set and corresponding ECU- and pin-labels $Y_{(r,e)} = \{y_{(1,e)}, y_{(2,e)}, \dots, y_{(R,e)}\}$, $Y_{(r,p)} = \{y_{(1,p)}, y_{(2,p)}, \dots, y_{(R,p)}\}$ respectively; and it predicts the ECU-labels and corresponding pin-labels as $\hat{Y}_{(r,e)} = \{\hat{y}_{(1,e)}, \hat{y}_{(2,e)}, \dots, \hat{y}_{(R,e)}\}$ and $\hat{Y}_{(r,p)} = \{\hat{y}_{(1,p)}, \hat{y}_{(2,p)}, \dots, \hat{y}_{(R,p)}\}$ respectively, where $e \in \gamma$, $p \in \zeta$, and $r \in R$. The classifier is trained on the dataset with three hidden layers, each with ν neurons (in our case $\nu = 10$), using the "scaled conjugate gradient backpropagation" method for weight optimization. For hyperparameter tuning, we used *GridSearchCV*.

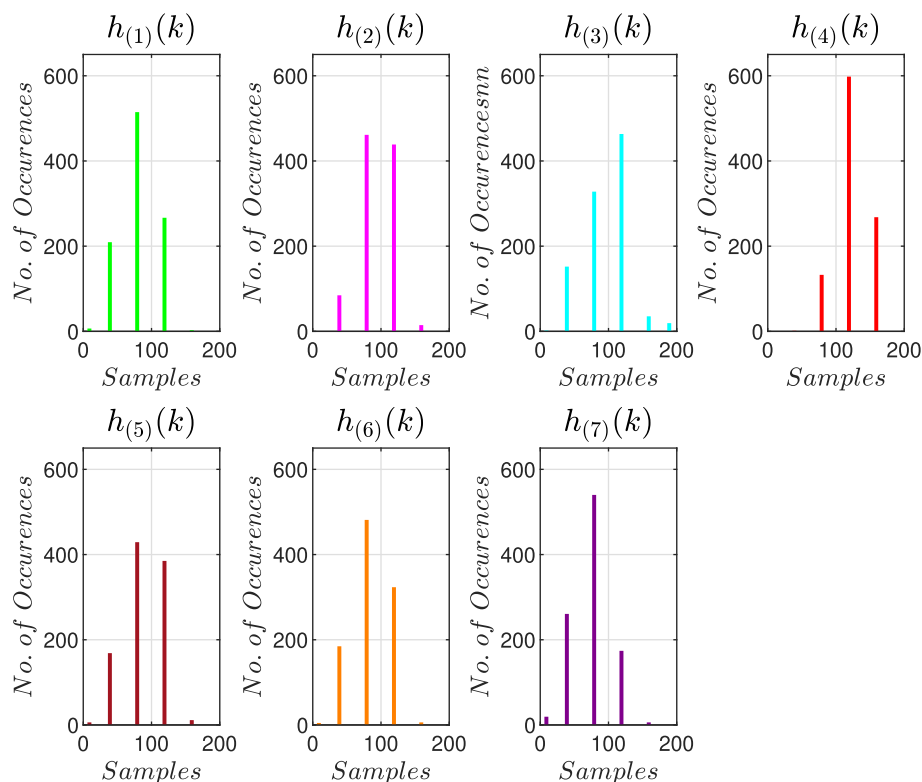


FIGURE 4
Histogram of $d_{ij}(n)$ for ECUs $E_{(1)}-E_{(7)}$.

(a scikit-learn library) to search for the best parameter values from the given set of parameters. Several hidden layers and neurons were tested using this library. Empirically, we chose three hidden layers and determined the number of neurons (Arif et al., 2023). During the training phase, the model learns the weight vector represented by: $\omega_e = \{w^{(1)}, w^{(2)}, \dots, w^{(m)}\}$ for all $E_{(i)}$ and $\omega_p = \{w^{(1)}, w^{(2)}, \dots, w^{(m)}\}$ for all $E_{(i,l)}$. The output of both networks is then merged as the final output during the testing phase (Figure 5). The ANN architecture for ECU recognition is presented in Table 1, and the same architecture is also used for ECU-pin recognition.

Table 1 summarizes the neural network structure used in this study, including the input layer, middle (hidden) layers, output layer, and error correction mechanism. The input layer, represented as $\vec{x}(r)$, receives data with dimensions $1 \times m$. The middle layers perform computations on the input vector $\vec{b}$ using a weight matrix U , resulting in an output vector $\vec{c}$ through an activation function f . Similarly, the output layer processes the input vector $\vec{d}$ using a weight matrix w , resulting in an output vector $\vec{e}$ through another activation function g . The error correction mechanism is described by the calculation of the cost function E and the update equations for the weight matrices w and U , denoted as ΔW_{ij} and ΔU_{mv} , respectively. These equations involve gradients of the error with respect to the weights and are updated using learning rates α and β for w and U , respectively. This comprehensive structure outlines the flow of information and the mechanism for adjusting weights to minimize errors during the training process of the neural network.

3 Results and discussion

3.1 Experimental setup

The proposed approach evaluates inter-class (amongst ECUs) and intra-class (amongst ECU pins) variability for message authentication. For inter-class variability, seven ECUs (transmitters) of the same make and model were used in this study, and data was recorded through the CAN-High (CANH) pin. For intra-class variability, six DAC pins of the same ECU were analyzed to determine the pin-level characteristics.

The hardware comprised seven Arduino UNO-R2 micro-controller kits; seven CAN-Bus shield boards with MCP 2515 CAN-bus controllers and MPC 2551 CAN transceivers; and a DS O1012A oscilloscope to record the voltage samples with a sampling rate of 20Msa/s, with 100 MHz bandwidth. Matlab R 2018a software was used for statistical data analysis of the sampled signals. A computer simulation that continuously transmitted messages from different ECUs and pins was written. Afterward, these messages were then used as the dataset for model training and evaluation.

3.2 Dataset description

The ECU identification dataset comprised 1,295 (7×185) records with 1,500 samples in each record. Whereas, for pin-level identification, a dataset was collected for six different pins of each transmitting ECU with 40 records for each pin. The dataset used

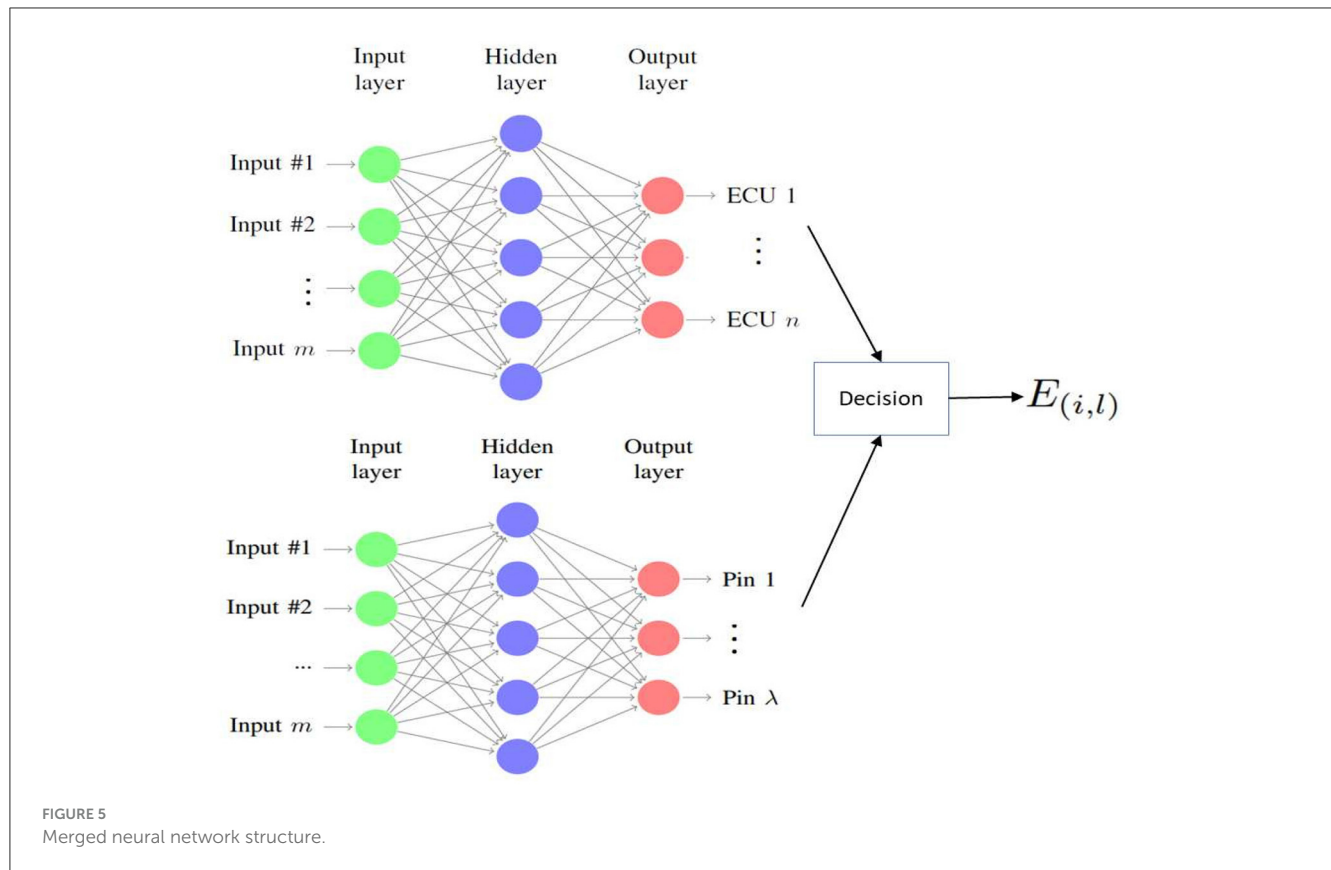


TABLE 1 Summary of neural network structure for every $E_{(i)}$ used in this study.

Input		
Input:	$\vec{x}_{(r)} = (x^{(1)}, x^{(2)}, \dots, x^{(m)})$	$\dim(\vec{x}) = 1 \times m$ (I.1)
Middle (hidden) layers (1-3)		
Input:	$\vec{b} = U \vec{X}$	$\dim(\vec{b}) = 1 \times v$ (I.2)
Output:	$\vec{c} = f(b)$	$\dim(\vec{c}) = 1 \times v$ (I.3)
U:	$m \times v$ weight matrix	
f:	$\frac{1}{1 + e^{-b}}$	
Output layer		
Input:	$\vec{d} = w_e \vec{c}$	$\dim(\vec{d}) = 1 \times v$ (I.4)
Output:	$\vec{e} = g(d)$	$\dim(\vec{e}) = 1 \times v$ (I.5)
w_e :	$v \times m$ weight matrix	
g:	$\frac{1}{1 + e^{-d}}$	
Error correction		
Cost:	$E = - \sum_{i=1}^n [y_{(r,e)} \log(c)]$	(I.6)
ΔW_{ij}	$= -\alpha \partial E / \partial W_{vm} = \alpha \delta_i c_j$	(I.7)
ΔU_{mv}	$= -\beta \partial E / \partial U_{mv}$	(I.8)

here was collected in the same environment, i.e., under the same temperature, and using an identical message to observe the unique variations of the digital signals.

To ensure the proposed classification scheme was robust and not overfitting to the training data, we implemented several validation techniques. The model was evaluated using various metrics: the dataset was randomly split into training and testing sets with a training ratio of 70% and a testing ratio of 30%. In addition, regularization techniques such as L1 (Lasso) were incorporated into the model. Moreover, early stopping criteria were introduced during the training phase. In addition to accuracy, the model was evaluated using precision, recall, F1-score, and AOC. By implementing these techniques, we ensured that the proposed classification scheme was validated rigorously and could generalize well to new, unseen data.

3.3 Performance evaluation measures

We used precision, recall, F_1 score, accuracy, and error rate as performance evaluation measures. To evaluate the effectiveness of the proposed method, we determined how many ECUs were correctly identified in response to messages sniffed by $E_{(i)}$. Let TP represent the true positive rate, FP represents the false positive rate, TN represents the true negative rate, and FN represents the false negative rate. Then precision can be defined as follows in Equation 13.

$$Precision = \left(\frac{TP}{TP + FP} \right) \quad (13)$$

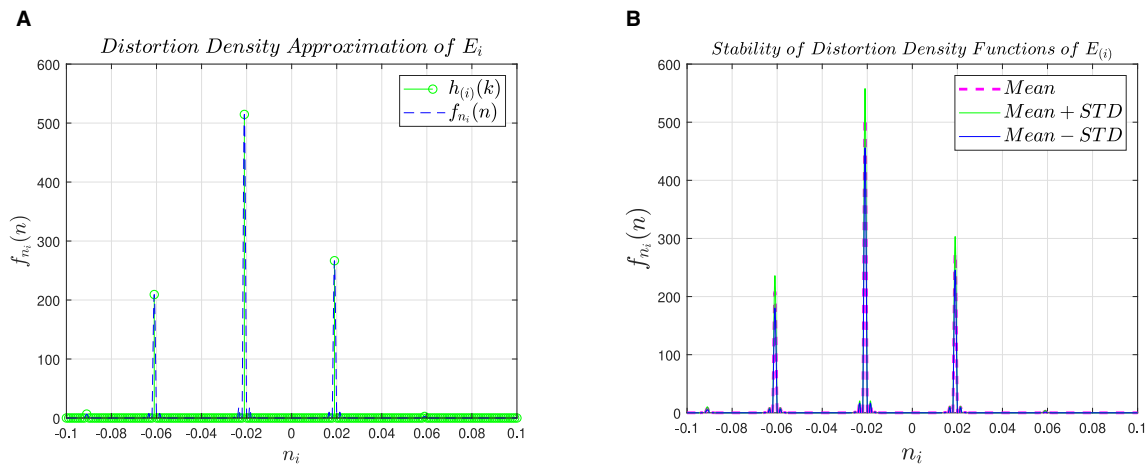


FIGURE 6

Density approximation of distortion and effectiveness of our feature set. (A) Density approximation of distortion. (B) Effectiveness of our feature set.

Precision was used to measure the ratio of the true instances against the retrieved instances for a particular class. To measure the sensitivity, we used the recall rates that can be computed as follows in Equation 14.

$$Recall = \left(\frac{TP}{TP + FN} \right) \quad (14)$$

Recall was computed to measure the total number of retrieved relevant instances. In order to combine both measures, i.e., precision and recall, we used the F_1 Score that was computed as below in Equation 15.

$$F_1 \text{ score} = 2 \times \left(\frac{Precision \times Recall}{Precision + Recall} \right) \quad (15)$$

The higher F_1 score signifies the robustness of the classification approach. In order to evaluate the overall performance by considering all the classes together, we computed the accuracy as follows in Equation 16.

$$Accuracy = \left(\frac{TP + TN}{TP + TN + FP + FN} \right) \quad (16)$$

Accuracy was computed to measure all instances that were correctly classified, despite the fact, whatever class they belong to. Moreover, by using the accuracy value, we also computed the overall error rate as follows in Equation 17.

$$Error \text{ rate} = 1 - Accuracy \quad (17)$$

3.4 Feature stability analysis

This experiment aims to validate that different ECUs, even of the same make and model, introduce different artifacts while transmitting an identical message, and this uniqueness can be exploited to counter spoofing attacks. To achieve this goal, all ECUs transmitted the same messages over the same channel with constant settings regarding temperature and environment. Data

was recorded for each ECU with identical channel inputs and transmission parameters to validate the claim of ECU-specific distortion. To verify the uniqueness, we estimated the distortion density function by applying the Spline function (Hirst and Espesser, 1993) over the histogram $h_{(i)}(k)$ to get $f_{n_i}(n)$ as shown in Figure 6A. The estimated distortion distribution represents the physical characteristics of each ECU. In order to find the stability, we generated the $f_{n_i}(n)$ of each ECU 100 times and computed the mean and standard deviation (STD). From Figure 6B, it can be observed that the difference between the mean and mean $\pm$ STD is negligible, which shows that the feature set remains constant over time for each ECU. Hence, it is proved that the proposed feature extraction approach induces the unique attributes for ECU representation that make it effective for ECU identification.

To further validate the uniqueness attribute of the proposed method, we plotted $f_{n_i}(n)$ for seven ECUs as shown in Figure 7, which clearly shows that each ECU has a unique representation. The benefit of the uniqueness is that the attacker cannot replicate an ECU's profile, thus signifies that our approach is robust against spoofing attacks.

3.5 ECU-level identification

The purpose of this experiment is to validate that any two ECUs exhibit different distortion profiles, i.e., any two ECUs, even from the same make and model, introduce different artifacts into the transmitted signal. To achieve this objective, seven ECUs were used to transmit the same message over the same channel with the same environmental settings, e.g., temperature, environment, etc. A seven-class multilayered neural network classifier is trained and tested on the data collected using data acquisition method described in Section 3. Shown in Tables 2, 3, and Figure 8 are the confusion matrix, performance evaluation in terms of selected performance metric, and detector receiver operating characteristic (ROC), respectively.

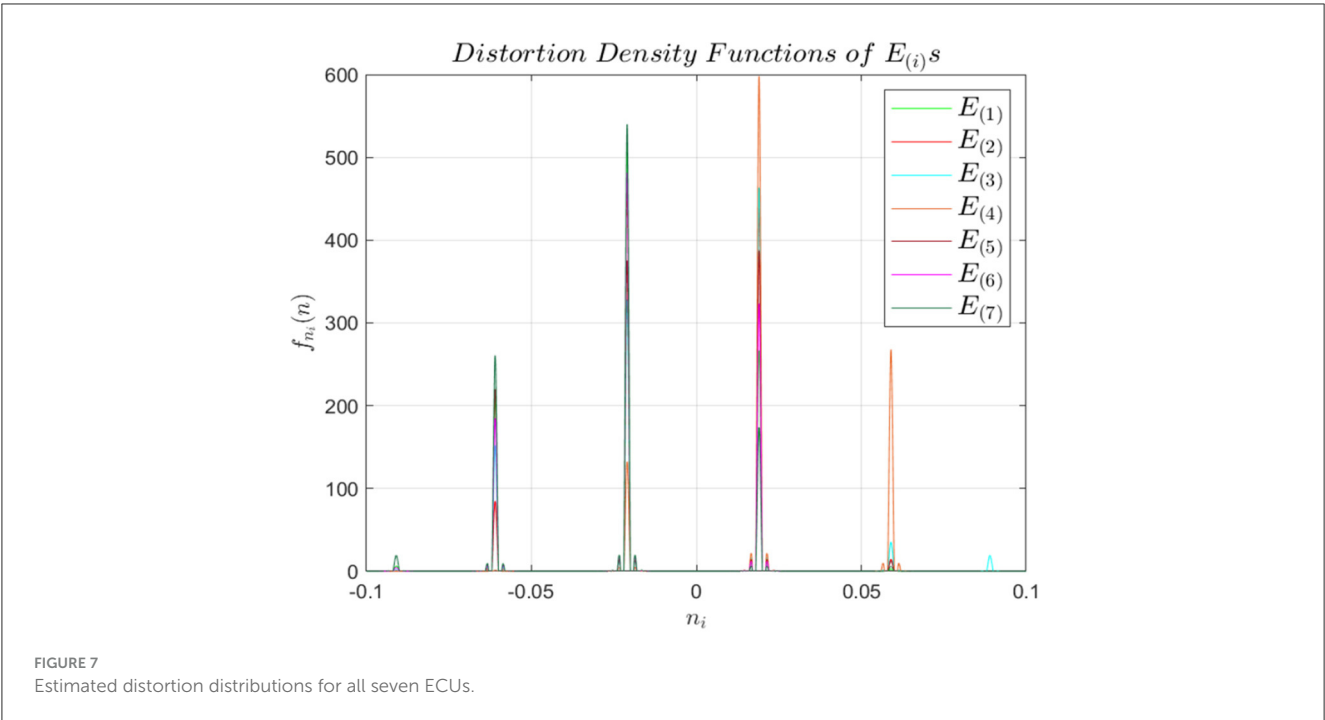


TABLE 2 Confusion matrix for ECU-level classifier.

		Target class							Total %
-	-	E ₍₁₎	E ₍₂₎	E ₍₃₎	E ₍₄₎	E ₍₅₎	E ₍₆₎	E ₍₇₎	
Predicted class	E ₍₁₎	185	0	0	0	0	0	0	100
	E ₍₂₎	0	184	2	0	0	0	0	98.9
	E ₍₃₎	0	1	183	0	0	0	0	99.5
	E ₍₄₎	0	0	0	185	0	0	0	100
	E ₍₅₎	0	0	0	0	183	3	0	98.4
	E ₍₆₎	0	0	0	0	2	182	0	98.9
	E ₍₇₎	0	0	0	0	0	0	185	100
	Total %	100	99.5	98.9	100	98.9	98.4	100	99.4

From Table 2, it can be observed that the proposed method achieves very high accuracy for ECU classification. The high accuracy signifies that the distortion introduced in each ECU due to DAC imperfections and semiconductor impurities is unique, thus resulting in high accuracy for ECU identification. Moreover, it also validates our hypothesis that the distortion due to DAC and semiconductor impurities has the potential for ECU fingerprinting for attack detection. It can also be observed from Table 2 that $E_{(1)}$, $E_{(4)}$, and $E_{(7)}$ have 100% detection rates, which is mainly associated with the high distortion values appearing in the form of high peaks as shown in Figure 4. Moreover, by analyzing Table 2 from the perspective of Figure 4, it can also be observed from T that if distortion is concentrated in a certain region, it increases the inter-class variability, which is one of the targets of this research.

Similarly, Table 3 shows that $E_{(1)}$, $E_{(4)}$ and $E_{(7)}$ have 100% precision, recall, accuracy and F_1 Score rates. Furthermore, $E_{(2)}$

has a 99.5% recall, which means it has only one false negative record; the other performance evaluation rate slightly drops. On the other hand, $E_{(3)}$ has 99.5% precision; it has only one false positive, therefore slightly affects the precision.

Figure 8A shows the graphical representation of the ECU identification results. A high correlation among the performance evaluation measures for all ECUs signifies the reliability of the proposed method. Furthermore, the area under the curve analysis of the ROC plots, as shown in Figure 8B, confirms our claim that unique distortions in the output of electronic devices can be used for fingerprinting and identification. This has many applications, ranging from hardware authentication to attack detection and localization.

3.6 Pin-level identification

The purpose of this experiment is to validate that different ECUs exhibit different distribution profiles, meaning that is, different pins of the same transmitting ECU introduce different artifacts into the transmitted signal. Pin-level fingerprinting can be utilized for reliable attack profile generation. To achieve this objective, six pins of the same transmitting ECU were used to transmit the same message over the same channel with the same environmental settings, such as temperature and environment. A six-class multilayered neural network classifier was trained and tested on the data collected using the data acquisition method described in Section 3. Shown in Table 4 is the performance of the proposed method in the confusion matrix (in terms of the number of samples per class).

Shown in Table 5 is the performance of the proposed method in terms of selected performance evaluation measures, including precision, recall, F_1 score, accuracy, and error rate. It can be

TABLE 3 Performance matrix of the ECU-level classifier.

-	Precision (%)	Recall (%)	Accuracy (%)	F ₁ score (%)	ERR (%)
E ₍₁₎	100	100	100	100	0
E ₍₂₎	98.9	99.5	99.8	99.2	0.2
E ₍₃₎	99.5	98.9	99.8	99.2	0.2
E ₍₄₎	100	100	100	100	0
E ₍₅₎	98.4	98.9	99.6	98.6	0.4
E ₍₆₎	98.9	98.4	99.6	98.6	0.4
E ₍₇₎	100	100	100	100	0

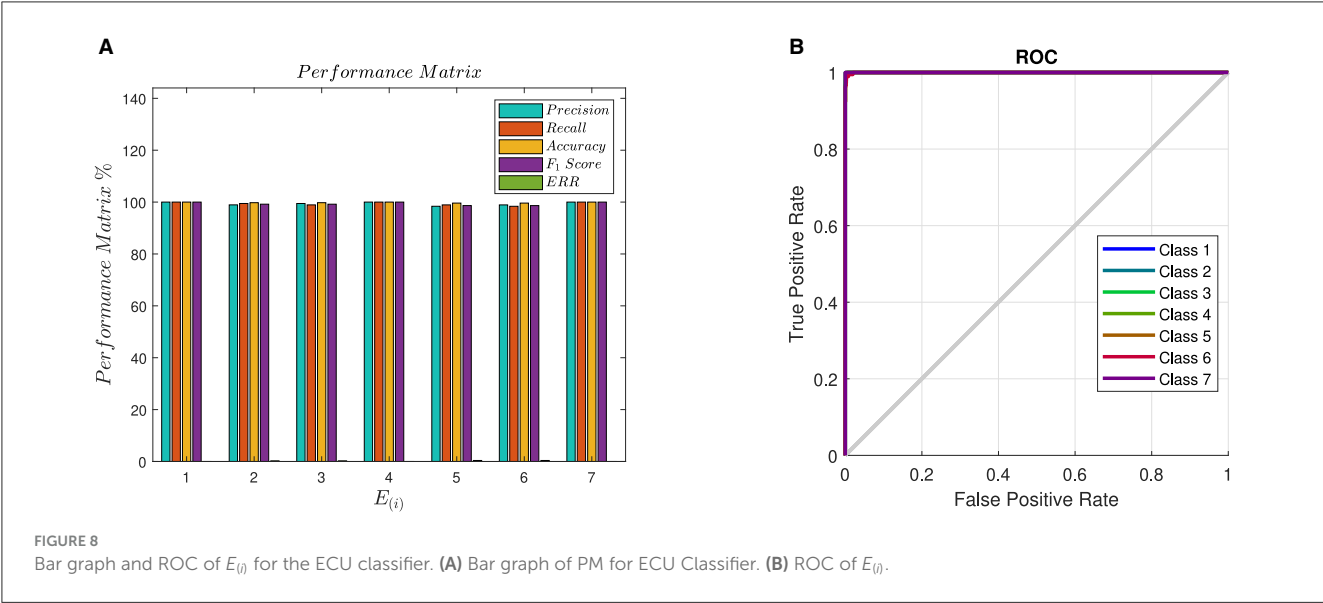


TABLE 4 Confusion matrix for pin-level classification.

		Target class						Total %
-	-	Pin ₍₁₎	Pin ₍₂₎	Pin ₍₃₎	Pin ₍₄₎	Pin ₍₅₎	Pin ₍₆₎	
Predicted class	Pin ₍₁₎	38	2	0	0	0	1	92.7
	Pin ₍₂₎	2	37	0	0	0	0	94.9
	Pin ₍₃₎	0	0	40	0	0	0	100
	Pin ₍₄₎	0	0	0	40	0	0	100
	Pin ₍₅₎	0	0	0	0	39	1	97.5
	Pin ₍₆₎	0	1	0	0	1	38	95
	Total %	95	92.5	100	100	97.5	95	96.7

observed from Tables 4, 5 that pins # 3 and 4 achieved perfect detection, with 100% precision, recall rate, accuracy, and F₁ score. Overall, the proposed method achieved an overall 96.7% detection rate for pin-level identification. Although detection rates for pin-level identification are slightly lower than ECU-level detection. The same results are graphically presented in Figure 9A. The area under the curve results, as presented in Figure 9B, show that the pin-level detection of the proposed approach is still satisfactory. It is important to highlight that the notion of pin-level detection is a novel concept; research efforts focused on finding why edge

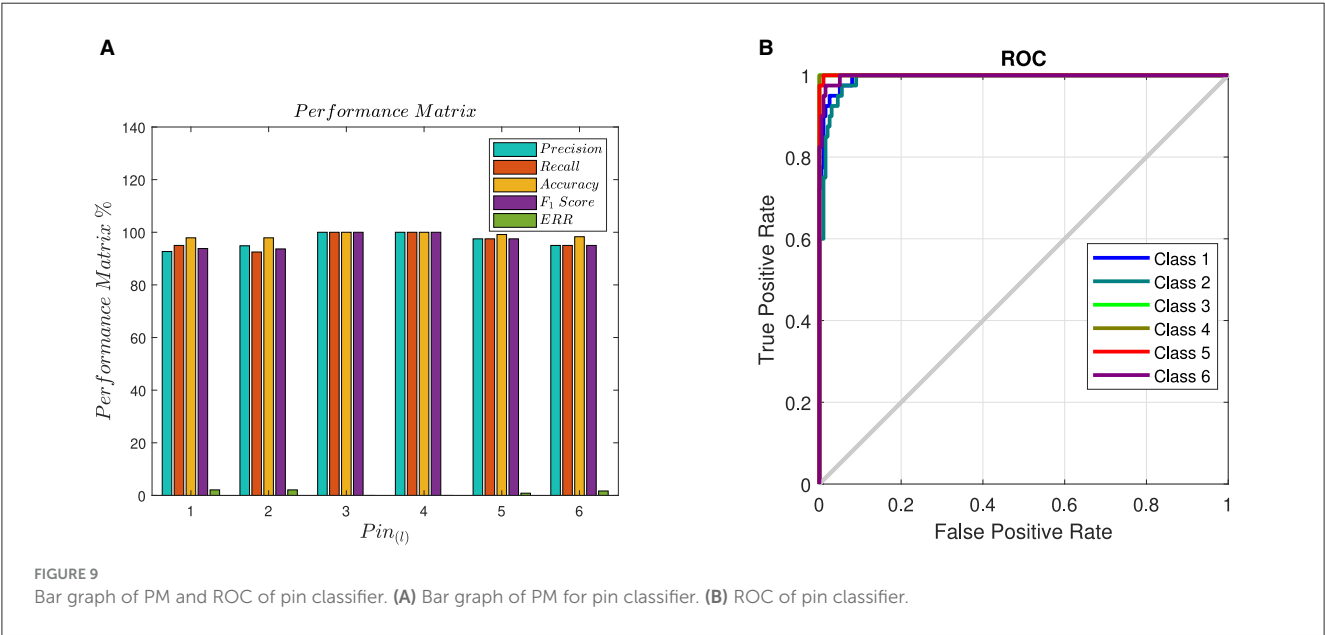
pins, e.g., pins 1, 2, 5, and 6, are expected to generate more interesting results.

3.7 Comparison against state-of-the-art

In this section, the proposed method is compared against state-of-the-art methods that are also doing the ECU identification. The performance is compared against ECU detection using Viden (Cho

TABLE 5 Performance matrix of pin-level classification.

-	Precision (%)	Recall (%)	Accuracy (%)	F ₁ score (%)	ERR (%)
Pin ₍₁₎	92.7	95	97.9	93.8	2.1
Pin ₍₂₎	94.9	92.5	97.9	93.7	2.1
Pin ₍₃₎	100	100	100	100	0
Pin ₍₄₎	100	100	100	100	0
Pin ₍₅₎	97.5	97.5	99.1	97.5	0.9
Pin ₍₆₎	95	95	98.3	95	1.7



and Shin, 2017), Inimitable characteristics of CAN signal (Choi et al., 2018a), and VoltageIDS (Choi et al., 2018b).

3.7.1 Comparative studies

Cho and Shin (2017) proposed a method called *Viden* that utilized the voltage profile of acknowledgment (ACK) bits for transmitter identification. In the first phase, the ACK bit was used to determine if the message originated from the genuine transmitter. Afterward, voltage measurements were used to generate ECU fingerprints. Based on these fingerprints, the attacker ECU was identified. In Choi et al. (2018a), a monitoring unit was installed in the vehicle that analyzes the electrical CAN signals and computes the statistical features. These features were then classified to identify the ECU. In Choi et al. (2018b), ECU detection based on inimitable voltage characteristics technique was proposed. The feature vectors proposed by Choi et al. (2018a) were extended both in time- and frequency domains and were classified for ECU identification by Choi et al. (2018b).

3.7.2 Performance comparison

Table 6 shows the performance comparison of our method against (Cho and Shin, 2017; Choi et al., 2018a,b). The results show

that our method is giving higher accuracy compared to Choi et al. (2018a,b), whereas it is giving almost the same performance as Cho and Shin (2017). However, the main advantage of our method is that feature extraction and message authentication can be done in any part of the signal without latency. Whereas, in the case of Viden, the voltage profile is estimated for message authentication during reception of the ACK bit, but it also introduces latency. Hence, from the aspect of latency, our method is more robust than the Viden.

4 Conclusion

In this research study, a novel approach for electronic control unit (ECU) identification in a vehicular network is presented. The main motivation behind ECU identification is that the CAN protocol lacks sender ECU information, which makes it prone to spoofing attacks. Our proposed device identification framework allows cybersecurity professionals to detect and localize transmitting devices even without transmitter information.

This study utilizes electronic device-specific distortion for device identification. The proposed approach utilizes unique but physically unclonable distortions in electronic device output stemming from the material's intrinsic imperfections and the manufacturing process. We have demonstrated that device-specific

TABLE 6 Comparison with other methods.

Research study	Method	Accuracy
Cho and Shin (2017)	Viden	99.57%
Choi et al. (2018a)	Inimitable Char. of CAN Signal	96.48%
Choi et al. (2018b)	VoltageIDS	95.54%
Our method	Distortion based IDS	99.4%

distortions are unique for all ECUs, even from the same make, model, and manufacturer. We have shown that device-specific distortions can be used to detect and localize devices on the network. We realized the proposed device identification framework using a 4-layered artificial neural network architecture. We have also demonstrated that the proposed framework can be used for device output-pin identification (e.g., device pin being used for message transmission).

For future study, we will investigate the robustness and reliability of the proposed framework under different environmental conditions, i.e., temperature, humidity, electromagnetic interference, sampling rate, etc. We also plan to evaluate its performance of the proposed system in real-time applications. In this regard, fuzzy logic-based decision modeling will be considered for the proposed ECU and pin identification framework under various environmental and data acquisition conditions.

Data availability statement

The raw data supporting the conclusions of this article will be made available by the authors, without undue reservation.

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Author contributions

AH: Conceptualization, Data curation, Formal analysis, Investigation, Writing – original draft. HM: Conceptualization, Project administration, Supervision, Writing – review & editing. AI: Conceptualization, Validation, Writing – review & editing. MU: Supervision, Validation, Writing – review & editing. FN: Formal analysis, Writing – review & editing.

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Conflict of interest

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